



Enabling Emerging, Heterogeneous Memory Systems

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Enabling Emerging, Heterogeneous Memory Systems

A DISSERTATION PRESENTED

BY

LILLIAN COSTON PENTECOST

TO

THE DEPARTMENT OF COMPUTER SCIENCE

IN PARTIAL FULFILLMENT OF THE REQUIREMENTS

FOR THE DEGREE OF

DOCTOR OF PHILOSOPHY

IN THE SUBJECT OF

COMPUTER SCIENCE

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Enabling Emerging, Heterogeneous Memory Systems

ABSTRACT

Optimizing data storage and data movement remain critical roadblocks to overall computing performance and efficiency. These barriers are due to a convergence of motivating factors that begins with the memory wall and limitations of well-established memory technologies and design paradigms, and is compounded by increasingly data-intensive applications. This shift can and should be answered by many research thrusts to address the density and efficiency of memory systems, including technological heterogeneity in on-chip memory, increased specialization and cross-computing-stack choices in system and device design, and re-thinking the relationship and co-location of memory and compute resources. This thesis proposes and evaluates both specific solutions and broad design methodologies, uncovering enormous potential for increased memory system efficiency while unlocking numerous doors for further exploration and innovation.

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Previous Work

Portions of this dissertation appeared in the following works:

- L. Pentecost*, A. Hankin*, M. Donato, M. Hempstead, G. Wei, D. Brooks. "NVMExplorer: A Framework for Cross-Stack Comparisons of Embedded Non-Volatile Memories" IEEE International Symposium on High-Performance Computer Architecture (HPCA) Proceedings, April 2022. *Authors contributed equally.
- L. Pentecost*, M. Sharifi*, R. Rajaei, A. Kazemi, Q. Lou, G. Wei, D. Brooks, K. Ni, X. Hu, M. Niemier, M. Donato. "Application-Driven Design Exploration for Dense Ferroelectric Embedded Non-Volatile Memories (eNVMs)" ACM/IEEE International Symposium on Low Power Electronics and Design (ISLPED 2021), July 2021. *Authors contributed equally.
- T. Tambe, C. Hooper, L. Pentecost, T. Jia, E. Yang, M. Donato, V. Sanh, P. Whatmough, A. Rush, D. Brooks, G. Wei. "EdgeBERT: Sentence-Level Energy Optimizations for Latency-Aware Multi-Task Natural Language Processing Inference" MICRO-54: 54th Annual IEEE/ACM International Symposium on Microarchitecture, October 2021.
- E. Pyne*, **L. Pentecost**, U. Gupta, G. Wei, D. Brooks. "Quantifying the impact of data encoding on DNN fault tolerance" Workshop on Performance Analysis of Machine Learning Systems (FastPath at ISPASS 2020), August 2020. **Undergraduate Mentee*.
- **L. Pentecost***, U. Gupta*, E. Ngan*, G. Wei, D. Brooks, J. Beyer, M. Behrisch. "CHAMPVis: Comparative Hierarchical Analysis of Microarchitectual Performance Visualization" Workshop on Programming and Performance Visualization Tools at Supercomputing 2019 (ProTools at SC19), November 2019. **Authors contributed equally*.
- M. Donato, L. Pentecost, D. Brooks, G. Wei. "MEMTI: Optimizing On-Chip Nonvolatile Storage (eNVM) for Visual Multi-task Inference at the Edge" EEE Micro, Vol. 39, Issue 6, November 2019.
- L. Pentecost, M. Donato, B. Reagen, U. Gupta, S. Ma, G. Wei, D. Brooks. "MaxNVM: Maximizing DNN Storage Density and Inference Efficiency with Sparse Encoding and Error Mitigation" 52nd IEEE/ACM International Symposium on Microarchitecture (MICRO-52), October 2019.

B. Reagen, U. Gupta, L. Pentecost, G. Wei, and D. Brooks. "Ares: A Framework for Quantifying the Resilience of Deep Neural Networks" Design Automation Conference (DAC) 2018, June 2018.

M. Donato, B. Reagen, L. Pentecost, U. Gupta, D. Brooks, G. Wei. "On-Chip Deep Neural Network Storage with Multi-Level eNVM" Design Automation Conference (DAC) 2018, June 2018.

I was excited at something new, always liked something new, but give credit to everybody who helped. I didn't do anything alone.

Katherine Johnson, 2017

Introduction

Data movement and data storage remain a critical roadblock to overall computing performance and efficiency. Data-intensive workloads, often running on increasingly heterogeneous systems (i.e., computers that comprise more than one type of processing unit or more than just a general-purpose CPU), would particularly benefit from drastic increases to on-chip memory capacity and storage density, as well as the energy efficiency and performance of memory access, as

discussed in detail in Chapter 1.

This thesis identifies and explores game-changing design choices that could enable substantial gains in future memory systems, and also provides far-reaching approaches to identify additional design choices and optimizations. In a variety of application and system settings, my work reveals that exposing and co-optimizing design choices at different levels of the computing stack (for example, algorithmic optimizations alongside architecture and device-level choices) is the bedrock of future efficiency gains and effective future memory system design. For example, several works presented in this dissertation identify embedded, non-volatile memory (eNVM) technologies as a compelling opportunity to address on-chip memory efficiency, but the design space is complex and each memory cell configuration has unique characteristics that can and should be calibrated to system needs. This work has the dual contribution of proposing and evaluating specific memory system solutions, often specifically designing towards optimal on-chip memory for machine learning hardware accelerators, as well as providing reusable and extensible methods, tools, and takeaways to inform a broader class of future memory systems.

0.0 Publications & Research Experiences

Throughout this dissertation, I will present specific system designs and evaluations and the accompanying methodologies and tools that enable efficient, effective future memory system design and optimization. Much of this work is underpinned and informed by experiences early in my PhD contributing to the design, fabrication, and testing of specialized accelerators for deep neural network (DNN) inference, consistently in tandem with application-level optimizations ^{243,81}. I was also fortunate to complete research internships in the first half of my PhD that honed deep skills in workload characterization and hardware profiling at IBM (2016) and Microsoft Research (2018). Additionally, I contributed in the nascent stages of MLPerf, a benchmarking effort to standardize

and democratize expectations and conditions for measuring and reporting machine learning performance, in the form of discussions for benchmark selection and responsibilities in benchmark and dataset testing and validation ¹⁵⁹.

Through my development and study of various machine learning algorithms, and particularly in managing the training and optimization of different DNNs, the resilience of DNN parameters and execution to various forms of normalization and value manipulation inspired the development of Ares, an application-level framework for quantifying the resilience of DNNs¹⁹³. Ares proved extensible and expressive for a variety of application and fault-mode pairings, and I collaborated and advised projects leveraging Ares to study resilience of different number representations and data formats ¹⁹¹, sensitivity to hyperparameters and bit-flip errors during training ¹⁹¹ and DNN resilience to approximate multiplication techniques ¹⁸⁶. Crucially, I also spent time early in my PhD understanding and developing fault models for multi-level-cell (MLC) programmed embedded non-volatile memory technologies. Specific programming choices (e.g., number and spacing of programmed values into a memory cell) could be customized according to the number formats and encoding strategies I was already studying deeply to optimize DNN weight storage ⁶⁰, and Ares was once again leveraged and extended to represent and quantify application-level impacts of different DNN parameter storage schemes.

After observing that cross-computing-stack interdependent design factors and sensitivities to cell-level settings were clear determinants of potential memory solution efficiency ⁶⁰, it became clear that a system-level evaluation and more comprehensive design study could unlock the huge potential of MLC eNVMs, which led to the MaxNVM project ¹⁸³. The building blocks of MaxNVM, in turn, were sufficiently modular and robust, and I applied similar co-design strategies to exciting new application spaces and system contexts (e.g., multi-task DNN inference ⁵⁹, graph processing kernels ²⁰⁴, hybrid SRAM-eNVM memory systems, and natural language processing ²²⁶).

I continually developed and applied these building blocks, including a memory-fault-centric im-

plementation of the PyTorch Ares interface, in new contexts, revealing open research questions and complex design considerations along the way. I realized that the versatility and efficacy of a first-order design guidance tool could be an invaluable framework for a broader set of design studies. This realization spurred the formalization and development of NVMExplorer ¹⁸⁵. My background and experiences in data visualization platforms, including a web-based tool for visualizing microarchitectural performance counters ¹⁸⁴, made it clear that the wide adoption and accessibility of such a design space exploration framework would be boosted by providing the users from different backgrounds and expertise with comprehensive and interactive ways to explore, filter, and refine their memory system design solutions. The resulting framework, NVMExplorer, in turn quickly and effectively spurred additional collaborative efforts in optimizing device and circuit design choices towards application-level and system-level constraints and goals ^{204,226}.

O.1 THESIS CONTRIBUTIONS

Several key themes and takeaways for the broader discipline of memory systems architecture and computer architecture will emerge and recur throughout the discussion of my PhD work:

- Reliability: Careful consideration of hardware reliability must be a first-order design consideration in future memory systems, owing both to the unique properties of emerging technology solutions and application-level definitions of program accuracy and resilience.
- Designing towards heterogeneity: Specialized hardware components communicating and coordinating to form a single, efficient system presents unique challenges and opportunities to the computer architect and system designer.
- Flexibility: In customizing a memory architecture towards emerging technologies, system requirements, and application properties, the extreme efficiency made possible by specializa-

tion must be balanced by a sense of sufficient flexibility and generality to support application changes and broader use-cases.

- Technology-aware principles and design methodologies: Numerous emerging technology
 solutions hold promise for improving the efficiency of computing systems at every scale,
 particularly in the context of memory solutions, but integrating them effectively will require
 exposing and balancing their properties with creativity and collaboration.
- First-order design space exploration to inform and shape further study: Probing design considerations and quantifying their impacts at many layers of the computing stack is an intensive, complex task, demanding tools and methods to effectively and efficiently explore such trade-offs at a high level to guide further investigation or investment.

Through the course of my PhD, I've unearthed and evaluated a selection of promising architecture solutions garnering appreciable potential benefits and takeaways compared with competitive baseline systems, a few examples of which include:

- Fault-prone, highly dense emerging memory technologies can be effectively employed to store sparse-encoded deep neural network weights for image classification. However, there is a tension between the two: sparse encoding increases fault vulnerability, limiting the efficacy of a fault-prone dense memory. Employing a proposed, lightweight error mitigation scheme with faulty, dense memory and sparse encoding can provide higher storage density than any of the techniques in isolation, providing up to 29× memory area reduction compared to less-faulty memory solutions ¹⁸³ (Chapter 2.3).
- Work presented in this thesis also probes the viability of a specific, even-less-mature memory technology solution in multiple application contexts (image classification, natural language processing, social network graph search) and identifies device- and circuit-level opportunities

for innovation and co-design for efficient future memory systems 204 . For example, memory area can be reduced by over $12 \times$ and energy-per-memory-access by $2.6 \times$ relative to a competitive baseline without degrading accuracy for a natural language processing task as a result of customizing memory device properties and programming schemes (Chapter 2.4).

- A proposed and evaluated new memory system for NVDLA²¹², an industry-grade convolutional neural network (CNN) accelerator, used co-designed emerging memory technologies to unlock up to 3.5× lower energy per image classification task and 3.2× lower system power, enabling entirely on-chip ResNet50 inference in about 2mm² of total system area ¹⁸³ (Chapter 3.1).
- System-level performance and energy efficiency of embedded, multi-task image processing improves dramatically by integrating a hybrid memory system of both emerging and traditional memory technologies, again made possible by tuning application optimizations and system settings for compression and data format ⁵⁹. For example, energy per input image frame reduces by over 10× while reducing the memory area by about 3× (Chapter 3.2).
- Applying fault-prone, dense memory technologies towards other applications, such as natural language processing, similarly achieves significant efficiency benefits (e.g., reducing energy-per-input-sentence due to memory access by 66, 000×)²²⁶, while showing that the intuitions and infrastructure developed in this thesis and used to evaluate emerging technologies can be effectively generalized and empower efficient future studies (Chapter 3.3).

These evaluations are made possible by a selection of software tools, simulation frameworks, and benchmarking efforts I've built or contributed to over the course of my PhD, which are described and cited throughout this thesis. Many of these contributions and evaluation efforts culminated in the development of NVMExplorer ¹⁸⁵, a unified platform to explore the viability of emerging

memory technologies in specific application and system settings. The NVMExplorer approach reveals cross-stack dependencies and optimization opportunities, in addition to reproducing and expanding previous published studies, (e.g., ¹⁸³, ⁸⁴, ²⁰⁴). NVMExplorer similarly enables co-design studies of application properties, system constraints, and devices in order to bridge the gap between architects and device designers for future memory solutions. Example co-design studies in Chapter 4 reveal both opportunities and potential disconnects among current research efforts ¹⁸⁵.

0.2 THESIS ORGANIZATION

In Chapter 1, I present background on the fundamental costs and limitations of current on-chip memory solutions, as well as introducing the properties and potential of a selection of eNVMs that can be considered to address memory limitations in a variety of computing systems. Next, Chapter 2 introduces reliability as a key design consideration for the next generation of memory systems, describing the tools and modeling efforts I have contributed to throughout my PhD to quantify memory reliability in application-specific contexts and presenting results from several publications that highlight interdependent choices in data format, number representation, circuit- and device-level design choices, and algorithmic optimizations. Each of the studies in Sections 2.2-2.4 apply application-aware reliability analysis to improve memory system efficiency.

Chapter 3 deepens the findings of Chapter 2 by carefully co-designing specific system solutions for deep neural network inference across different application domains, optimization goals, and operating conditions. Together with Chapter 4, Chapter 3 details system evaluations that reveal significant energy/power benefits and increased storage density via comprehensive and technology-aware cross-computing-stack co-design. Chapter 4 goes above and beyond the evaluation of specific solutions by proposing and describing an open-source, modular design space exploration framework (NVMExplorer) to empower future collaborations and effective memory systems research

where application experts and device designers can similarly identify and contextualize different optimizations and system, algorithm, and technology choices. Chapter 5 revisits several key themes and takeaways, and proposes potential future contributions and research directions informed and enabled by the work presented in this dissertation.

1

Background: Memory Technologies & How to Leverage Them

EFFICIENCY AND INNOVATION in future computer systems is increasingly bound by the capabilities of the memory system. The convergence of fundamental scaling limitations, increasing hardware specialization and system heterogeneity, and a broad set of essential computing applications

that are voraciously data-intensive prompt consideration and investment in new memory design strategies and technologies. In this chapter, I motivate the need for fundamental breakthroughs and fresh approaches in improving memory systems, as well as introduce several compelling research directions for addressing the memory needs of modern and future computing systems.

1.0 Motivation: Limitations of Today's Memory Systems

Even before Moore's Law waned and ended, the slowing of scaling trends has been disproportionate in favoring the efficiency of compute vs. memory ²⁴⁹. Namely, the improvement in microprocessor speed was observed to exceed the improvement in DRAM memory speed year-over-year. In blunt terms, the inevitable conclusion of this trend is that the fastest, most efficient possible processing unit will result in no system-level performance benefit because it will be stalled, waiting on interactions with memory. The disproportionate scaling in terms of achievable power efficiency, data delivery, and cost-effective density is sometimes referred to as the 'memory wall' problem, and it motivates much need for reinvention and careful consideration of memory system design choices.

The memory subsystem has remained a crucial performance bottleneck in the intervening years of innovation in computing at different scales (e.g., the emerging prominence and diversity of mobile phones and embedded technology, as well as datacenter-scale computing), with critical applications facing limits in terms of both the availability and utilization of memory bandwidth. The quantity of data required by applications often out-paces the power, energy, and physical distance to store sufficient data capacity. As such, there are broad classes of applications that can be identified by whether their end-to-end performance is limited by the availability and utilization of compute resources (compute-bound) versus those limited by the availability and utilization of memory resources (memory-bound). The identification of compute-bound vs. memory-bound application behavior on a given system is sometimes visualized via the Roofline Model ⁹⁰, and analysis is typi-

cally conducted via microarchitectural performance analysis (described in Chapter 4.0.5).

In response to the explosive number and variety of data-intensive, often memory-bound applications that are critical in computing today (from facial recognition and image processing on mobile phones, to recommendation systems and graph processing in the datacenter, to a variety of scientific computing applications), this thesis focuses on innovating and intervening in the design of memory systems, with a particular eye towards heterogeneity in two senses: (1) the integration of technological heterogeneity on-chip, that is, emerging and alternative memory device technologies, and (2) customizing memory systems towards hardware accelerators and heterogeneous systems. To understand these trends further and to motivate the solutions and methods put forth in Chapters 2-4, I next provide some discussion and key terms in memory architecture and organization, then dive into the strengths and limitations of today's most pervasive memory technologies (SRAM and DRAM). Then, I describe the area, latency, and energy costs of memory access with a simple example and identify cell-level and memory architecture design trade-offs that will recur in later design studies. Finally, this section will touch upon some compelling research and industry efforts to address the memory wall that are outside the scope of this thesis.

1.0.0 SRAM, DRAM, AND THEIR LIMITATIONS

Different memory technologies find use in different parts of a traditional memory hierarchy due to their varying characteristics and varying system requirements. What are the most common memory technologies used in computing systems today, and how are they typically employed in memory hierarchies? Answering these two questions will motivate and inform what the research goals are for augmenting, replacing, or improving these technologies in future designs.

In Figure 1.0, different memory technologies and key terms are organized by their typical relative capacity and proximity to compute units, with some relevant characteristics listed on the right going from top to bottom. In most modern computing systems, the memory resources physically closest

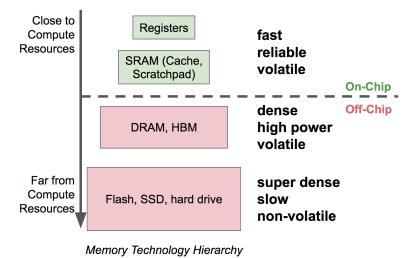


Figure 1.0: A majority of modern computing systems rely on SRAM for fast on-chip memory resources physically close to compute resources (green), rely on additional off-chip (red) memory resources with increased density (e.g., DRAM) during computation, and also incorporate higher-capacity, non-volatile memory sources for permanent storage, as described in Section 1.0.0.

to compute resources is implemented using SRAM, and is integrated directly on the same chip as compute resources for fast, efficient data access (shown in green in Figure 1.0). Higher-capacity memory resources are accessed via an interface to another fabricated component (e.g., a DRAM chip or Flash memory), referred to throughout this thesis as "off-chip" (shown in pink/red in Figure 1.0). SRAM and DRAM are volatile technologies, meaning they must constantly receive power to retain stored data values, while Flash is a non-volatile memory technology, meaning it retains data values in memory when powered off. While the relative capacities and protocols applied to each memory technology may vary depending on system context, the broad organization of on-chip SRAM resources for performant memory, off-chip DRAM for density, and a more physically distant, slower, non-volatile resource for permanent storage is visible in mobile phones, laptops, and datacenters alike.

SRAM provides reliable, fast on-chip memory, used as cache memory in a vast majority of mod-

ern systems of varying scales. However, one can only provision so much SRAM on-chip, and there are power and efficiency trade-offs when doing so. When data needs outstrip the capacity of on-chip memory, higher-capacity, off-chip resources repopulate the on-chip resources with the more frequently used or otherwise critical data for use by compute resources. An enormous amount of microarchitectural design effort and system-level innovation is poured into retaining, rejecting, or fetching the right data from off-chip at the right time in order to minimize data movement and/or maximize overall system efficiency due to the costs (energy, time or otherwise) associated with off-chip access. This is true both historically, in continued development of cache protocols for general-purpose systems at varying scales, and in specialized hardware, where energy and power due to data movement remains a key limitation to system efficiency.

The memory wall has already forced system designers to think beyond the memory technology hierarchy in Figure 1.0, and current and future generations of computing systems will require different tools, methods, and technologies to make meaningful breakthroughs. To achieve fundamental future improvements, particularly for applications that are heavily memory-bound and for specialized systems with unique data needs, system designers must bring denser, more efficient memory resources closer to compute resources.

1.0.1 MEMORY ORGANIZATION AND MEMORY ARRAY ARCHITECTURE

This section summarizes some fundamental costs of storing and retrieving data. With SRAM properties and limitations as context, we can zoom in on an example of the costs associated with data storage and data movement in a memory array architecture. What are the area, energy, and latency trade-offs when memory devices are organized into subarrays, banks, and arrays? This discussion will introduce architectural design choices and trade-offs that are unique to each proposed physical mechanism for storing bits. As alternative memory technologies for on-chip storage solutions are introduced and evaluated in future chapters, some array architecture characteristics introduced here

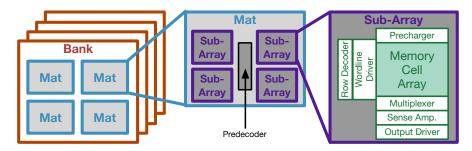


Figure 1.1: A memory array is comprised of a set of banks (left), each of which in turn may exhibit a hierarchical organization into mats and sub-arrays. Within each sub-array, there is additional periphery circuitry to read and write data in and out of a set of memory cells (right), adapted from ^{60,62}.

will be revisited and innovated in the context of optimizations that consider memory-device-level attributes, all the way up to practical, application-level impacts.

There are a variety of architectural choices when organizing memory cells and sub-arrays into larger blocks of accessible memory, some of which are evident in observing a basic memory array organization in Figure 1.1. An array of individual memory devices with corresponding circuitry to read and write data values is labeled as a sub-array, and those sub-arrays are, in turn, hierarchically organized into mats and banks. Large cell arrays can consume significantly more power and take more time to read/write to, so provisioning a block of memory using many smaller sub-arrays with a network of multiplexing and routing to determine when and how to fetch data can reduce the average energy and latency per access. Hierarchical organization choices can also enable simultaneous parallel access to multiple columns of data, sub-arrays, mats, or banks, improving the bandwidth of the overall memory. Sub-array dimensions, column muxing and complexity of selection circuitry, and hierarchical organization into mats and banks are just a sampling of design choices that impact the area efficiency, latency, and energy of access to the memory array. These trade-offs will vary per-technology, and can be tuned according to the system setting and use case. Figure 1.1 shows a schematic example of the memory organization considered by a previously-validated memory array simulation tool 62, applicable both in the context of an SRAM array and for embedded, non-volatile

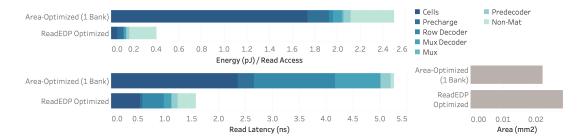


Figure 1.2: For a small (8KB) SRAM array under two different organization choices (simulated using NVSim ⁶²), different memory organization choices impact the area efficiency, latency, and energy of access. Organizing the 8KB array as a single bank (Area-optimal) requires less area, while using 16 smaller sub-arrays (the read-energy-delay-product-optimal (ReadEDP) choice) drastically reduces energy and latency of access at the cost of additional periphery circuitry and routing from sub-array to port. The energy and latency per access can be further broken down into the contributions from individual architecture components, as labeled in Figure 1.1.

memory technologies (eNVMs).

Figure 1.2 presents a simple example of how the components of a memory array contribute to the access energy and latency for a simple example under two different architectural organizations. An area-optimal array organization uses a single bank; the resulting read latency and energy-per-access are then dominated by the subarray components. In constrast, a read-energy-delay-product-optimal organization will drastically reduce the latency and energy of access by fracturing the array into multiple banks and subarrays. As a result, the area slightly increases due to duplicated sensing circuitry and network overhead to route data from subarray to port, and non-subarray energy and latency comprises a larger portion of total access costs. Thus, memory architecture choices and optimization goals heavily determine the performance and efficiency of memory resources, and application performance and efficiency, in turn, can be sharply impacted by these choices. The precise trade-offs and optimal memory architectures will vary significantly depending on array capacity and optimization goals. These design considerations are made even more complex by how periphery overheads and cell characteristics differ for alternative, emerging memory technologies discussed in Section 1.2.

1.0.2 3-D Integrated Memory Solutions

Another key innovation that has been deployed in several contexts is using the third dimension (e.g., stacking chips on top of one another and connecting them vertically) to improve the effective density and potential bandwidth of memory systems. This is a well-established technique in the context of stacked DRAM using through-silicon-vias (e.g., Hybrid Memory Cube⁷⁵) or the proliferation of High-Bandwidth-Memory (HBM) for increased bandwidth in GPU systems (graphics processing units, e.g., ¹⁵⁸). The radical increase in effective density and bandwidth provided by 3D-stacked memory solutions has also been applied towards specialized systems for deep neural networks (e.g., ⁷³). However, these performant, high-density solutions remain power-intensive, suffer efficiency limitations, and many, being DRAM-based, are volatile memory solutions.

3D-integration techniques have also been demonstrated as compatible with several of the technologies presented in Section 1.2^{247,41}. Though the studies presented in this thesis do not consider 3D-integrated memory solutions, typically due to the energy and power constraints of edge and embedded computing contexts considered, there are many compelling memory system design directions left for future work described in Chapter 5.

1.0.3 Processing-In-Memory, Processing-Near-Memory Architectures

For several of the technologies leveraged in this thesis, their analog device properties allow them to be used effectively as processing elements ^{222,50,105,89}, as discussed in Section 1.2. For example, researchers have crafted circuit and array architectures to add or multiple input currents as values, and applied such systems towards neural network inference and training ^{42,8,201}. Additionally, there is a long history of investigation towards near-data processing (i.e., co-locating compute resources or performing computation in a memory process like DRAM to bridge the distance and unlock efficiency for data-intensive tasks ¹⁴⁰). While these approaches are exciting disruptions of the traditional

memory hierarchy (e.g., Figure 1.0) and have seen application in some targeted settings, they are accompanied by distinct manufacturing and scaling challenges.

Processing-in-memory and near-memory will remain important and exciting paradigms in the future, but the challenges to their scalability, capacity, and density are somewhat orthogonal to the study of emerging materials innovations and devices as memory cells. Though understanding and methodologies for design space exploration at the device and circuit level of emerging memory devices, such as are presented in the following chapters, could be incredibly informative towards in-/near-memory architectures in the future, specific evaluations fall outside the scope of the work presented in this thesis. As the next section will highlight, the size and data-intensiveness of mission-critical applications necessitates sheer density and energy efficiency of on-chip memory and data delivery more broadly. Thus, the scope of this thesis aims to be creative in augmenting and improving memory systems, while keeping them distinct from compute elements, to best complement the enormous diversity of modern heterogeneous computing systems.

I.I MOTIVATION: DATA-INTENSIVE APPLICATIONS

In response to the end of Moore's Law, architects and system designers have gravitated towards specialized hardware solutions and increasingly heterogeneous computing systems in order to achieve the next generation of performance and power advantages over previous systems. The past decade has seen an explosion of investment and interest in specialized hardware and heterogeneous computing platforms. As this trend continues, customized memory system support will continue to be essential. In this Section, I will begin by briefly reviewing examples of heterogeneous systems and the importance of hardware accelerators (Section 1.1.0). Next, I describe the specific needs and ongoing architectural innovations around two broad classes of data-intensive workloads that are the target of studies presented in this dissertation: Deep Neural Networks (DNNs, Section 1.1.1) and

the processing and analytics performed on very large graphs (Section 1.1.2).

1.1.0 HARDWARE ACCELERATORS, HETEROGENEOUS SYSTEMS

Specialization allows for significant gains in compute efficiency. Identifying applications that will benefit from specialization and whose acceleration will provide benefits to the overall system setting has become a touchstone of computer architecture research and future system efficiency. For example, datacenter architectures now rely on a huge mix of server-class CPUs, GPUs, FP-GAs, other accelerators, and various network and memory interfaces to serve user requests, and high-performance-computing has similarly benefited from the integration of hardware accelerators ^{87,111,114}. Alternatively, resource-constrained and battery-powered devices (e.g., mobile phones, deployed sensor nodes, drones, self-driving cars) have similarly shifted towards a model of many accelerator blocks and integrated components, for example in a system-on-chip (SoC). In fact, the number of identifiable, specialized IP blocks in Apple iPhone SoCs has increased dramatically over the last decade, from fewer than 10 in the A4 to over 40 in the A11 (iPhone X), comprising a majority of on-chip area ²⁰³. These blocks need to maintain and access data, as well as communicate with one another, and data transfer and data management become even more critical (e.g., ^{155,182}).

These specialization efforts at varying scales of computing power and complexity require customizing computing blocks (i.e., accelerators) and end-to-end systems to the needs and characteristics of the most pervasive or otherwise bottlenecked pieces of software that will execute on these systems. The efficiency and effectiveness of specialized hardware accelerator blocks rely on their understanding of application-level properties (i.e., which patterns of computation, memory access, and individual operations to support), as well as sufficient programmability and flexibility (i.e., if the target application changes slightly in terms of size or pattern, is the accelerator still useful?). Thus, cross-computing-stack design considerations and first-order understandings of application-level impacts of system designs become necessary tools in the face of heterogeneous system design and

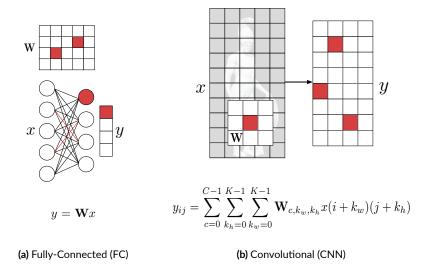


Figure 1.3: Two common types of neural network layers, either fully-connected (left) or convolutional (right) can be summarized as operations on an input (x) using trained parameters referred to as weights (W) to produce a transformed output (y), adapted from (y), adapted from (y), adapted from (y), and (y), and

optimization.

The rise of heterogeneous computing platforms is similarly reflected in the sheer volume of academic publications, silicon-based start-ups, and industry investment in hardware accelerators for a range of target applications (indisputably dominated by innovations and investment in machine learning) 192,159. In these myriad proposals, dense data storage, efficient data movement, and minimizing reliance on off-chip memory resources remain essential challenges. Thus, innovation and improved efficiency in on-chip memory would provide drastic improvement to systems with different target applications, different design constraints, and different scales of operation.

1.1.1 DEEP NEURAL NETWORKS

Deep learning is a field of machine learning that leverages large neural networks (often referred to as Deep Neural Networks or DNNs) to tackle challenging classification and regression tasks. In recent years, DNNs have become essential to a huge breadth of application domains, including

image recognition and detection, language processing, and translation. DNNs have two operating modes: training and inference. Training is the process of fitting neural network parameters (weights) to labeled data. Inference is the process of using a previously trained DNN to predict labels for new input data. Figure 1.3 provides highlighted examples of the inputs, weights, and outputs, as well as the equations describing the basic operations of two common types of neural network layers.

Perhaps the most prominent example of the co-development of data-intensive applications and specialized hardware solution has been the seismic rise of acceleration of tensor operations in machine learning algorithms and in DNNs in particular. Machine learning researchers and engineers over the past two decades realized that the computation underlying their algorithms was often deployable and exhibited enormous performance benefits when offloaded to a graphics processing unit (GPU). Where neural-network based solutions were previously impractical due to their intensity and runtime, neural network algorithms trained using GPUs began outstripping state-of-the-art metrics in classification and prediction tasks 197,88. GPU manufacturers like NVIDIA responded in force by building out support and tailoring GPUs to the needs of such algorithms, leading to a proliferation of GPU-based machine learning acceleration. This increase in popularity, together with the continued proliferation of low-power embedded devices, has similarly motivated the design of DNN-specific hardware accelerators ^{192,111}. While many energy-efficient DNN hardware implementations have been proposed, a major challenge remains: the large memory requirement to store DNN parameters. Although entirely on-chip or on-device storage would guarantee better inference efficiency, limited on-chip SRAM capacity inevitably leads to reliance on costly off-chip memory accesses to DRAM or cloud-based execution of tasks like speech recognition.

An incredible body of work and many continued research efforts over the past decade have targeted reducing the computational complexity and memory requirements of DNNs. These techniques include innovations in data format (such as reduced precision even down to binarized DNNs), model architectures (such as reducing depth, or removing trained parameters), and training

techniques (such as transfer learning, adaptive techniques) ¹⁹². However, the overarching trend of state-of-the-art proposals is a staggering increase in memory requirements to push the boundaries of machine intelligence, as well as sometimes unforeseeable changes in data flow and data access that complicate deployment to existing, fixed neural network accelerators ¹²⁶, ²²⁶. Thus, the continued development and specialization of systems to support and accelerate DNNs and related, data-intensive applications must balance flexibility in the face of future trends with the persistent need for increased memory capacity and efficiency. Careful consideration of algorithmic advancements and memory-reducing optimizations play a key role in maximizing the effectiveness of system and architecture decisions, as explored in detail in future chapters.

The myriad proposals for reducing DNN size and complexity often degrade the achievable application-level accuracy (e.g., reducing the likelihood of correctly classifying input images by up to 10% while reducing storage requirements by an order of magnitude ⁹⁴). While many efforts have considered trading accuracy for efficiency in deep learning systems, the most convincing demonstration of a new technique for a practical system must address and preserve a DNN's baseline model accuracy. One proposal for quantifying and bounding acceptable variation in the inherently stochastic baseline performance of DNN's on a given task is **Iso-Training Noise (ITN)** ¹⁹², which is employed in several studies presented in this dissertation (e.g., ^{193,183}). The intuition for this method is that accuracy varies for DNNs repeatedly trained with identical hyperparameters. The resulting variance in the accuracy can be used as a bound for final classification error. As long as model alterations do not result in error exceeding this bound, they are said to be indistinguishable from ITN and therefore iso-accuracy. Examples of ITN for different DNNs and application of ITN as a reasonable bound on acceptable model accuracy are found in Chapters 2.0, 2.3, 2.4, and 3.1.

1.1.2 GRAPH PROCESSING

Analytics and processing tasks on very large graph structures underlie a variety of important and emerging applications today, including social network analysis, data mining, computational chemistry, and, perhaps most recently and prominently, COVID-19 drug discovery ^{17,13,135,262}. There has been a recent rash of interest and exploration of graph processing acceleration and optimization in the systems research community ^{55,83,265,39,11,54,54,13}. These works span system settings, from FPGA-based acceleration of graph kernels ¹¹ to hardware accelerators ⁵⁴ to general-purpose cache replacement policies ¹³. Each of these works highlights and designs towards the large memory capacity and bandwidth required for efficient execution of graph processing ^{17,83}.

The challenges of customizing the memory architecture to support graph processing in a variety of system settings are similar to deep neural networks purely in that they are both data-intensive in terms of bandwidth and capacity. Otherwise, at least in broad strokes, the challenges of graph processing are quite distinct. Namely, memory access patterns exhibit much more randomness than the embarrassingly predictable patterns when operating on a fixed neural network model architecture, and graph-related kernels (e.g., search, extracting graph properties) tend to require a more balanced mix of read-write traffic than DNN use cases considered in this thesis. Additionally, the resilience of graph processing to different value manipulations and fault models is less thoroughly studied than for DNNs, as discussed and probed in Chapters 2.4 and 4.2.

1.2 EMBEDDED, NON-VOLATILE MEMORY (ENVM) TECHNOLOGY LANDSCAPE

This section briefly reviews the device physics and relevant low-level details of various embeddable, non-volatile memory technologies. Each technology offers unique design challenges and opportunities as a storage device, and are at widely varying stages of development and investment across industry and academic efforts. More detailed surveys and comparisons of subsets of these technologies

are helpful resources (e.g., ³¹) in understanding the capabilities and physical mechanisms for each eNVM proposal. The following sections will briefly describe each technology, with an emphasis on both promising and limiting characteristics of each technology. For more established technologies (i.e., RRAM, PCM, and STT), I reference compelling recent examples and mention their broad characteristics, while for more emerging solutions (i.e., FeFET, CTT), I provide some discussion of their physical design and operation. Finally, this section also discusses the eNVM landscape as a whole, including contextualizing the potential benefits and limitations of different technologies and describing multi-level-cell programming.

1.2.0 RESISTIVE RAM (RRAM, RERAM)

Resistive RAM (RRAM) encompasses a wide variety of cell-design implementations that all encode data via the variable resistance of a thin layer of material 92,121,190,109,79, most frequently a metal oxide 254,255,49,128,29. RRAM is particularly compelling and relatively mature among eNVM proposals, having been demonstrated by major industry fabrication facilities at relatively mature technology nodes 48,77,104. Previous and contemporary work has also employed RRAM to neural network acceleration 239,95,183, including leveraging the variable resistance readout for analog compute capabilities in neuromorphic and compute-in-memory architectures 247,235,38,261,162,215,141,42,201. Such processing-in-memory implementations for deep neural networks are part of a broader class of research probing the analog computation capabilities of RRAM cells 147,252,68,222,36,246,50,149,21.

There are compelling RRAM solutions using either diode access for crossbars or CMOS access for more traditional memory array architectures. Crossbar arrays offer the best cell area (as low as 4F²), but they are subject to higher access times than using a dedicated CMOS transistor to access the RRAM cell ^{258,120,12,148}. The larger cell area can be overcome by increasing storage density via multi-level-cell (MLC) programming ^{209,33,251,37,189,146,269,268,53}. RRAM arrays have also been demonstrated in 3D-integrated settings (e.g., 3D-stacked dies of RRAM, vertical integration of

RRAM) ^{236,153,23,96,139,138,250}. However, RRAM device and array design is still an area of active development and innovation, with constantly evolving studies of reliability and endurance ¹⁷¹, in addition to alternative cell configurations and array architectures ^{45,18,263,180,4,154,15,145,150,143,156,241,151}.

1.2.1 Phase-Change Memory (PCM, PCRAM)

Phase-Change Memory (PCM), like RRAM, relies on a variable resistance in a material to encode a value. However, PCM relies on the transition between two phases of the target material, namely crystalline (low resistance) or amorphous (high resistance). PCM arrays have been the subject of study and demonstration for a longer period of time than most other eNVMs considered, and have been demonstrated in 20nm-90nm technology nodes ^{248,47,46,10,218}. PCM similarly requires careful study of reliability, endurance, and cell structure and fabrication variations ^{40,43,152,118}, and has been recently probed as compatible with 3D-integration ⁴¹, processing-in-memory architectures ⁷⁴, and multi-level-cell-programmed implementations ^{78,119,221,52}.

1.2.2 FERROELECTRIC-BASED MEMORY CELLS (FERAM, FEFET)

While Ferroelectric RAM (FeRAM) is a more mature memory solution with a DRAM-like structure relying on a ferroelectric capacitor ^{179,259}, FeFET-based eNVM is an emerging, embeddable technology that achieves much higher density by modifying a FET with a layer of ferroelectric material ^{161,58,174,166,196,175,108,231}. More precisely, the FeFET device is made by replacing the normal gate dielectric in a MOSFET with a ferroelectric layer ¹⁰⁸. For this reason, FeFET devices can be easily integrated in existing CMOS processes, especially when high-*k* dielectrics such as hafnium oxide can be used as the ferroelectric layer ¹⁶¹.

FeFET devices can be programmed to different values through partial polarization switching of the ferroelectric layer, and have been demonstrated as CMOS-compatible memory devices down to 14nm ^{205,27,227,64,124,131,132,122,70,206}. This is achieved by tuning the portions of the switched ferroelectric domains in a multi-domain FeFET, and this mechanism can be leveraged to program a single FeFET as a multi-level-cell memory device ^{5,58}, for analog compute capabilities ^{6,223,168,107}, and 3D integration ⁶⁹, though these are recent efforts under active research development. Recent works ^{5,166,168,64} use FeFETs for ultra-dense, low-leakage, and fast memories.

1.2.3 Magnetoresistive RAM (Spin-Transfer, Spin-Orbit Torque)

Two popular implementations of magnetoresistive RAM (MRAM)^{65,169,76} rely on different physical mechanisms to encode data via the spin of electrons in a magnetic tunneling junction (MTJ) element ^{214,142,16,110}. Spin-Transfer-Torque (STT) is the more mature of the two, with compelling applications for last-level-cache ^{170,98,7,229,177,176,106}, other embedded and storage-class demonstrations ^{133,3,173,57,113,25,51,230,32}, and explorations into their applicability in near-memory-computing contexts ^{30,105,89}. By contrast, Spin-Orbit Torque (SOT) is an emerging solution with exciting potential for dense, performant storage rivaling STT characteristics and capabilities, though current implementations tend to be less dense and their scalability is not as proven ^{172,93,71}. Over the past six years, the studies in this thesis have focused on STT as a promising eNVM solution in multiple system contexts, though future work should incorporate SOT more centrally in the design space of eNVM options. STT-RAM is under active development to verify and improve reliability ^{238,44,72,256,237,217,207} and scalability ^{164,22,240,129,61,130,134,216,211,198}, as well as probing the possibility of different device and array architectures and MLC-programming via vertically stacked MTIs ^{208,199,97,228,19,267,266}.

1.2.4 Charge-Trap Transistors (CTT)

Previous work has shown that a single, standard-sized NMOS device can be used as a cost-effective embedded non-volatile memory cell 117,116,115,165, including multi-level-cell 157 and process-in-memory capabilities. Data is stored in the resulting memories, often referred to as charge trap transistors (CTTs), by trapping charge in the gate oxide using hot-carrier injection, which alters the threshold voltage of the device. Charge trapping impacting the threshold voltage is itself a well-studied effect typically indicative of device aging 165.

As a result, a single transistor can be programmed to exhibit different saturation currents, which are read out and decoded as distinct stored values, as discussed in Section 1.2.6 and studied in detail in Chapter 2.1. In this way, CTT-based memory arrays can be fabricated in industrial-grade, cutting-edge standard CMOS technologies with zero added manufacturing cost. In addition to low cost, CTT has many other desirable properties: the memory arrays are analogous to programmable NOR ROM arrays and have similar-scale read latency; devices display very low leakage currents, as stored information is preserved in the transistor's threshold voltage with high retention independent of the applied voltage; and as each cell consists of only a single NMOS transistor, density can be extremely high.

The benefits of CTTs come at the expense of two undesirable properties, limiting their deployment and rendering them ineffective for many general-purpose compute tasks: (1) extremely long write-latencies and (2) potential for high cell fault rates. Writes involve changing the physical device properties, which takes 100ms or more. Write latency is long because CTTs are programmed by iteratively injecting increments of charge and reading until a desired shift is achieved. This is an imperfect process – charge injection is a random process that inevitably leads to a spread in current distribution, and this spread translates to increasing likelihood of misreading the stored value. Work presented in Chapters 2 and 3 demonstrate how CTT may yet be useful for on-chip MLC weight

1.2.5 OTHER MEMORY TECHNOLOGIES (EDRAM, ROM, ETC.)

There are additional alternative memory technologies with varying advantages and limitations. For example, the storage requirements for the DNN architecture for deeply embedded inference, where weights need not be re-written, could be met by read-only memories (ROM) as well. ROMs ensure the best density for storing read-only parameters. However, ROMs also require configuring the network at fabrication time, which makes the design less scalable and cost-effective. One-time programmable (OTP) memories such as anti-fuse, while being amenable to post-fabrication configuration, are far less dense than other memory solutions, even when compared to SRAM.

Embedded DRAM (eDRAM) is another compelling response to the need for dense and performant on-chip memory, and over the past 10-15 years eDRAM has seen deployment in industry products (e.g., the XBox 360°) and, more recently, in accelerator systems for data-intensive applications 35,83. While eDRAM continues to be developed, optimized, and deployed, issues of technology scaling persist, and the complexity and power-intensiveness of DRAM refresh preclude its use in energy-constrained devices and environments. In contrast, embedded non-volatile memories typically have extremely low leakage and can be powered off without losing data, while still offering increased density, adequate performance, and CMOS-compatibility. In fact, several of the eNVMs discussed require few to no CMOS process changes compared with the complexity and potential fabrication issues of embedding DRAM.

It is also possible to embed traditional flash memory (eFlash), which is provided as a production-grade on-chip memory solution by top industry foundries (e.g., for IoT settings from TSMC ²³²) despite known issues and large overheads due to reliability of eFlash technology ^{210,253}. eFlash has been deployed as a compelling solution for increased on-chip density in IoT and automotive applications, but has several key limitations to wider adoption and future development ¹⁴⁴. Most promi-

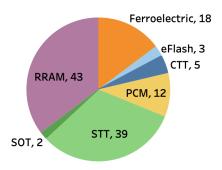


Figure 1.4: Number of ISSCC, VLSI, and IEDM papers published per eNVM technology, 2016-2020, showing sustained interest in RRAM, STT, and emerging interest in ferroelectric-based devices, taken from ¹⁸⁵.

nently, eFlash has not proved compatible with advanced technology nodes, and it exhibits markedly higher power consumption and typically worse read/write performance than RRAM, MRAM, and other eNVMs considered in this work.

1.2.6 MULTI-LEVEL NVM STORAGE

Storing multiple bits in a single memory cell (multi-level-cell, MLC) is desirable for increasing storage density, and has been demonstrated using many eNVM devices, including resistive random access memories (ReRAM), phase change memories (PCMs), charge-trap transistors (CTT), and FeFETs^{31,115,183}. In order to discriminate among the different programmed levels, the memory sensing circuit needs to perform an analog to digital conversion (ADC), decoding each detected level of current or resistance from the memory cell to the corresponding binary word ⁵³. However, the relatively high area and power overhead of the ADC can limit the relative benefits of MLC implementations compared to single-level-cell (SLC) storage. Thus, MLC NVM storage, while offering exciting opportunities for increased density, introduces performance/power/area trade-offs and may be accompanied by reliability issues, as studied in detail for several technologies in Chapter 2.

1.2.7 DISCUSSION: OPPORTUNITIES AND LIMITATIONS

In evaluating the varied landscape of non-volatile memory devices, we are interested in identifying implementations that achieve low read latency, high storage density, and proven ability to scale to advanced process nodes. Figure 1.4 summarizes the number of publications pertaining to each flavor of eNVM over 5 years of the top device and circuit conference venues, including both academic and industry demonstrations and manufactured examples. There is clear variety and active research in the design, optimization, and application of eNVM solutions, but there is a severe imbalance of available data and detail across these technologies according to device maturity, corporate transparency, and the potentially limited scope of demonstration in each published example.

For a sampling of memory-array-level capabilities of eNVMs, Table 1.0 shows several published and validated examples, spanning different technology nodes (90nm to 20nm) and architectures (crossbar vs. CMOS-access), taken from 183. Several RRAM solutions have been demonstrated in relatively advanced technology nodes, and CMOS-access architectures for embedded applications indicate compelling storage density and read characteristics (i.e., lower area and comparable read latency compared to iso-capacity SRAM). STT strictly improves storage density and read/write performance compared to RRAM and PCM owing to demonstration in a more advanced technology node (28nm) and higher area efficiency despite larger cell footprint. The advantages of PCM are less uniform; though 40nm PCM appears to exhibit lower write latency than similar-capacity RRAM examples, the publication lacks sufficient data to contextualize this with the achievable density and read characteristics. This incomplete context and inconclusive comparison at the device or array level is representative of a key research challenge in trying to identify and evaluate the most relevant and most compelling memory solutions for a particular use case.

To better contextualize and more precisely evaluate the trade-offs among eNVM proposals, Figure 1.5 shows simulated, iso-capacity 4MB memory array characteristics derived from pub-

Reference	eNVM type	Tech. Node	Access	Cell Area	Capacity	Array Area	Read latency	Write latency
29	RRAM	28nm	CMOS	39 F ²	ıМb	0.56 mm ²	6.8ns	500ns - 100 <i>µs</i>
128	RRAM	40nm	CMOS	53 F ²	1.4Mb	0.28 mm ²	ions	_
2.3	RRAM	24nm	diode	4 F ²	32Gb	130.7 mm ²	40µs	230µs
52	MLC PCM	90nm	CMOS	25 F ²	256Mb	120 mm ²	320ns	_
47	PCM	40nm	CMOS	_	ıМb	_	_	12ons
46	PCM	20nm PRAM	diode	4 F ²	8Gb	59.4 mm ²	1 20ns	150ns - 100 <i>µs</i>
256	STT	28nm	CMOS	75 F ²	ıМb	0.214 mm ²	2.8ns	20ns

Table 1.0: Characterization of several example non-volatile memory arrays circa 2017, adapted from ¹⁸³.

lished characteristics, optimized for read energy-delay-product in NVSim ⁶². We also include an optimistically-scaled version of a RRAM array based on a $10F^2$ cell size as a way of evaluating the maximum potential of promising technology advances, which is shown scaled to several process nodes in Figure 1.5. As demonstrated in proceeding studies (e.g., Section 3.1), compromising on the read latency may not be acceptable for DNNs and other data-intensive use cases, as this could reduce available bandwidth. While the need to accommodate larger access devices requires some area overhead, the arrays implemented using CMOS access devices provide read latencies more in line with potential application requirements. Based on these iso-capacity results, we can identify a subset of eNVM solutions that maximize density and maintain read performance in future evaluations.

To push the boundaries of dense on-chip storage, we are interested in evaluating memories with the capability of storing multiple bits in a single cell using MLC storage. In fact, the projected memory array characterizations shown in Figure 1.5 include several MLC design points. MLCs encode multiple bits of data per cell, which offers maximal storage density with minimal read latency and read energy for the examples shown compared to corresponding SLC (one-bit-per-memory-cell) solutions. However, MLCs come with design challenges (e.g., reliability issues, as studied in Chapter 2) and additional opportunities for optimization (Chapter 3). Both the opportunities and limitations of results in Figure 1.5 lay foundation for the modeling approach that is proposed, expanded, and developed into a more comprehensive design space exploration and evaluation framework in Chapters 3 and 4.

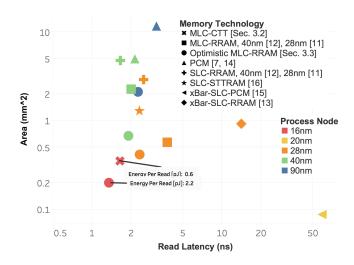


Figure 1.5: Characterization of published and extrapolated eNVM proposals comparing area and read latency for a fixed capacity (4MB) using read-energy-delay-product-optimized results from NVSim ⁶²; section and citation numbers are to references in original publication context, taken from ¹⁸³.

Table 1.1 summarizes device characteristics per-eNVM extracted from surveyed publications in 2016-2020 (Figure 1.4). This data, in turn, informs a library of cell characteristics and technology capabilities further developed and leveraged in Chapter 4. For each eNVM listed, we consider a range of published properties of array-level and device-level characterizations. Parameter ranges for a subset of parameters that are particularly relevant to guide system design and determine viability are shown for illustration purposes in Table 1.1.

In Table 1.1, there are several key characteristics that an architect will care about for a memory technology, including the range of demonstrated cell area [F²], demonstrated fabrication process [nm], endurance [write-cycles before degradation], and data retention [s]. It is important to note that the technology classes considered are at different levels of maturity. For example, SOT is a relatively recent technology, and while it boasts very impressive write speed and lower write current compared to STT, published data is extremely limited. We also see that endurance varies by multiple orders of magnitude across different technologies. Thus, adoption will depend on the write

	SRAM	PCM	STT	SOT	RRAM	CTT	FeRAM	FeFET
Cell Area [F ²]	146	25-40	14-75	[20]	4-53	I-I2	-	4-103
Tech. Node [nm]	7-16	28-120	22-90	[1000]	16-130	14-16	40	45
Read Latency [ns]	0.5-1.5	[1-100]	1.3-19	1.4-11	3.3-2e3	-	14	-
Write Latency [ns]	0.5-1.5	10-3e4	2-200	0.35-17	5-1e5	6e7-2.6e9	14-1e3	0.93-1.3e3
Read Energy [p]	1.1-2.4	-	0.21-1.2	-	1e-3	,	0.001	-
Write Energy [p]	1.1-33	-	0.6-4.5	[0.015-8]	0.68		-	0.0003-0.01
Endurance [Cycles]	N/A	105-1011	105-1015	-	103-108	104	104-1011	107-1011
Retention [s]	N/A	108-1010	108	108	103-108	108	105-108	-

Table 1.1: High-level listing of memory cell technologies and ranges for key characteristics; recent publications are complemented by simulation and industry references to form technology cell definitions discussed, indicated by [X], though some emerging technologies have limited available data for certain metrics, indicated by '-', adapted from ¹⁸⁵.

intensity of target applications and system dynamics, so incorporating memory lifetime estimation becomes a critical design consideration.

For some technologies considered, the survey of published examples 2016-2020 provided insufficient data to conduct cell-level characterization, or even to fill in some cells in Table 1.1. Insufficient data could be for reasons of propriety from industry fabrication, waning interest in a given technology resulting in fewer up-to-date publications, or experimental constraints. However, to guide system design, it is critical to have some concept of the possible range of values associated with these parameters. In those cases where SPICE models for a technology are available, we used device and circuit simulations to fill in missing parameters. Alternatively, we consider older publications (dating back to 2009) and consult with device experts to reason about cell and array parameters. This includes identifying either optimistic or pessimistic ranges of achievable cell characteristics to establish reasonable bounds of the design space per technology, as described in Chapter 4.

1.3 IDENTIFYING CROSS-STACK DESIGN CONSIDERATIONS

Data-intensive applications such as DNNs and graph analytics have emerged as dominating work-loads in the current computing landscape and have influenced many research efforts and advances in computer architecture in the past few years. Specialized hardware architectures deliver outstanding

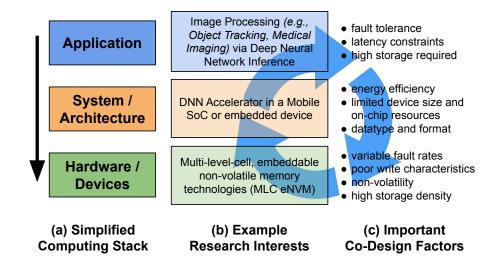


Figure 1.6: The optimizations and opportunities, as well as the design methodologies and frameworks, proposed in the following chapters both expose and exploit the interdependent nature of design choices at the memory device, architecture, system, and application level of the computing stack.

performance and computational efficiency. However, the size and complexity of critical workloads continues to outstrip available on-chip memory, making data movement and efficient on-chip storage a challenge in scaling these applications. Efficient future memory systems must leverage information across the layers of the computing stack – that is, identifying and exploiting opportunities to alter the applications, the system architecture, and the hardware/devices in a collaborative manner to improve end-to-end metrics like energy-efficiency, performance, and accuracy. Design considerations and optimization goals become interdependent across the levels of the computing stack, as pictured in Figure 1.6, highlighting example research interests and important factors for co-design methods that are featured in the remainder of this thesis.

Embedded non-volatile memories offer a promising technology alternative to traditional onchip SRAM³¹. Smaller memory cell size and the ability to store multiple bits in a single memory cell can translate to a denser array implementation, which can be optimized to store entire DNNs on-chip ¹⁸3,59. In addition to storage density improvements, non-volatility increases energy efficient due to lower leakage power and the ability to retain information in power-off state for intermittent computing. The advantages of eNVMs are often countered by lackluster write performance and reduced reliability, which may be exacerbated with multi-level cell programming.

In today's varied application space, meeting different memory read and write access patterns requires careful consideration of eNVM memory characteristics. An application-driven approach can provide a clear path towards design optimization targeting a specific application and coincidentally offer insight regarding which device-level specifications such as reliability, endurance, write performance, and cell area should be further improved to serve a wider set of applications. Discussions in the following chapters will present both application-driven studies to guide better device development and careful system-level optimizations and evaluations to determine the viability of different memory solutions in specific system contexts.

"Marilla, isn't it nice to think that tomorrow is a new day with no mistakes in it yet?"

"I'll warrant you'll make plenty in it," said Marilla. "I never saw you beat for making mistakes, Anne."

"Yes, and well I know it," admitted Anne mournfully. "But have you ever noticed one encouraging thing about me, Marilla? I never make the same mistake twice."

"I don't know as that's much benefit when you're always making new ones."

"Oh, don't you see, Marilla? There must be a limit to the mistakes one person can make, and when I get to the end of them, then I'll be through with them. That's a very comforting thought."

L.M. Montgomery, "Anne of Green Gables", 1908

2

Reliability as a First-Order Design Concern

HARDWARE IS NOT PERFECT, but conventional computing systems typically require near-perfect execution to guarantee correctness. In the shift towards increasingly heterogeneous systems, specialized hardware blocks, and integration of different memory technologies, taking a critical approach to hardware reliability and application resilience can unlock opportunities for increased efficiency. Such a critical approach will require a re-imagining, re-examining, or even retracting of existing

efforts in the deep and complex field of resilience and error correction, as well as a willingness to incorporate the unique fault and failure modes of emerging technology choices in order to extract potential system benefits. In the context of supporting specific classes of workloads, as for hardware accelerators or in computing systems of varying scales, driving resilience studies with a consideration of application-specific metrics, including accuracy, performance, and fault tolerance, is a key advantage of the work presented in this chapter, with a specific emphasis on the co-design of deep neural networks (DNNs) and emerging, embedded non-volatile memory (eNVM) technologies.

Resilience and error correction standards in general-purpose CPUs tend to be brittle and typically require fault probabilities as rare as on the order of 10⁻¹⁵. This level of correctness levies a high design cost at both the device and microarchitecture levels, but has some key advantages. Namely, an operating application can reliably assume perfect accuracy or the hardware, and the user or developer requires no specific awareness or understanding of potential faults or failures. However, relaxing these requirements can enable significant savings, and certain classes of emerging, data-intensive workloads (e.g., deep neural networks) exhibit increased fault tolerance compared to conventional workloads (e.g., one-in-every-thousand values being manipulated or mis-read ¹⁹⁴), providing a plethora of optimization opportunities, particularly in specialized hardware systems.

Ample opportunities to improve performance and efficiency are possible at varying levels of the computing stack if the requirements for absolute correctness can be relaxed. For example, at the architecture level, conventional devices can be pushed to their operational limits by increasing clock frequency to boost performance or reducing V_{dd} to maximize energy efficiency. Additionally, DNNs require substantial amounts of memory to store weights and intermediate values. While modern high-density storage solutions (i.e., flash memories) require expensive redundancy and error correction hardware, such costly safeguards may be overkill for DNNs. At the device level, many emerging technologies promise high efficiency, but have seen little adoption in general-purpose contexts due to their instability and poor yield. At the fabrication level, design rules may be aggressively

pushed to maximize densities at the expense of reliability. These opportunities motivate us to thoroughly understand the extent of resilience in order to safely push the limits of DNN hardware from architectures to devices.

This chapter first presents and describes a DNN-specific, application-level fault injection framework: *Ares* ¹⁹³, including some background on hardware fault and prior work in fault modeling and resilience analysis. Ares studies and simulates faults at the application level while executing DNNs directly on GPUs, which is necessary for conducting large-scale fault studies in a timely matter as part of a larger system design process. Traditional fault injection frameworks based on dynamic instrumentation can experience kernel slowdowns of up to $488 \times ^{86}$, making them prohibitively slow to sweep many fault patterns across a range of fault rates varying orders of magnitude. As such, Ares can provide accurate, first-order design space exploration that can guide lower-level tools on where to conduct detailed microarchitectural fault analysis. Additionally, Section 2.0.1 will review characterization of DNN fault tolerance, which may be broadly applied across a range of devices and architectures that run DNNs. This is vital because DNNs run on CPUs, GPUs, and accelerators, and the optimal microarchitecture for each DNN hardware device is still widely debated ¹⁹².

Next, Section 2.1 presents fault models and reliability-related characterization efforts across different memory technologies. These fault modeling efforts, in turn, underpin the exploration and careful co-design of specific memory solutions paired with different application characteristics and algorithmic optimizations.

The remainder of this chapter (Sections 2.2-2.4) will present several critical examples of the memory system benefits and fundamental device-application trade-offs that can be uncovered by taking reliability as a first-order design concern. Namely, I will discuss multi-level-cell (MLC) encoding strategies for optimized, dense DNN weight storage using CTT devices (Section 2.2 60), the intersection of MLC programming choices for both CTT and RRAM with sparse-encoded data and sparse encoding strategies (Section 2.3 183), and the application of the resulting design methods to

maximize storage density without impacting application accuracy for FeFET-based storage for both natural language processing and graph search workloads (Section 2.4²⁰⁴). I also leveraged the intuitions and methodology developed in this chapter towards a collaboration applying approximate matrix multiplication methods to DNN inference and robotics kernels¹⁸⁶, which is omitted from the following discussion for brevity.

2.0 Application-Aware Resilience Studies

Today, deep neural networks (DNNs) are being deployed at all computing scales—from energy-constrained IoT devices to cost-optimized data centers. Given the complexity and growing scale of DNN deployment, work has mainly focused on optimizing performance and energy efficiency. The initial Ares work ¹⁹³ explored and quantified the inherent resilience of DNNs to hardware-level faults, opening up new directions for design and optimization strategies.

Designing towards a specific application space (DNNs) allows us to re-examine the (possibly overprovisioned) reliability standards and correction strategies. Alternatively, emerging memory technologies have unique failure modes and fault models by virtue of their unique underlying physics. The application-level impacts of various storage technologies, schemes, and systems must be studied fresh and remain front-of-mind during system design and optimization processes for maximum efficiency.

This section will first discuss a bit of context about hardware faults, and in understanding the potential resilience of DNNs. Next, we present some preliminary results from Ares ¹⁹³ to build intuition about DNN fault tolerance and the impacts of application-level-resilience-aware design. Additionally, this section will describe ways Ares has been extended and modified to support fault injection during training and under varying number representations and data formats ¹⁹¹. Studying DNN resilience under more fixed definitions of hardware errors (e.g., bit-flips and power failures)

will prepare us to intersect these intuitions with additional application, system, and device-level optimizations including sparse encodings and varying eNVM solutions.

2.0.0 HARDWARE FAULTS

Hardware designers have been forced to consider hardware reliability since the inception of the field. There are many sources of faults in modern semiconductor chips and storage media; manufacturing process variation, voltage noise, and temperature contribute to unpredictable circuit delays that force designers to use worst-case margins. Attempts to minimize these wasteful margins risk the occurrence of timing violation faults under certain operating conditions. SRAM circuits are exposed to large delay variation because they use smaller devices than logic cells, and there are a vast number of cells on each chip, which increases the chance of an outlier. Similarly, heavily-scaled DRAM and flash memory cells now store so few electrons to encode each bit that they are occasionally flipped by common noise events. Cosmic particle strikes are a concern for some applications, as in datacenters, because they result in single-event upsets. Datapaths can also fault when operating conditions introduce excess delay. Another source of fault for both memory and datapaths can be self-induced via approximate circuits. The Ares work focused on static faults in memory, which are pertinent to DNNs due to the large storage requirement for weights and intermediate states ¹⁹³.

Faults can be further classified into transient and static varieties. Transient faults come and go over time and are caused by abnormal conditions or events (e.g., resonant supply voltage noise and particle strikes). Static faults persist in the affected device and occur in cases such as "weak" SRAM bit cells due to process variation or flash life-time wear problems. Because static fault persist in time, we see them as a superset of transient faults, and studying their effects first provides a lower bound on DNN fault tolerance. While there are well-known methodologies for identifying and quantifying reliability per-component at an architectural level ^{167,234} or for application-level studies of architectural resilience ²³³, Ares aims to perform fault injection and quantify fault tolerance at an

application-level, using an understanding of fundamental technology fault models.

2.0.1 ALGORITHMIC RESILIENCE OF DNNs

Previous work suggests DNNs can be robust to faults ¹⁹². For example, eliminating individual nodes or parameters leads to a graceful degradation in model accuracy, as also described in Chapter 1.1.1. Furthermore, regularization and simplification techniques can also be employed to develop, bolster, or otherwise reflect DNN robustness, as DNN models demonstrate correct operation with perturbed parameters. However, these cases are not representative of how faults typically manifest in hardware. Thus, the Ares work intersected some of the most popular DNN optimizations and simplification techniques with bit-flip fault models and error rates to develop a fuller picture of the fault tolerance of DNNs.

The hardware community has also shown interest in understanding DNN fault tolerance. For automotive applications, transient faults can lead to problematic image misclassifications; in order to meet ISO standards, techniques to improve reliability are required ¹³⁷. Finally, Minerva ¹⁹⁴ proposes fault mitigation techniques to reduce SRAM supply voltage in order to save energy while preserving inference accuracy in fully connected DNNs.

2.0.2 Ares: A Tool for Quantifying the Resilience of DNNs

We developed Ares to be a fast, scalable fault injection framework that enables rapid fault analysis demonstrated with fully connected (FC), convolutional (CNN), and recurrent architectures such as gated recurrent unit (GRU) based DNNs. We found that model-specific, more-aggressive data types can provide 10× more resilience, and fault tolerance can vary across models by several orders of magnitude. The basic methodology flow of Ares is depicted in Figure 2.0. Treating training as a large, one-time cost that has been performed in a fault-free environment, we focus on the execu-

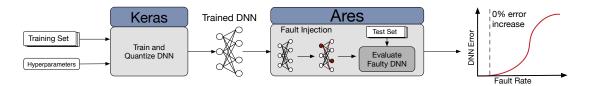


Figure 2.0: Given a trained and quantized DNN model, Ares can simulate the impact of transient and static fault models on DNN execution, resulting in concrete application-level impacts (e.g., image classification accuracy) under varying fault rates, adapted from ¹⁹³.

tion of inference. We can characterize the accuracy of trained models by performing inference on a particular set of previously unseen inputs.

The studies presented as part of the demonstration of Ares also revealed optimization opportunities by considering faults at a per-layer granularity. For example, within a DNN, the per-layer fault tolerance can vary by up to $2781\times$. We also found that increases in fault rate incur graceful degradation in accuracy for FC and GRU layers, and less so for CNN layers. Additionally, the different types of data of DNNs (i.e., weights, activations, and hidden state) exhibit different fault tolerance behaviors; the activations of a model can be up to $50\times$ more resilient than the weights.

2.0.3 FAULT INJECTION MECHANISMS

Ares establishes baseline classification error by executing inference on a pre-trained DNN with floating point data types for all structures (defined as Iso-Training Noise, as in Chapter 1.1.1). Next, structures are quantized to desired fixed point representation (no accuracy loss is permitted).

Ares has two modes of fault injection: static (for trained weights) and dynamic (for activations). Static faults are injected off-line, before inference is executed. Injecting faults statically is preferred as it introduces no performance overhead during execution. Dynamic faults are injected during the execution of a DNN inference. In Ares, the overheads of dynamic fault injection are minimized by using native tensor operations to emulate fault behavior. So long as the fault model under study can be cast as an element-wise operation or a linear transformation of the state (e.g., random or

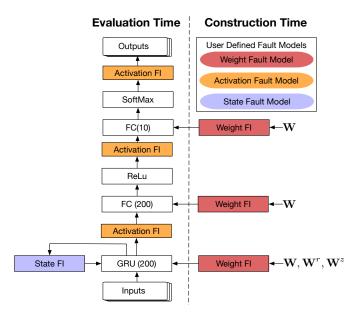


Figure 2.1: Ares is an application-level fault injection framework for simulating both static and dynamic fault models, incorporated in program execution either during evaluation or off-line during model construction, taken from ¹⁹³.

systematic noise), it can be implemented as a tensor operation and run on a GPU. For GRU layers, we inject faults into hidden states at each time step between state updates, as highlighted in Figure 2.1. The maximum slow down from dynamic fault injection is less than $3.5 \times$.

Ares performs fault injection at designated points across the weights, activations, and hidden states, depending on the DNN topology and planned experiment. Each fault injection experiment requires the bit error rate (BER) and faulty structures to be specified. Fault patterns are generated by sampling a uniformly random distributed process, which identifies which bits will fault per structure. Ares models faults by using the bit-level fault pattern to mutate stored values. By sweeping fault rates, the user can then analyze the fault tolerance of the DNN.

To further illustrate how Ares performs fault injection, Figure 2.1 lays out the network topology and fault injection points for TiGRU: a three layer DNN with one GRU and two FC layers. Ares is built on top of Keras¹, with an alternate PyTorch^{181,191} implementation with similar scope and

functionality.

Fault injection is performed at two stages: construction time (static injection) and evaluation time (dynamic injection). Once the DNN is trained, the weights are known, and Ares injects weight faults at construction time by manipulating saved weight values. In contrast, injecting faults into activations requires changes to the program execution. Activation and state fault injection operators are implemented as GPU-compatible element-wise tensor operations. Datapath faults can also occur (e.g., in MAC units), which Ares models by manipulating the data before and after compute operations. The presented results focus on memory faults at all memory fault points, as a precursor for modified and optimized memory system solutions in the proceeding sections. To demonstrate that Ares can accurately capture bit error behavior exhibited by real hardware, we validated simulation results against measurements using a DNN accelerator capable of inducing and measuring SRAM faults ^{242,194}, as detailed in ¹⁹³.

2.0.4 Models

Six diverse DNNs are used to evaluate Ares. The datasets used to train these DNNs represent major application domains in which DNNs are commonly used: image classification (*CiFar-10*, *MNIST*, *ImageNet*) and speech classification (*TIDIGITS*)^{125,197}. Datasets have varying sizes and models vary greatly in depth and composition. For larger DNNs, we group clusters of similar, adjacent layers into *layer blocks* (LBs) to increase interpretability and ease organization.

2.0.5 Summary of DNN Inference Resilience to Hardware Faults

We use Ares to quantify the relationship between faults and accuracy in DNNs. Results reveal over an order of magnitude variance of fault tolerance across models, between layers, and across structures. Hence, Ares enables exploration of optimizations using different aspects of DNN resilience.

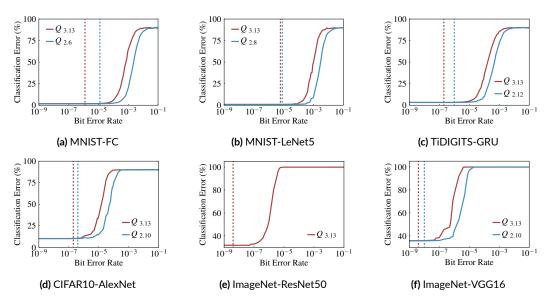


Figure 2.2: The tolerable bit-error-rate (BER) before classification accuracy degrades (marked by vertical, dashed lines per model) varies per DNN and per quantization choice, taken from ¹⁹³.

We simulate the impact of bit errors at a model granularity by injecting faults across all weights in each model. Figure 2.2 shows the resulting BER-accuracy curves, and the dashed vertical lines indicate the highest tolerable BER without loss of inference accuracy. The two vertical lines show two quantized types for each model: one using only the minimum number of bits required to run inferences without loss of accuracy (blue) and the other (red) a global type that is sufficiently accurate across all DNNs considered.

All models show heavily thresholded accuracy degradation: small BERs have a negligible impact on accuracy up to a certain threshold point. At BERs beyond this point, model accuracy degrades exponentially. The point of o% accuracy loss is consistently orders of magnitude lower than the knee of this curve. Between these points, the curves typically exhibit a gradual decline in accuracy. Depending on the application, this could be an interesting operating regime in which a loss in accuracy may be exchanged for efficiency gains.

Resilience varies across DNNs. There is a large spread in fault sensitivity between DNNs; de-

pending on the model and data type, the knee of the BER-accuracy curve can vary by multiple orders of magnitude. Considering the unified data type for all models, we see that the o% accuracy loss can vary from 4×10^{-9} (VGG16) to 6×10^{-6} (LeNetCNN). These results suggest protection mechanisms and optimizations used (e.g., ECC and optimal voltage reduction) should be engineered on a per-model basis.

Quantization impacts resilience. To improve efficiency, DNNs are quantized to minimize the total number of bits. We study weight fault tolerance with two data types for each network: a unified $Q_{3.13}$ (i.e., 3 integer and 13 fractional bits) and a per-model optimized type. $Q_{3.13}$ is the minimal type needed for no loss of accuracy across all models, with ResNet50 requiring the most bits. While a single global type would be needed to build flexible hardware that supports all 6 DNNs, model-specific types are the most aggressive quantization per DNN without increasing baseline error.

Figure 2.2 shows that $Q_{3.13}$ (red) consistently results in lower fault tolerance than model-specific types (blue). For LeNetFC, the optimized $Q_{2.6}$ data type is $10 \times$ more fault tolerant. This is because the number of integer bits used (3 versus 2) determines the range of representable values, and models other than ResNet50 can be clipped to just 2 integer bits with no loss in accuracy. The unnecessarily large range of possible values allows for faults of greater magnitude to occur, and hence increases the potential impact of a fault (e.g., flipping the MSB of a near-zero valued weight results in a greater change in value if 3 integer bits are used rather than 2). Flipped superfluous fractional bits do not have as much of an impact on parameter value. By reducing the number of integer bits from 3 to 2, model-specific quantizations consistently demonstrate increased fault tolerance.

Some image classes are consistently mispredicted. We found the variance of the distribution of test-case mispredictions (i.e., the number of mispredictions for each class at a sampled fault pattern) is positively correlated with classification error. While fault patterns with high per-class misprediction variance are outliers, it can skew the average classification error. In LeNetCNN, 100 samples at a fault rate of 1×10^{-3} resulted in an average error of 12.2% while the median was only 6.8%.

There are two ways a network can exhibit high class-misprediction variance: one class can constantly be misclassified or half the classes tend to be wrong and half correct. We found the latter case had a greater effect on model classification error. It is worth noting we also found specific fault patterns where a single class was consistently mispredicted.

2.0.6 Sensitivity to Data Formats

How weights are represented in memory changes the impact of certain bit flips, potentially affecting the model's vulnerability. In a standard two's complement representation, flipping the sign bit is equivalent to subtracting the largest expressible value for positive weights. DNN weights are clustered near zero, with fewer than one-in-a-million weights in CiFar10-VGG12 larger than 1.5, so sign bit flips are likely to have a large impact on the magnitude of the weight. To determine the extent of this impact, Ares is used to simulate oracle protection to the sign bit of weights, where flips that would occur in the sign bit are silently suppressed. Fault injections are performed on the weights of the six baseline models used in Ares (MNIST FC, MNIST LeNet5, ImageNet VGG16, ImageNet ResNet50, CiFar10 VGG12 and TIDIGITS GRU), with per-model customized quantization at iso-accuracy, as detailed in ¹⁹³.

For each model, an instance is trained, then trials are performed across 50 bit error rates (BER) in the range [1e-9, 1e-3]. For each trial, the test set error after fault injection is recorded. At each BER, 20 trials are averaged to estimate the mean model degradation at that fault rate. Taking the model test error with no fault injections as the baseline, the maximum bit error rate with less than a relative .5% increase in error is calculated and displayed for different NNs per data format in Figure 2.3. Compared to standard two's complement, oracle protection of only the sign bit gives up to an order of magnitude improvement in fault tolerance, as shown in Figure 2.3.

Unfortunately, completely protecting sign bit errors may require architectural tradeoffs, such as storing the bits on a different memory technology or with error correction, which introduces

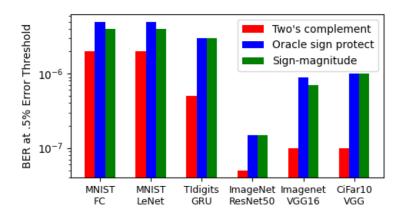


Figure 2.3: Sign-magnitude encoding results in models tolerating up to $10 \times$ higher bit error rates (BERs) than two's complement, nearly matching the effect of oracle protection of the sign bit, taken from ¹⁹¹.

overhead. This study achieved the benefits of suppressing sign errors without special protection by changing the bit representation of the weights to sign-magnitude (SM). In sign-magnitude encoding, flipping the sign bit transforms x to -x while leaving the absolute value unchanged. If M is the largest representable value, for x such that $|x| \leq M/2$, a sign bit error changes x by less in sign-magnitude representation than in two's complement. For values close to 0, the effect on weight magnitude of a sign bit error is much larger in two's complement than sign-magnitude. For CiFar-VGG12, which has an approximately normal distribution of weights with mean zero and standard deviation less than .02, for over 99.9% of weights a sign bit flip in 2c will have more than $10 \times x$ greater impact on weight magnitude than the equivalent flip in SM. We use Ares to analyze the same set of model instances using sign-magnitude representation. For CiFar10-VGG, ImageNet ResNet50 and TiDigits GRU, sign-magnitude provides equal resilience as oracle protection in terms of tolerable bit error rate (Figure 2.3), achieving all of the benefit without special protection of sign bits. For the MNIST FC and LeNet and ImageNet VGG16 DNNs, the maximum sign-magnitude BER is within 20% of that achieved with oracle protection. Thus, SM representation could allow sign bits to be stored on the same memory as other data, while obtaining the benefit of "protecting" them

from error. Despite the design and implementation costs, moving to a sign-magnitude encoding for weights could result in 10x higher acceptable bit error rate while using the same number of bits per weight value.

2.0.7 DNN Resilience Across Training Runs

Implicit in experiments that examine the fault tolerance of a single instance of a model ^{137,193} is the assumption that identically trained instances have equivalent fault tolerance. However, this assumption does not necessarily hold. Even after isolating a specific dataset, model architecture, and training hyperparameters, we find a wide distribution of fault tolerance. Furthermore, changes to bit representation can affect the variability of resilience to faults (i.e., how predictable the fault tolerance and resulting accuracy is), in addition to the absolute level of fault tolerance. As with variance of hardware characteristics between devices, variance in the fault tolerance of models on a fixed device may result in a requirement to design for the worst case. Thus, studying the variability in fault tolerance has implications for provisioning hardware if a specific model architecture is continually re-trained and re-deployed ⁸⁷.

This follow-on study to Ares¹⁹¹ quantified the variation of fault tolerance across training runs of a VGG12 model on the CiFar10 dataset, as summarized in Figure 2.4. Each model instance is trained to convergence with identical hyperparameters (e.g., learning rate = 0.1, L2 = 5*e*4) and results in classification accuracy well within previously measured bounds ^{192,193}. For each potential fault rate, as in the previous study (Section 2.0.5), uniformly random bit flips are injected across all weight values, with 50 trials per BER per model instance.

A key finding, made clear in Figure 2.4 is that there is nontrivial variation in fault tolerance across identically-trained models, which should inform future quantification and analysis of DNN reliability and fault tolerance. In fact, if we consider the BERs at which different model instances experience 0.5% error degradation, we observed up to 4× discrepancy for two's complement number

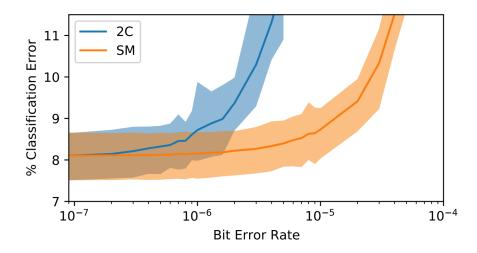


Figure 2.4: Mean and min/max (indicated by shaded region) classification error across 20 instances of VGG12 trained on CiFar10 for varying BERs and different data encodings (two's complement, 2c, and sign-magnitude, SM). SM exhibits slower degradation and lower variance, visible as a smoother curve, taken from ¹⁹¹.

representation. In more concrete terms, a design target that identified a one-every-million-bits error rate permissible, if applied to another instance of an identically trained model and re-deployed, could actually experience accuracy degradation below acceptable threshold $4\times$ as often as predicted. For sign-magnitude (SM) encoding, BER at the 0.5% threshold varies by up to 2.6 \times . In fact, for a CiFar-VGG instance retrained with identical hyperparameters, assuming equal fault tolerance results in over 5% increase in missclassified images across the test set. Classification error for each model instance initially degrades smoothly below a BER of approximately 2 \times 10⁶ for SM and 2 \times 10⁻⁷ for 2c, then rapidly decay (Figure 2.4).

These results suggest that the acceptable BER to minimize impact on classification accuracy differs even among identically trained instances. A given model can be associated with a specific BER above which the model accuracy is unacceptable across all instances, but architects working at lower error tolerances must consider per-instance variation. For situations where models must adapt on device, or where multiple training runs sweeping fault tolerance are cost-prohibitive, sign-magnitude

encoding allows tighter control on the worst case error at a given bit error rate, in addition to lower mean error. More broadly, these results reveal the sensitivity of co-design strategies to the naturally stochastic behavior of DNN training and the important role of bringing re-training, fine-tuning, and other algorithmic choices in-the-loop with reliability analysis and system design choices, as investigated in other studies presented in this dissertation (e.g., Chapters 3.2, 3.3).

2.1 FAULT MODELING OF MULTI-LEVEL-CELL (MLC) ENVMS

Next, we discuss the underlying physical behaviors and circuit design considerations that inform the development of fault models for multi-level-cell programming of several compelling eNVM technologies. These fault models, in turn, inform system-level and application-level evaluations of eNVM memory solutions across several projects presented in this dissertation.

One key challenge in MLC storage is that an increase in number of levels is more susceptible to device-to-device variations, making the memory less reliable. In addition, the ADC cost in terms of area and power can limit in some cases the effectiveness of MLC implementations. In order to discriminate among different programmed levels, the memory uses analog to digital converters (ADCs). The ADCs translate each programmed I_D target level to the corresponding binary word. Individual circuit and device design choices have direct impacts on the resulting reliability of MLC-programmed memory cells and memory array architectures.

2.1.0 Multi-Level CTT Characterization

As discussed in Chapter 1.2.4, the benefits of CTTs come at the expense of two undesirable properties, limiting their deployment and rendering them ineffective for many general-purpose compute tasks: (1) extremely long write-latencies and (2) potential for high cell fault rates. Despite these limitations, several works, including those presented in this thesis, have demonstrated CTT-based mem-

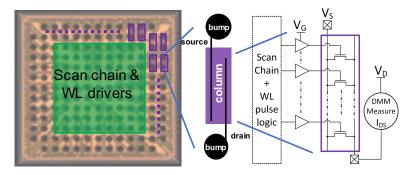


Figure 2.5: Test chip die photo and column schematic for Charge-Trap-Transistors (CTT), adapted from ^{183,157}.

ories as a near-ideal solution for enabling efficient, embedded DNN inference with careful architectural co-design. This is possible because deeply-embedded DNN weights are updated infrequently and repeatedly used to make inferences, and DNNs are known to be fault tolerant.

Several projects presented in this thesis rely on direct measurements of fabricated MLC-CTT test structures in a commercial 16nm FinFET process ^{183,157}. A die photo of the chip is shown in Figure 2.5, which contains 36 columns of 128 cells each; internal scan chain and driver circuits mimic wordline drivers. The bumps expose column bitlines to flexibly read and write individual cells via external test equipment. Figure 2.6 (left) shows the distribution of read currents at different wordline supply voltages for 8-level programmed CTTs (a 3-bit MLC). The different colors represents unique levels (i.e., programmed values), and each level is measured from 128 unique devices.

Figure 2.6 (right) shows a slice of the current distributions at the nominal wordline voltage of 0.8V. The yellow cluster corresponds to unprogrammed cells with intrinsic V_{th} variations inherent to the process technology. Other program levels exhibit tighter distributions due to the iterative write-and-check process. In both cases, the histograms are well approximated using a Gaussian distribution. As more levels are encoded per cell, the current distributions tend to overlap, which increases the probability of misreading the programmed value. This plot highlights two co-design opportunities: first, as more levels are encoded, the current bands tend to overlap, increasing the

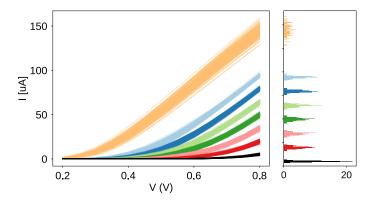


Figure 2.6: For programmed MLC-CTT test devices, we measure the distribution in current corresponding to each programmed level across voltage (left) and at fixed 0.8V (right), adapted from ^{183,157}.

probability of read errors. Given the wider distribution produced by intrinsic process variations, we purposely add separation between the unprogrammed and first programmed state to minimize read error probability of the unprogrammed state.

2.1.1 GENERALIZABLE MULTI-LEVEL-CELL FAULT MODELS

In developing a fault model for eNVM implementations, we focus on two primary sources of uncertainty: (1) the intrinsic randomness associated with the process of setting the value of different programmed levels in a memory cell, and (2) the effect of mismatches in the sensing circuitry.

A fault in an MLC is defined as a device being incorrectly read as a level adjacent to the intended one. The probability of a fault occurring for a given level, L_n , is determined by two reference thresholds: $I_{\text{ref n}}$ and $I_{\text{ref n-1}}$; reference thresholds discretize current ranges of the cell. Figure 2.7 shows the current level distributions for a generic MLC. ΔI_{init} represents the distance between the mean value of the initial state, L_0 , and the mean value of the first programmed state, L_1 . The remaining programmed levels are equally spaced by ΔI_{prog} . These deltas are tunable and determine the fault rate between levels, which introduces co-design opportunities.

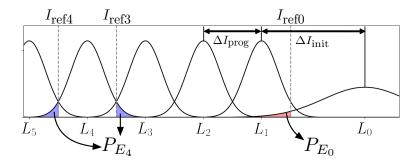


Figure 2.7: For a generic multi-level-programmed cell (MLC), each level can be represented as a gaussian, and the probability of mis-reading a programmed value corresponds to the overlap or intersection of adjacent programming distributions, labeled as P_E , adapted from 60 .

Figure 2.7 highlights the probabilities of two possible cell faults: P_{E_0} and P_{E_4} . We first consider the case of a fault in a cell set to the initial state, L_0 , to be incorrectly read as a cell in the first programmed state, L_1 . The probability of this fault (P_{E_0}) occurring is given by the total probability of the initial state's distribution beyond the reference current I_{refo} .

$$P_{E_0} = P(I_{\text{cell}} < I_{\text{refo}}) = \int_{-\infty}^{I_{\text{refo}}} P_{L_0}(x) dx$$
 (2.1)

In the case of a cell programmed to a specific level (i.e., not in the initial state), the cell could fail by being read as either of the two adjacent levels. In Figure 2.7, this is illustrated as P_{E_4} . Thus, the probability of a fault is given by the sum of the probabilities of having L_4 fall in either of the erroneous ranges:

$$P_{E_4} = [1 - P(I_{\text{cell}} < I_{\text{ref}_3})] + P(I_{\text{cell}} < I_{\text{ref}_4})$$
 (2.2)

Given $\Delta I_{\rm init}$ and the number of levels per cell, a detailed fault model for CTT-MLC is constructed. $\Delta I_{\rm prog}$ results from partitioning the remaining ΔI after $\Delta I_{\rm init}$ is allocated. This allows us to make P_{E_0} arbitrarily small in order to protect the initial state at the cost of increased fault rates in the programmed levels. The effects of current to voltage conversion on the distribution must also

be taken into account.

The intrinsic distributions for the stored levels in a RRAM implementations are extracted from published data for 3-bit MLC programming ²⁶⁸. For MLC-CTT, inter-level fault rates are determined by directly measuring the current distributions from our test chip, as shown in Figure 2.6. For both technologies, current distributions can be modeled as gaussian distributions, and the overlap of level distributions determines the rate at which that value will be misread as an adjacent programmed level. In both cases, we use SPICE simulations to derive the distributions at the output of the current-to-voltage converter. When packing many levels into a single cell, the likelihood of misreading a programmed level can be high in both CTT and RRAM (e.g., fault rates for MLC-3 range from 10⁻³ to 10⁻⁵). However, errors typically result in reads to an adjacent level*. By carefully arranging how data is encoded into MLCs and exploring the impact of varying number of bits per cell on DNN classification error, the effects of faults can be mitigated (Section 2.3). For each cell design, we complement measured or extracted characteristics with SPICE simulations to determine equivalent current or resistance and read voltage descriptions for each programmed level ^{60,183,59}.

The design and optimization of sensing circuitry also determines the fault rate of a multi-level-programmed memory cell for the eNVMs discussed. The sensing circuitry must perform the analog-to-digital conversion from the measured cell current to the binary-valued memory array output, as summarized in Figure 2.8. More precisely, an n-bit MLC outputs a current level of I_i in the read mode when programmed to level i. The sense amplifier (SA) circuit detects the level (one of 2^n) of the cell from I_i and outputs an n-bit signal. The designed SA circuit amplifies I_i , compares it to some quantized level, and concludes the saved level delivering an n-bit output. The analog-to-digital converter (ADC) employed by the SA circuit is similar to a flash-type ADC and has $2^n - 1$ quantized levels for an n-bit MLC. The SA circuit, in turn, suffers from D2D variation and affects the reliability of the memory.

^{*}Misread probability of non-adjacent level is 1.5×10^{-10} or below.

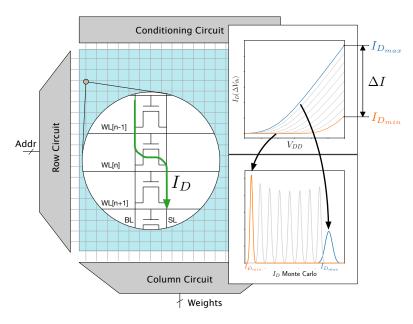


Figure 2.8: At a memory array level, the programmed level per MLC device correlates to a current value in a fixed, programmable range, and is directed to sensing circuitry for read-out, adapted from ⁶⁰.

Reading back the stored value requires converting the programmed analog level to a binary word, and can be done using parallel sensing or sequential sensing schemes. Parallel sensing is similar to using a flash ADC, and requires each bitline to have dedicated sense amplifiers for each possible stored level. Sequential sensing uses a single sense amplifier and recovers the stored binary word iteratively for each bit. While sequential sensing reduces the overall number of sense amplifiers, we noticed that implementing parallel sensing with small sense amplifiers does not incur an excessive area penalty.

Several of the works presented in this thesis focus on a specific sense amplifier (SA) design, which is characterized by having low static power and input referred offset primarily determined by the input differential pair of transistors (Figure 2.9, left). We characterize the input offset voltage for a latch-based sense amplifier 33 through Monte Carlo SPICE simulations. For these simulations, we swept the width of the input transistors as they are primarily responsible for setting the input offset voltage. These results allow us to quantify the resulting inter-level fault rate per device. For example,

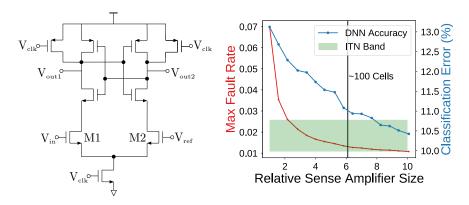


Figure 2.9: Example schematic for a sensing circuit for multi-level programming (left), highlighting the relative size of M1/M2 as a determinant of fault rate, and simulated fault rate results for 3-bit MLC programming including ADC (right), plus application-level impact on ResNet/CiFar10 image classification error, adapted from ¹⁸³.

in Figure 2.9, right, varying the relative sense amplifier size has a direct impact on both the raw fault rate and the resulting impact on classification accuracy for a workload of interest (CiFAR-10 image classification using a ResNet model).

Based on these simulations, we choose a SA size such that the overhead incurred due to SA choice for the overall array in our final results never exceeds 1%, and the inherent inter-level MLC fault rates are altered by less than $2\times$. For the readout architecture, we consider a parallel sensing scheme, whose operation is similar to a flash ADC: the bitline is connected to a number of SAs equal to N-1, where N is the number of levels that can be stored per cell, with each SA connected to the appropriate reference voltage. A parallel sensing scheme can decode the stored value in a single conversion step. However, the number of required SAs increases exponentially with the number of stored bits. This overhead is mitigated by multiplexing the memory array columns. The resulting design is integrated with NVSim 62 to characterize the overall memory architecture.

2.1.2 Ares Framework Extensions

Several projects presented in this thesis extended Ares to model MLC eNVM faults, accommodate sparse-encoded values, and simulate error mitigation strategies. Based on the results from Section 2.1.1, we generated a inter-level fault probability map for each MLC configuration (either CTT or RRAM). The baseline fault probability map is modified to include the effects of the sensing circuitry by shifting the mean of each level distribution by an amount sampled according to the characterized input referred offset.

Fault injection is performed by first converting the weight values into an MLC representation. Then, for each eNVM cell, we sample a gaussian random variable from the appropriate level distribution and check if the result crosses the thresholds which are used to sense the cell's content. If they do, a fault has occurred and the faulty memory cell's value is updated to the value represented by the adjacent level in Section 2.3. After determining fault locations, the modified values are used to perform inference across the entire test dataset. Experiments are repeated over many trials, and presented results are averaged over many unique generated fault maps.

Sparse encodings require separate fault injections on each structure (e.g., the bitmask and non-zero weight values), and we vary the number of bits per cell used to store each structure. When implementing support for ECC, we must ensure a level-to-level fault correlates to a correctable (single bit flip) error by using Gray coding to store binarized values in the MLCs. Dynamic error correction and mitigation strategies are integrated such that faults are detected and values are updated as appropriate prior to evaluating model accuracy. Ares enables the evaluation of DNN classification error over many randomly seeded trials for various MLC eNVM configurations and encoding strategies. From this analysis, we definitively determine the optimal number of bits stored per MLC and the minimal number of memory cells required for each encoding strategy for each DNN such that there is no loss in accuracy.

2.2 Customized, Iso-Accuracy MLC storage of DNN weights

This study describes how CTT-MLCs can be leveraged to eliminate off-chip neural network (NN) weight accesses ⁶⁰. To address the faults incurred from storing multiple bits per device, these experiments leverage the fault model and fault infrastructure presented in Section 2.1 to vary how many cells and levels per cell to use for fixed-point weight representations. To optimize this implementation, we co-design DNN weights and CTT-MLC device properties by: (i) clustering the weight values to require fewer memory cells, (ii) using non-sequential level encodings to mitigate the effects of faults, and (iii) pruning the network to skew weight values and leverage the non-uniformity of fault probability in CTT-MLCs. While this sections focuses on CTT-based eNVMs, this approach is deepened and applied to other MLC eNVM technologies in Sections 2.2 and 2.4 and again in Chapter 3.

2.2.0 DATA FORMATS

This study considers two types of weight quantization: fixed-point and clustering. **Fixed-point** data type quantization is a well-known and effective hardware optimization technique; reducing the width of data types can substantially reduce area and power dissipation while improving performance. In NNs, fixed-point quantization can be aggressively applied to weights. The 32-bit floating point types used for training weights can be reduced to use only 8-12 bits for inference without compromising accuracy. The second weight quantization approach we consider is k-means clustering ¹⁹², in which NN weights are mapped onto a set of k values on a per-layer basis. Clustering is advantageous as only the indexes need to be stored per weight and $\lceil \log k \rceil$ is typically significantly less than the number of bits required with fixed-point quantization. The overhead for storing the look-up table of the actual k weight values is negligible (e.g., 4-16 8bit values per NN layer).

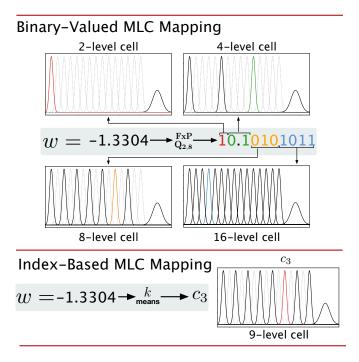


Figure 2.10: In assigning DNN weight values to MLC programmed levels, data format and programming choices will determine reliability of storage; mis-read MSBs may have outsize impact on application accuracy and may be selectively stored reliability in 2-level cells, or programming settings can be customized for clustered DNN values corresponding to non-power-of-two number of programmed levels, adapted from ⁶⁰.

2.2.1 Data Encodings to MLC Storage

As a example, consider the conversion of the value $(-1.3304)_{10}$ and its fixed-point representation $(10.10101011)_2$ using two integer bits and eight fractional bits. If this value is stored using binary encoding with CTT-SLC, then 10 cells are needed (one per bit). The same value can be stored using only 3 CTT-MLCs if the most dense, 16 level (4 bits), cell configuration is used. This uniform configuration reduces the area cost per bit at the expense of a higher fault rate. Since faults on higher order bits have a disproportionate impact on the stored value, we anticipate that it is advantageous to selectively mitigate these faults.

Figure 2.10 shows an alternative encoding to MLCs better tuned to both memory cell properties and DNN fault tolerance: a non-uniform encoding where 4 cells are used with 2, 4, 8, and 16 levels

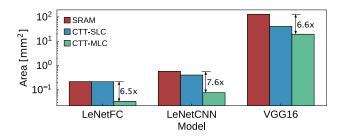


Figure 2.11: Area reductions for fixed-point NN weights, with no loss in DNN accuracy, for MLC-CTT storage vs. on-chip SRAM, adapted from ⁶⁰.

per cell, from most-significant-bit (MSB) to least-significant-bit (LSB). The example shows that the sign and integer portion of the weight can be protected by using a CTT-MLC with fewer levels. Compared to uniform encoding, this non-uniform encoding requires only one additional cell and substantially reduces the fault rates in weight MSBs.

Storing cluster indexes (Figure 2.8, bottom) enables two optimization opportunities. The first technique is an intra-cell optimization to protect the initial state from faults by increasing $\Delta I_{\rm init}$. This technique is particularly effective when the majority of the parameters are assigned to a single cluster, a property we actively enforce with weight pruning. We further consider distinct ways to map clusters to levels to mitigate the magnitude and fault rates, varying the relative level-to-level fault rates of the most common cluster indexes vs. ordering by corresponding weight value. The second technique is an inter-cell, multi-cell per weight optimization. If more than one cell is required to store the cluster index, then some clusters can be more protected via the asymmetrical fault rates of MLC and non-uniform encoding, (a non-uniform allocation of cluster index bits across cells).

2.2.2 REDUCING WEIGHT STORAGE FOOTPRINT (EVALUATION)

In Figure 2.11, the area benefits of using CTT-MLC eNVMs with a fixed point encoding are measured against two baselines: SRAM and CTT-SLC (storing one bit per CTT cell). We evaluate

Model	Weights (#)	Error (%)	Q (int.frac)	Encoding	# Clusters (Max)	Config (lvls/cell)	Area (mm ²)
				FxP	ē	4,8,8	0.033
LeNetFC	270K	1.91%	2.6	C	16	16	0.008
				P+C	I 2	12	0.008
				FxP	÷	4,4,8,8	0.076
LeNetCNN	600K	0.85%	2.8	C	8	8	0.021
				P+C	II	11	0.021
				FxP	*	4,4,4,4,16,16	40.53
VGG16	135M	37.8%	2.10	C	64(CNN) / 8(FC)	2,2,4,4 (CNN) / 8 (FC)	5.7
				P+C	64 (CNN) / 9 (FC)	4,4,4 (CNN) / 9 (FC)	4.9

Table 2.0: For each model, we report the quantization (Q) in the form integer.fractional bits, the number of clusters, and the configurations for each encoding scheme: fixed-point (FxP), clustered (C), and pruned & clustered (P+C). For VGG16, the two values of number of clusters, config, and level map given represent the values for CNN layers vs. FC layers, respectively, adapted from ⁶⁰.

the effectiveness of CTT devices to store the weights of three prototypical NNs listed in Table 2.0: LeNetFC is a three-layer fully-connected (FC) network, LeNetCNN is a five-layer convolutional neural network (CNN), and VGG16 is a much larger CNN that also contains large, prunable FC layers ²¹³. LeNetFC and LeNetCNN use the well-known MNIST dataset for handwritten digit classification, and VGG16 uses the popular ImageNet dataset of colored images to be classified into 1000 possible classes ¹⁹⁷.

The fault injection simulations are implemented using Ares ¹⁹³. For both fixed-point and cluster representations, a corresponding encoding transform function is defined. Starting from a trained model, the encoding transform is applied to the original parameters on a per-layer basis. Next, the faulty cells are randomly chosen using the error probability based on the number of levels per cell and the specific level value. The transformed parameters are used to evaluate the accuracy of the model for a specific encoding configuration. To find the configuration that minimizes the area footprint while maintaining model accuracy, all possible configurations of levels per cell and number of cells are tested for each NN and each variation of data format.

A comprehensive exploration of data encoding strategies and NN accuracy impacts identified that fixed-point CTT-MLC encoding requires 3 transistors per parameter for LeNetFC and 6 for VGG16, as listed in Table 2.0. Compared to an SRAM baseline, this results in a total area reduc-

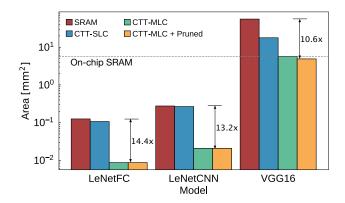


Figure 2.12: Area reductions for clustered NN weight values, with no loss in DNN accuracy, for MLC-CTT storage vs. on-chip SRAM, adapted from ⁶⁰.

tion of up to $7.6\times$, as shown in Figure 2.11. Clustering is often preferable to fixed-point quantization because only cluster index pointers are stored. All three NN models require between 8 and 64 clusters to preserve model accuracy, which requires just 3 to 6 bits to represent each parameter, as summarized in Table 2.0.

When encoding clustered parameters in CTT-MLC, the benefits are immediate. For LeNetFC and LeNetCNN, each parameter can be stored in a single transistor because only 16 and 8 clusters are needed to preserve accuracy for each model. Compared to fixed-point quantization, k-means clustering saves 3.8× and 3.6× area when storing LeNetFC and LeNetCNN in CTT-MLCs. Compared to SRAM, storing the clusters in CTT-MLCs saves 14.4× and 13.2× for LeNetFC and LeNetCNN, respectively, as shown in Figure 2.12.

2.2.3 Customizing with Per-Layer, Pruned DNN Weights

One additional step of customization that unlocks significant density benefits begins with the observation that in the VGG16 network, 89.5% of the parameters are in the FC layers. These FC layers need only 8 or 9 clustered values, while the CNN layers need up to 64 to maintain model accuracy

(Table 2.0). Using the same number of cells for parameters in the FC and CNN layers is wasteful as, once again, we need a single transistor per parameter to encode the parameters in the FC layers. For VGG16, 1 cell is used for each FC parameter and 4 cells are used for each CNN parameter, and this use of heterogeneous configurations saves us 3× area. For VGG16, CTT-MLCs provide 9.8× total area savings compared to SRAM.

This study also leveraged the particular MLC properties of CTT devices, where the initial state (see Figure 2.7) provides a unique opportunity for optimization: because it can be deliberately separated from the programmed states, the fault probability can be skewed to protect the most frequent cluster. To further leverage this protected state, the potentially high sparsity of stored weight values allows us to selectively map zero-valued weights to the less-fault-prone cell level. In VGG16, we prune 97% of all parameters in the first FC layer, which has over 100M parameters. This effectively protects a much larger proportion of the parameters compared to the unpruned network. In the CNN layers, pruning reduces the number of cells needed to store parameters from 4 to 3, leading to a further area reduction of 1.17× over CTT-MLC with clustering. Resulting sparsity and corresponding CTT-MLC area are reported in Table 2.0.

2.2.4 Discussion

The preceding results provide a concrete demonstration of the potential of concurrent optimization of MLC eNVM and neural network storage requirements. The storage of fixed-point parameters can be optimized using a non-uniform encoding that protects the sign and integer bits using fewer levels per cell, and this solution provides up to $7.6\times$ area savings. As a further optimization, using k-means clustering together with MLC storage requires just a single transistor for each parameter in fully-connected NN layers. Additionally, pruning NN parameters and fine-tuning the cell level distributions protects the most frequent stored values and allows for more aggressive encoding configurations for CNN layers as well. The concurrent adoption of these optimizations reduces the

memory footprint of VGG16 to a total area of 4.9 mm², which can be reasonably integrated in a modern SoC. While this work focused on density-reliability trade-offs for a specific eNVM implementation, the interactions between fault tolerance and storage formats, as well as the co-design methodology developed in this work, proved extendable to other system, application, and technology settings.

2.3 FAULT TOLERANCE IN-THE-LOOP WITH SPARSITY AND ERROR MITIGATION

In this section, we will study how DNN weight resilience under MLC programming varies as we tune application-level choices such as quantization, clustering, and sparsity. Additionally, we will study the intersection of resilience with system-level storage choices such as selective/dynamic MLC programming, sparse encodings, and memory allocation and partition choices. This section extends previous demonstrations of CTT memory using the previously described multi-level-programmed fabricated test chip (MLC-CTT, Section 2.1.0), and our careful algorithmic and architectural codesign can overcome the high fault rates that accompany high-density MLC storage. A more comprehensive system-level evaluation is reserved for Chapter 3.

2.3.0 Sparse Encodings

Employing lossless sparse encodings to reduce required storage would be strictly beneficial for traditional memory technologies like SRAM. However, sparse-encoded values are no longer particularly fault tolerant due to the vulnerability of encoding metadata structures compared to the resilience of weight values. Therefore, additional analysis is required to determine optimal MLC storage schemes with sparse encodings ¹⁸³.

Compressed Sparse Row Storage (CSR) uses three data structures to encode a sparse matrix: an ordered list of all non-zero data elements by row (Weight Values), the column indexes of each non-

Model	LeNet5	VGG12	VGG16	ResNet50
Dataset	MNIST	CiFar10	ImageNet	ImageNet
Layers	4	I 2	16	54
Parameters	600810	7899840	138084352	24585472
Classification Error	0.83%	10.38%	35.07%	31.15%
Error Bound	0.05%	0.40%	0.57%	1.02%
Cluster Index Bits	4	4	6	7
Sparsity (% zero-valued)	89.9%	40.9%	81.1%	64.84%
16b Size	1.26MB	15.4MB	270MB	70MB
Pruned & Clustered (P+C)	316KB	3.86MB	101МВ	30.6MB
CSR	84KB	3.78MB	30.2MB	25.1MB
BitMask	107KB	3.23MB	35.5MB	11.2MB

Table 2.1: DNN models for MaxNVM evaluation including baseline classification error, computed error bound, and storage requirements under varying quantization and encoding schemes, taken from ¹⁸³.

zero element (Column Index), and counters of the non-zero elements for each row in the original matrix (Row Counter). The relative overhead of CSR varies proportionally with sparsity. Model layers with less sparsity do not benefit from this encoding, so CSR is applied on a per-layer basis where worthwhile. Convolution layer weights are typically 3-D (filter width, height, and channels), and thus must be mapped to 2-D to be packed using CSR. NVDLA has specific requirements for data formatting into the convolutional core, which dictates a 2-D mapping that can be unpacked to be compatible with the datapath ²¹².

BitMask Sparse Encoding Format is natively supported in the NVDLA framework, and it utilizes an indicator bitmask for whether each weight is zero-valued and stores all non-zero data values in packed, 128 byte aligned groups. A simple example of a bitmask-based sparse encoding is also provided in Figure 2.13. We refer to this bitmask-based sparse encoding method as **'BitM'**, and we maintain compatibility with the defined NVDLA sparse format ²¹². The effective compression of each encoding method is shown in Table 2.1, and the relative overhead of each method is dictated by the sparsity of the weights.

2.3.1 Error Mitigation

Existing, well-established schemes to mitigate the impact of faults are effective in the context of detecting and correcting eNVM-related faults for DNN inference execution. However, we find that existing schemes are often over-provisioned in light of the relative fault tolerance of stored DNN weight values, and we explore variations as well as propose stripped-down alternative strategies for maintaining accuracy of DNN inference under fault-prone MLC storage with minimal overhead.

Error-correcting codes (ECC) are a pervasive mechanism to detect and correct errors in stored values. The basic mechanism is to, at the granularity of a block or page of data, maintain metadata (e.g., Hamming-style parity bits) computed according to the correct stored values, which can then be re-computed and validated against the data block when retrieved at another time. In this way, ECC introduces a nominal, constant overhead in terms of storage requirements and error-checking. Many variants of ECC exist in terms of what metadata is collected and maintained, how much metadata is used, and whether it is hard-coded into the memory array architecture or software-implemented. ECC may be deployed at different granularities and levels of the memory hierarchy.

Different amounts of relative ECC overhead were tested for each model in the studies that follow based on the number of correctable errors encountered, but over many trials the lowest overhead configuration (maintaining 24 parity bits for each 4KB of row counter values for CSR, for example) is sufficient to safely allow MLC-3 to be used. The resulting ECC bit storage overhead is strictly less than 1% per layer across all models. Additionally, ECC is single error correct, double error detect (SEC-DED), but the probability of a DED within the row counter structure, even for the largest model considered (ImageNet-VGG16), is on the order of 10^{-18} . This probability is less than the standards for mass-produced standard memories 2 , so our design accepts this risk.

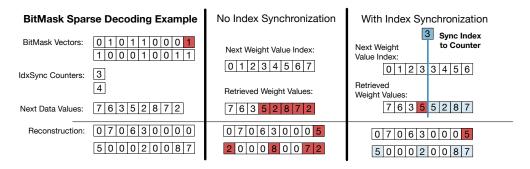


Figure 2.13: Introducing a lower-overhead approach to protecting bitmask-style sparse encoding than ECC; single bit-flip errors can be catastrophic in the BitMask structure by resulting in mis-assigned data values to the entire remaining matrix during reconstruction, but lightweight counters can prevent faults from propagating to preserve task accuracy by periodically synchronizing the index into the vector of non-zero data values, taken from ¹⁸³.

2.3.2 Lighter-Weight Error Mitigation with Index Synchronization

For even lighter weight error mitigation in the BitM sparse encoding, we additionally propose a method in which we store a counter for the number of non-zero (or zero, depending on which is lower) weight values we should have read in each 128 byte aligned block. When an error in the bit-mask causes the wrong number of weights to be read, the counter is used to dynamically update where to read from next in the packed data array. We refer to this dynamic error mitigation strategy as Index Synchronization of the bitmask or 'IdxSync'. Index synchronization of the bitmask does not correct any errors within 128 byte groups—rather, it prevents errors in previously read values from propagating to cause additional errors during weight matrix reconstruction, as demonstrated with a simple decoding example in Figure 2.13 and detailed in ¹⁸³.

2.3.3 Cross-Stack Evaluation of Sparse Encoding Choices

The model optimizations and sparse storage schemes employed to minimize the burden of storing the DNN model parameters significantly impact the fault tolerance during inference tasks. Metadata structures become a significant portion of the storage needs per DNN layer, but bit errors in

the metadata of any of the sparse encoding formats can have catastrophic impact when decoding the original matrix of values.

Thus, co-design with our carefully developed MLC eNVM fault models is required to enable dense, efficient storage. This evaluation uses the Ares fault-injection framework (Section 2.0) to quantify the impact of different encoding strategies on DNN classification error. This methodology was also applied across a wider range of models and encoding strategies with MLC-RRAM, as further explored in Chapter 3. The results of these experiments guide us in incorporating existing and proposed error correction and mitigation techniques in order to maximize the effectiveness of dense MLC eNVM storage for the execution of DNN inference.

2.3.4 Vulnerability of Encoding Strategies

Sparse encoding strategies allow fewer bits to be used when storing DNN weights, but experiments reveal that this compact data storage is in general much less tolerant of inter-level MLC faults. For all DNNs, weights that could be safely stored by their cluster index values in MLC-3 can no longer be safely stored in MLC-3 with sparse encoding. This is an incredibly important observation – as an example, there are layers of ImageNet-VGG16 in which, despite CSR enabling fewer bits to be stored, this does not overcome the area penalty of storing that layer's parameters sparsely in SLC or MLC-2 rather than storing them densely in MLC-3.

These instances can be explained by the extreme vulnerability of certain sparse encoding data structures, as highlighted for MNIST-LeNet5 (Figure 2.14). This example demonstrates that the row counter (RC) and column index data structure of CSR exhibit exceptionally high vulnerability – storing with 3 bits per cell, which is equivalent to approximately one level-to-level fault in the row counter vector, causes a pronounced degradation in DNN accuracy. This is because a single misread RC value may cause an offset when reading from the non-zero data values such that all remaining non-zero data values are incorrectly assigned during re-construction. Similarly, column index

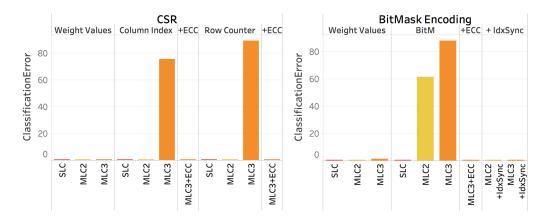


Figure 2.14: Introducing sparse encodings will have a direct impact on the resilience of our stored values. Existing, well-established schemes to mitigate the impact of faults are effective in the context of detecting and correcting eNVM-related faults for DNN inference execution, but a lower-overhead proposed technique (ldxSync) offers a compelling alternative, taken from ¹⁸³.

values (CI) are stored as relative indexes to the previous non-zero value within each row. Thus, a misread CI value may offset the remaining assigned data values, though the impact will be restricted to a particular row. The vulnerability of column indexes may be mitigated by using absolute rather than relative column indexes, but we find that this requires strictly higher overhead than integrating lightweight ECC. Relative to CSR, the bitmask-style sparse encoding is even more vulnerable, as shown in the right portion of Figure 2.14. A single bit flip in the bitmask will cause all remaining non-zero data values to be mis-assigned when reconstructing the matrix (illustrated in Figure 2.13). Thus, the bitmask structure cannot safely be stored in MLCs without some protective technique.

2.3.5 IMPACT OF ERROR CORRECTION AND MITIGATION TECHNIQUES

Looking at the classification error for the sample model in Figure 2.14, we see that both ECC and the IdxSync method enable the use of MLC-3 to store the bitmask without degrading model accuracy. The bitmask can account for a significant portion of the total storage per DNN layer (up to 65% of total bits to be stored in a given layer), so enabling denser storage of this structure can result

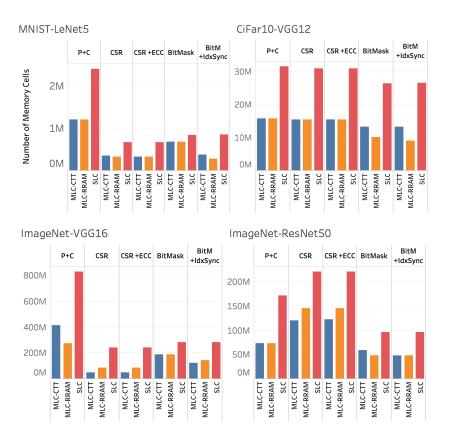


Figure 2.15: The storage strategy in terms of data format, sparse encoding, and MLC programming varies per layer for each DNN studied, with lightweight error mitigation via ldxSync corresponding to the lowest number of memory cells required for several models by enabling use of MLC-3 at no loss in application-level accuracy, taken from ¹⁸³.

in significant area benefits. For all models considered, applying IdxSync is sufficient to enable the safe use of MLCs for the bitmask. This configuration requires less storage overhead compared to ECC and reduces the decoder complexity compared to Hamming-style ECC.

Figure 2.15 summarizes the total number of memory cells required for each pairing of DNN model and encoding scheme such that there is no loss in application-level accuracy (i.e., image classification accuracy). In several cases, integrating index synchronization with the bitmask-style encoding results in the lowest required storage footprint by enabling use of MLC for the BitMask. However, the optimal storage strategy per DNN depends on model properties (per-layer sparsity, bits

per weight value) and desired optimization target (lowest area dictating fewest cells, performance, energy) based on use case, so we continually leverage the results of these fault tolerance studies in determining proposed memory solutions in Section 2.4 and Chapter 3.

2.4 MLC FEFET MEMORY FOR DNN INFERENCE AND GRAPH PROCESSING

This section considers yet another MLC-capable, promising eNVM candidate (FeFETs), develops detailed fault models and resiliency studies, and introduces several driving applications that can be co-designed with memory device properties to maximize storage density and energy efficiency. These studies, detailed in ²⁰⁴, took full advantage of the intuitions, tools, and methods described to co-design MLC CTT and RRAM towards image classification DNNs and applied them to the unique properties of FeFETs and additional, data-intensive applications with unique fault tolerance properties (a DNN for natural language processing and graph processing kernels).

This work advanced the study of FeFET as a compelling eNVM solution by modeling and quantifying the impact of device-level design choices such as size and programming scheme on the accuracy of target workloads. FeFET memories promise dense storage with competitive read characteristics, but their viability and achievable density in a realistic application use case, such as DNN inference or graph analytics, was not previously explored. Optimizations targeting device size and iterative programming schemes must be co-designed with application-level metrics and memory array-level properties to maintain benefits, especially in multi-level-cell configurations.

This section begins by evaluating the device-to-device (D2D) variation for different FeFET cell areas using a previously validated device model 58. Based on these preliminary device-level results, a proposed, customized programming scheme can increase the programming reliability for single-level cell (SLC) and multi-level cell (MLC) configurations. To quantify the relationship between resulting device characteristics and memory array performance, this work built on existing extensions of

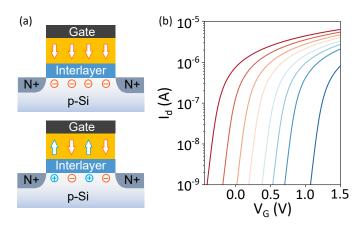


Figure 2.16: (a) FeFET device schematic; (b) Transfer characteristics of a FeFET device in 8 different resistance levels for 3-bit-per-cell (MLC-3) programming, adapted from ²⁰⁴.

NVSim⁶² tool to support a FeFET cell model and MLC programming ¹⁸³. Additionally, we develop a FeFET fault model based on the variability both at the memory and sensing circuitry level, and we extend an application-level fault injection framework ^{193,185} to evaluate the impact of FeFET device size and programming scheme on application accuracy.

2.4.0 FEFET DEVICE AND ARRAY ARCHITECTURE

FeFETs, like other emerging devices, suffer from device-to-device variation issues. The sensitivity to variation increases when the number of levels increases in a multi-level-programming scenario. This problem even gets worse in smaller FeFET devices 5. In the presented results, we account for D2D variations by employing a stochastic model of polarization switching in FeFET devices 58,204. This device-level model has been previously experimentally validated 58.

Figure 2.16 shows the programmability at 3 bits or 8 distinct levels per cell. This work employs 1FeFET AND arrays (Figure 2.17), where the source lines (SLs) connect the elements along the columns), which are more promising than the other types of FeFET memories (e.g., NOR arrays, where the SLs connect the elements along the rows) for two reasons: first, the cell size is equal or

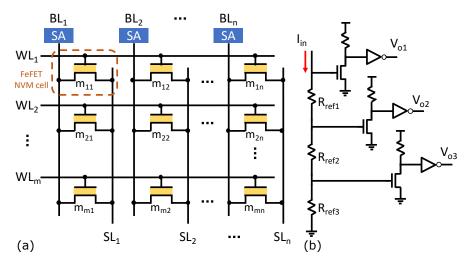


Figure 2.17: (a) AND array memory architecture highlighting a single FeFET NVM cell; (b) Customized multi-level sensing circuit for four programmed (2-bit) MLC, taken from 204 .

even smaller than NOR arrays; second, the parallel bit-lines and source-lines reduce the write disturbance 58,196 . Note that the write disturbance can be further mitigated by applying a half-bias scheme (e.g., applying $V_{write}/2$ to deselected cells)⁵.

The model uses Monte Carlo sampling on independent polarization domains in the ferroelectric layer. In doing so, it captures the essential behaviors required to perform scalability and reliability studies on FeFET memories, including (i) D_2D variation as the cell size changes (different number of $10nm \times 10nm$ ferroelectric domains); (ii) stochasticity of domain switching; and (iii) the accumulation of domain switching probability when a train of pulses are applied to the gate of the FeFET device. Thus, we can effectively project the impact of write schemes and device size on D_2D variation, and, consequently, the impact of D_2D variation on application level accuracy. Using this model and a bottom-up approach, we can effectively project the impact of D_2D variation in terms of application level accuracy.

This work employs 1FeFET AND arrays (Figure 2.17, where the source lines (SLs) connect the elements along the columns), which are more promising than the other types of FeFET memories

(e.g., NOR arrays, where the SLs connect the elements along the rows) for two reasons: first, the cell size is equal or even smaller than NOR arrays; second, the parallel bit-lines and source-lines reduce the write disturbance 58,196 . Note that the write disturbance can be further mitigated by applying a half-bias scheme (e.g., applying $V_{write}/2$ to deselected cells)⁵. Additionally device-level details are provided in Chapter 1.2.2.

2.4.1 TARGET APPLICATIONS

We evaluate the impact of FeFET reliability on target applications that are (1) data-intensive in terms of required capacity and read bandwidth and (2) have infrequent or highly batched write accesses. In embedded and mobile SoCs, the high density and compelling efficiency of FeFET memories could be crucial in enabling on-device or otherwise power-efficient execution of these critical applications, and performance will not be debilitated by potentially long writes.

Deep Neural Networks (DNNs) for vision tasks and natural language processing (NLP) are well-studied driving applications that have demonstrated compatibility with multi-level-cell (MLC) eNVM storage ^{183,59,226}. DNN inference performance and power-efficiency benefits significantly from increased on-chip memory density, particularly for storage of weight parameters that are infrequently updated. Thus, we evaluate the viability of FeFET MLC memories for two representative DNN workloads: ResNet18 for image classification (CiFar10), and the ALBERT transformer-based model for natural language understanding (MNLI) ^{88,126}. Another critical category of data-intensive workloads are those that operate on large graphs ¹⁷, such as search tasks on social network graphs, though resilience of graph formats to MLC eNVM storage has not been addressed in prior work. Thus, we additionally evaluate the viability of FeFET MLC memories for storing social network graphs with sample graphs representing Wikipedia article voting patterns and anonymized social circles from Facebook ¹³⁵.

2.4.2 ARRAY-LEVEL MLC FEFET SIMULATION

The FeFET Monte Carlo current model allows us to characterize the behavior of FeFET memory cells in isolation. NVMExplorer leverages extensions to NVSim^{62,204} to extrapolate our device-level characterization study to memory array architecture, as discussed further in Chapter 4. NVSim does not natively support FeFET memory cells, so we added a customized memory cell definition ²⁰⁴. Moreover, we modify NVSim to estimate the costs of MLC sensing circuitry.

We integrate a new memory cell definition using the energy, delay, and area results collected from SPICE simulations. Our evaluation considers two programming schemes, which are discussed in detail in Section 2.4.5. For the read operation of the SLC memories, we employ a simple voltage based sense amplifier. However, for the MLC memories, we integrated the energy, area, and delay results that we collected in HSPICE simulations for our design discussed in Section 2.1.1. For the single-pulse programming scheme, we derive the set and reset energy and latency for different cell sizes from HSPICE simulations. The write energy and latency for other programming schemes use the average number of set and reset programming pulses computed by sampling 1500 FeFET cells for D2D variation, in addition to estimated energy and latency due to write circuitry.

To model the write energy, we first obtained the energy required to program the FeFET cell for a single reset pulse and a single set pulse with different gate sizes using SPICE simulation. Second, we calculated the average number of sets and resets needed for 1500 FeFET cells (to have enough data points) to reach their target levels for different programming schemes (single-pulse and write-verify). Then, we can calculate the write energy of FeFET cells by multiplying the number of set and reset pulses to the obtained pulse energy. An additional change to improve model fidelity is that FeFETs fabrication requires adding a ferroelectric layer on top of a traditional MOSFET gate, so the gate equivalent oxide thickness is increased and, therefore, the gate capacitance. We captured this in NVSim by setting FeFET gate capacitance 1.73 × bigger than the default CMOS gate capacitance 58 .

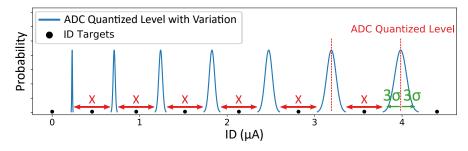


Figure 2.18: ADC quantized levels with variation as a Gaussian function with 3σ deviation of 5% and target currents for 3-bit FeFET MLC, taken from 204 .

2.4.3 SENSING CIRCUIT DESIGN

For a 1-bit read operation, this work employed a simple voltage-based sense amplifier characterized in SPICE. The same sense amplifier model is employed to build a parallel sensing scheme for MLC FeFET operation. As discussed in Section 2.1.1, a parallel sensing scheme compares the current from a memory cell against $2^n - 1$ reference levels and returns an n-bit binary word. Figure 2.17 depicts the circuit schematic of the sensing circuit for a 2-bit read operation. The operation is similar to a Flash ADC and is similarly affected by D2D variation, which in turn has an impact on memory reliability, as in the RRAM and CTT fault models developed in Section 2.1.1.

We model the effects of D2D variations on transistor dimensions, resistance values, etc., as a Gaussian function with a 3 σ deviation of 5%. As a result, the quantized levels show variability proportional to the threshold currents. Based on this constraint, we propose spacing the MLC programming currents such that the sensing threshold distributions are equally spaced. This approach uses the extra margins in low-current levels to distribute the read errors due to sensing threshold shifts more evenly across the entire programming window, as depicted in Figure 2.18. Figure 2.18 shows a MLC-programmed schematic at 3 bits or 8 distinct levels per cell, displaying changes in variation per ADC quantized level over the programmable range of current values. This is distinct behavior from the technologies considered in previous sections (i.e., CTT and RRAM), but is sim-

ilarly expressible as a set of normal distributions for integration with application-level resilience analysis. The energy, latency, and area for the resulting design is incorporated in NVSim to model sensing at the memory array architecture level.

2.4.4 Cross-Application Fault Injection Framework

Evaluating device non-idealities at the application level requires a balance of fault model accuracy and simulation performance. We approach this problem by building a fault-injection simulator that integrates with existing application-level frameworks. As the focus of this work is on DNN and graph applications, we leverage two popular Python packages, namely PyTorch ¹⁸¹ and SNAPpy ¹³⁶ to execute representative workloads. Previous work has explored the evaluation of faults in DNN models ¹⁹³ and extended the study to eNVM errors in DNN applications ⁶⁰. We extend existing MLC eNVM fault modeling to simulate FeFET memories based on the model discussed in Section 2.1.1 and ADC quantized level variations. Moreover, we use a general input interface to process parameters from both DNN and graph workloads, a major extension and generalization of the capabilities of Ares ^{193,183}. For each target application, the fault injection tool reads in the application data to be stored in FeFET-based memory and applies a quantization transform followed by MLC encoding based on the selected configuration, quantifying application-level accuracy impacts over many random trials.

Stored data values are assigned to FeFET current levels, and we sample the current based on the FeFET Monte Carlo model. The currents computed in this first sampling process are then compared against current sensing thresholds which are also sampled based on ADC variations. The sensed currents are then converted back to quantized values and used to repeat workload execution and evaluate the effect of memory faults on the target application.

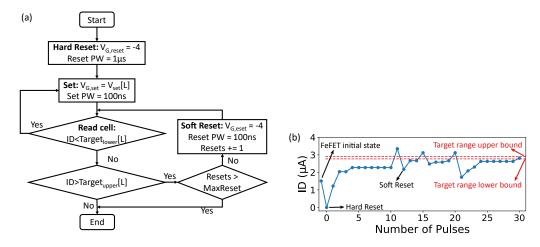


Figure 2.19: (a) Proposed write-verify programming schematic; (b) Example of pulse by pulse level tuning process of an individual FeFET device into the target range, taken from ²⁰⁴.

2.4.5 Optimization Study – Programming schemes

We consider two approaches for programming multi-level FeFETs, (i) single pulse; and (ii) write-verify. The single-pulse approach first performs a hard reset of the device (a -4V, 1 μ s pulse), followed by one pulse of amplitude dictated by the target level. An issue with this approach is that as we move to smaller FeFET devices, (i.e., lower number of domains) the D2D variation increases to a point where the rate at which we mis-read a programmed level will increase significantly. While appealing in terms of simplicity, the single-pulse scheme is not robust to D2D variation for smaller FeFET cells, as evident in the overlap of resulting current distributions in Figure 2.20 for 50-domain cells.

To solve this problem and realize multi-level FeFET cells with tightened cell distribution, we proposed a write-verify scheme described in Figure 2.19. First, we do a hard reset on the device to make sure that the device reaches the lowest state. Next, the set voltage is chosen based on the target level with the pulse width of 100*ns*. The set pulse is then applied to the gate of the FeFET device. After the first set pulse, the cell is read and the device current is measured. If the measured current is smaller than the lower bound of the target programmed range, another set pulse is applied to the

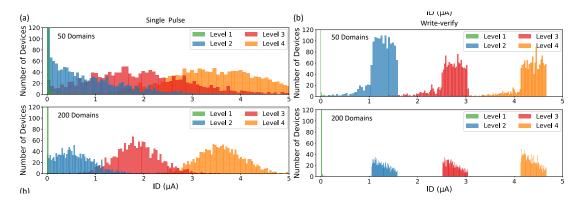


Figure 2.20: (a) Current distribution of 2-bit FeFET MLCs with single pulse programming for 1500 cells; (b) Current distribution of 2-bit FeFET MLCs with write-verify programming for 1500 cells, taken from ²⁰⁴.

gate of the FeFET. However, if the measured current is larger than the upper bound of the target range (and we have not exceeded the maximum number of resets), a soft reset pulse is applied on the gate of FeFET. Finally if the device is in the targeted range we stop the programming.

The write-verify scheme proposed in 204 (Figure 2.19) helps combat D2D variations by producing tighter level distributions. This scheme leverages the stochastic behaviour of the FeFET domain switching to reach the targeted level. The advantage of this approach is that it reduces the overhead in the write circuitry while reaching the tighter cell current targets. This scheme cannot be employed for other emerging devices such as PCM since they need programming pulses with increasing amplitude and width to reach the desired current targets. The soft reset pulse has two advantages: First, its pulse width is $10 \times$ shorter than the hard reset pulse, which will increase the performance of the write. Second, it does not fully reset the device to the zero state because of the shorter pulse. This gives us more control to set the device into the targeted range. A similar strategy has been applied to RRAM devices 33 , but this has never been studied for FeFETs before to the best of our knowledge.

Our evaluation showed that only a small subset of devices (less than 0.1% for 200-domain cells) fails to reach the target range after 10 soft reset cycles. Therefore, we fix the maximum number of

soft resets to prevent prohibitively long and energy-consuming programming sequences. The programming sequence is terminated either if the measured device current is within a target range, or if the number of applied soft reset pulses reaches its maximum count (Figure 2.19).

Figure 2.20 shows the current distribution of 1500 2-bit FeFET MLCs programmed using single-pulse and write-verify programming schemes. Figure 2.20 shows that there are large overlapping areas between different levels if the single-pulse scheme is used. The level distribution overlap worsens when the number of domains is reduced from 200 to 50. Figure 2.20 shows that using write-verify scheme, overlaps can be mitigated for 200 domain cells. However, even with the write-verify scheme, there is some overlap in 50 domain cells.

2.4.6 Memory design trends

For each of the programming schemes discussed in Section 2.4.5, we carefully consider the fault characteristics of the FeFET memory as we vary the number of programmed levels per cell and the cell size. The highest resulting inter-level fault rate per cell design and programming scheme is shown in Figure 2.21. Our fault model additionally captures the asymmetry of inter-level fault rates (i.e., when it is more likely for the programmed level to be mis-read as a lower level than a higher one), which is increasingly apparent in 2-bit and 3-bit programming. Thus, the raw maximal fault rate alone does not capture the complexity of the potential impact of MLC FeFET behavior on application accuracy. However, we note that the fault rates when we program 2 bits per cell, on average, exhibit more asymmetry than 3 bits per cell (i.e., it is more likely for a cell's value to be mis-read as a lower value than a higher one). Figure 2.21 highlights those fault rates which are strictly higher than similarly provisioned MLC-programmed RRAM storage as a proxy for the viability of different FeFET memory configurations.

As discussed in Chapter 1.1, there are multiple classes of important workloads that can be tolerant of long write latency and would benefit immensely from increased storage density and effi-

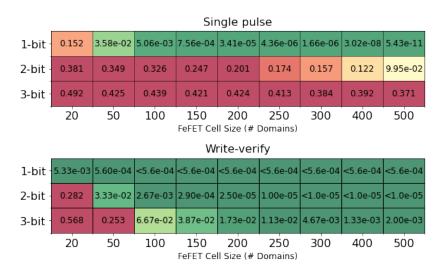


Figure 2.21: Shmoo plot of maximum read fault probability as a function of cell size and bits per cell for single pulse vs. write-verify programming, colored according to whether the raw fault rate is higher (red), lower (green), or similar to MLC-programmed RRAM taken from ²⁰⁴.

ciency. However, these workloads may exhibit different resilience to MLC fault characteristics. In light of the relatively high error rates for MLC programming across cell size shown in Figure 2.21, it is critical that we evaluate application-level implications to identify the densest allowable storage scheme without degrading application accuracy.

2.4.7 Application-level Evaluation

We evaluate two DNN workloads which differ both in terms of target application and eNVM utilization. The first is image classification using ResNet18. In this case, we consider a system in which the full model is stored in the FeFET memory. Figure 2.22 shows how application error relative to baseline classification accuracy increases significantly as cell size decreases across programming schemes. For 1-bit, single-pulse programming (no verify), the minimum FeFET cell size that preserves inference accuracy is 300 domains. Introducing the write-verify scheme allows the memory cell size to be scaled more aggressively, with 1- and 2-bit storage displaying minimum cell sizes of 50

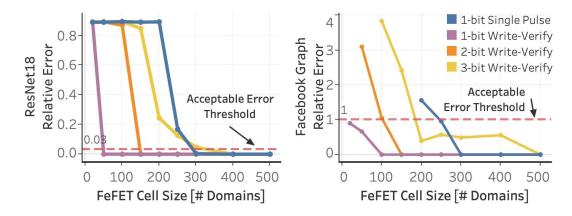


Figure 2.22: Fault injection studies are performed per cell size and programming scheme to evaluate resulting application error, as measured by relative degradation from baseline accuracy. The minimum cell size that keeps relative error below the acceptable threshold dictates densest storage per scheme, taken from ²⁰⁴.

and 150 domains, respectively.

The second DNN we consider is natural language inference running on ALBERT. BERT-based DNNs are often re-used across multiple tasks (e.g., translation vs. sentiment analysis), so we propose to store the shared embedding parameters in FeFET memory, while the task-specific portion of the network is stored separately, an approach that has proven successful in maximizing energy efficiency for multi-task DNN inference ⁵⁹. In this case, the 1-bit, single-pulse configuration requires 200 domains, while for the write-verify cases, 1-, 2-, and 3-bit configurations can be scaled down to 50, 150, and 150, respectively (summarized in Table 2.2). Compared to the relative vulnerability of 3-bit MLC for ResNet18, partitioning ALBERT to store more vulnerable parameters in SRAM provides the opportunity for more aggressive scaling in terms of FeFET cell size.

To determine the viability of different FeFET cell designs and programming schemes for graph analytics tasks, we measure the average impact to the accuracy of a set of breadth-first search queries on the example graphs as a proxy for maintaining network structure and resilience to faults ¹⁷. We store input graphs as adjacency matrices and perform fault injection to quantify resilience to varying MLC storage and programming schemes. The relative error increase for the Facebook graph,

BPC	Programming	ResNet18	ALBERT	Wiki	Facebook
I	Single Pulse	300	200	250	250
I	With Verify	50	50	50	20
2	With Verify	150	150	150	150
3	With Verify	400	150	400	200

Table 2.2: Summary of minimal cell size in number of domains per programming scheme and bits per cell (BPC) without application accuracy degradation, adapted from ²⁰⁴.

for example, (Figure 2.22), demonstrates workload-specific sensitivity to FeFET fault characteristics distinct from ResNet18 in terms of allowable cell size. The minimum cell size per workload and per programming scheme is summarized in Table 2.2. For example, we see that the Wikipedia input graph is resilient to FeFET storage at 50 domains, while both graph workloads are amenable to 2-bit FeFET storage down to 150 domains. Due to the varying sparsity and structure of the two graph workloads studied, we observe different fault resilience in the 3-bit configuration, and the Wikipedia voting graph degrades in terms of average query accuracy when using a cell with fewer than 400 domains. Each mis-read value in the adjacency matrix corresponds to an erasure or erroneous addition of a connection between two graph nodes, and it is possible that each mis-read value has a disproportionate impact on a graph that is less clustered and more sparsely affiliated (as in the Wikipedia graph) than one with more disparate, strongly connected social circles (as in the Facebook graph), but further analysis would be required to justify this argument.

2.4.8 Discussion

Our results show that implementations targeting inherently robust applications can lead to more efficient memory array architectures. For example, in most of the workloads we considered in this work, the SLC configuration using a write-verify programming scheme results in a memory array design which has the best combined density and read performance. This result can be explained by considering that replacing the SLC configuration with MLC2 requires increasing the cell size by $3\times$

Workload	Opt. Scheme	Storage	# Cell	Area	Read Latency	Read Energy	SET Latency	SET Energy
WOIKIOAU		Needs	Domains	(mm^2)	(ns)	(pJ / bit)	(µs)	(pJ / bit)
ResNet18	SLC	24MB	50	1.686	1.866	0.461	1.47	0.745
ALBERT	MLC2	4MB	150	0.313	1.20	0.189	1.80	0.438
Wikipedia	MLC2	6MB	150	0.682	1.268	0.278	1.80	0.422
Facebook	MLC2	2MB	150	0.241	0.976	0.169	1.80	0.234
SRAM (16nm)	Ref.	4MB	N/A	3.9	1.3	0.5	0.001	0.5

Table 2.3: Summary of optimal programming schemes, device characteristics, and resulting memory array characteristics provisioned per workload, adapted from ²⁰⁴.

in order to preserve application-level accuracy.

Though all of the examined workloads are amenable to MLC FeFET storage, in order to derive the most efficient FeFET storage solution per application, we must consider the achievable zero-accuracy-loss storage density in conjunction with memory array performance, area, and energy characterization. For example, we find that SLC FeFET under the write-verify scheme can be safely used down to a cell size of 50 domains or smaller (Table 2.2), so this programming scheme consistently provides highly dense, zero-accuracy-loss memory. On the other hand, our simulation on ALBERT give us an opportunity to uncover the potential of MLC storage solutions for data-intensive, fault-resilient workloads.

Figure 2.23 also demonstrates that 2-bit storage is a compelling design in terms of density and read characteristics, with 3-bit showing competitive performance in all metrics, though not providing the pareto-optimal configuration highlighted in Figure 2.23 and summarized for each workload in Table 2.2. The storage density achieved here is a full order of magnitude higher than that of 16nm SRAM, in which a 4MB array requires about $4mm^2$ according to NVSim evaluation, cited as a reference point in Table 2.3. Similarly, the observed read latency and energy per bit tends to match or improve upon SRAM. We repeat analysis of area, latency, and read energy per bit of provisioned memory arrays per-application to identify FeFET memory arrays that minimize latency and energy. In this way, the FeFET arrays highlighted in Table 2.3 will cause minimal impact to workload execution time while providing high density, energy-efficient embedded storage.

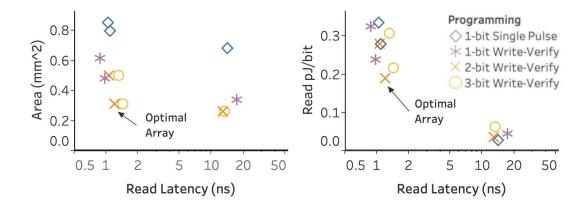


Figure 2.23: For an example workload (ALBERT, 4MB capacity), we show key characteristics for FeFET memory arrays with acceptable application accuracy across programming schemes and NVSim optimization targets. From these, we select optimal array configurations summarized in Table 2.3 that minimize array area, read latency, and read energy per access, taken from ²⁰⁴.

It is important to note that these design points are not fully optimized, which indicates that even higher density could be achieved with architectural enhancements ^{183,185}. For example, incorporating lightweight error mitigation or error correction strategies could enable use of even higher density storage solutions, such as 3-bit MLC with smaller cell sizes. Alternatively, more sophisticated programming schemes could be used to further tighten the current distribution of MLC FeFET configurations. However, more complicated programming schemes would add area, latency, and energy overhead due to write circuitry.

This work is a demonstration of the viability of MLC-programmed FeFET eNVMs with an application-driven evaluation of these memories in light of DNN and social network graph resilience and memory access patterns. The presented methodology and studies provide exposure to the critical trade-offs among write costs, storage density, fault characteristics, and application impact that will drive future development of highly-dense FeFET memory solutions. These results build confidence that MLC FeFET storage could be effectively leveraged for both deep neural network inference and graph processing tasks, even in the absence of more involved optimizations. Yet, these

results are limited by the lack of integration with system-level optimizations and a more comprehensive system evaluation, which is the focus of Chapter 3.

 $If it's so \ be autifully \ arranged \ on \ the \ plate, you \ know \ someone's \ fingers \ have \ been \ all \ over \ it.$

Julia Child, "The French Chef"

3

MaxNVM: Maximizing Memory Efficiency for ML Accelerators

DEEP NEURAL NETWORKS (DNNs) are a critical, driving application in a variety of systems and domains, from object-detection and object-tracking for self-driving cars, to classification and data analysis tasks on embedded sensor nodes and implanted devices, to diagnoses and drug discovery

in the medical field ¹⁹². Particularly in deeply-embedded computing environments, efficiency is paramount – both in terms of performance, operating power and thermal impacts, and in terms of energy usage and potential impacts on battery life and device lifetime.

For state-of-the-art DNN hardware accelerators, fetching weights from DRAM persists as a primary performance and energy bottleneck, limiting inference efficiency ¹⁹². Ideally, DNNs weights would be stored entirely in on-chip memory, such that DNN inference can be fully executed on-chip without costly DRAM data fetching. However, even with aggressive weight compression and layers of algorithmic optimization, capacity requirements of modern DNNs often remain unrealistic for storage in on-chip SRAM.

Emerging, embedded non-volatile memory (eNVM) technologies are one promising solution for eliminating DRAM inefficiencies, as described in Chapter 1. eNVMs provide high-capacity, low read-latency storage and are significantly more dense than SRAM. Many eNVMs, including RRAM, PCM, and CTT, also have multi-level cell (MLC) capabilities, allowing multiple bits to be packed into a single device to further increase density. eNVMs are not perfect—two main drawbacks are decreased reliability and high write latency. First, while single-level eNVMs are reliable, achieving the ultra-high densities necessary to eliminate DRAM for DNN inference requires aggressive MLC designs. However, these MLC designs can incur high memory access fault rates. Second, eNVMs offer fast read access, typically on the same order as SRAM. However, writes can be orders of magnitude slower, as well as more costly in terms of energy or operating conditions, as they often alter the physical property of the storage material. Given DNN inference in embedded settings often require infrequent weight updates and are implicitly fault tolerant (as explored in Chapter 2), we identify MLC eNVMs as a potential solution to improve DNN inference efficiency by storing weights on-chip and eliminating DRAM accesses.

With algorithm-hardware co-design, DNN inference is a near-ideal fit for MLC-eNVM as the benefits (high density, fast read, low-power) can be had without suffering from the cons (fault-

prone, slow writes). This chapter focuses on the demonstration and evaluation of MLC eNVM technologies applied towards highly-efficient DNN inference settings, in each case employing optimizations that span and exhibit inter-dependencies among multiple layers of the computing stack.

As an efficient baseline, MaxNVM starts at the algorithmic level, applying clustering, pruning, and sparse weight encoding (CSR and BitMask). To optimize for fault-prone MLC storage, MaxNVM co-optimizes the sparsity of DNNs with the storage density of MLC eNVMs, which directly impacts the reliability of the storage medium, to reduce the memory requirements without sacrificing accuracy. MaxNVM unveiled that sparse-encoded weights to be more vulnerable than dense weights, which limits MLC eNVM density, as previewed in Chapter 2.3. This chapter pushes the limits of storage density through increasing the levels-per-cell in MLCs and by applying protective mechanisms, ECC and IndexSync (proposed), in-the-loop with application-level metrics like accuracy and achievable frame rate and system constraints like total on-chip area and hardware accelerator bandwidth requirements.

To demonstrate the efficacy of our co-design approach on the actively evolving eNVM landscape, MaxNVM conducted detailed evaluations of two fundamentally different eNVMs: RRAM and CTT. Both are extremely dense, MLC-capable, and can be made CMOS-compatible for on-chip integration with logic in advanced process technology nodes. CTTs have fast read latency and are very low-power, but incur inordinately long write latency. Compared to CTTs, RRAM has a faster write latency, but is less energy-efficient.

The MaxNVM project conducted a crucial first step in developing models for RRAM and CTT, including MLC-programming, by extending NVSim⁶² in ways that were, in turn, significantly generalized and expanded in other research projects I have led or contributed to over the course of this thesis work ^{204,59,226,185}. RRAM cell parameters and fault models are taken from published work, and MaxNVM implemented two models to represent projected ^{268,269,53} and demonstrated ^{252,255,128} RRAM scaling to smaller technology nodes. CTT parameters and fault distribu-

tions are derived from a measured chip prototype 183,157.

Several studies presented in this chapter leverage NVDLA—a state-of-the-art, industrial-grade CNN accelerator. NVDLA is an open-source accelerator architecture from NVIDIA, providing performance modeling and comprehensive documentation so as to be extendable and effective as a system model for academic projects ^{212,178}. Experiments in Section 3.1 show that even large models, e.g., VGG16 and ResNet50 ^{213,88,197}, can reasonable fit on-chip with the MaxNVM co-design approach. CTT results in the lowest energy-per-inference design point at up to 3.2× reduced energy per inference. In a system where weights are updated more frequently, RRAM presents a nice compromise of writing weights orders of magnitude faster while sacrificing approximately 20% energy efficiency. In a case study, we constrain all on-chip memory to fit within 1mm² and sweep the percentage given to SRAM and eNVM. The study concludes a balance of eNVM and SRAM is ideal to maximize performance and efficiency.

After detailing the system configurations and evaluations and discussing the compelling results of MaxNVM, this chapter briefly summarizes two additional projects in specialized DNN accelerators to which I have contributed. In the contexts of both Multi-Task Image Processing (MEMTI 59, Section 3.2) and Natural Language Processing (EdgeBERT 226, Section 3.3), total system efficiency and on-chip storage density are significantly improved by proposals integrating multi-level-cell-programmed eNVMs, and each system design benefited from cross-computing-stack considerations in data format, fault tolerance, and memory technology characteristics.

3.0 END-TO-END, CO-DESIGN METHODOLOGY FOR MLC ENVM

The MaxNVM co-design strategy incorporates optimizations and techniques at algorithmic and architectural levels. After developing a fault model based on technology-specific device characteristics and SPICE models of sensing circuitry, we use a previously-validated fault injection framework

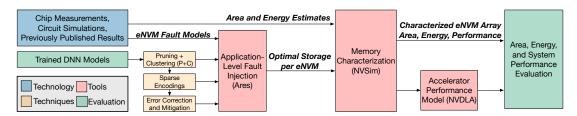


Figure 3.0: MaxNVM summary of the tools, optimizations, and intermediate results used in final system evaluations; technology characteristics, specific system requirements, and a variety of application-level optimization opportunities inform the proposed MaxNVM system designs, taken from ¹⁸³.

to determine whether a given storage scheme impacts DNN accuracy ¹⁹³, as described in Chapter 2. We leverage and extend well-known tools to model the energy, performance, and area of the proposed systems at the memory-array ⁶² and system-architecture ¹⁷⁸ levels. The interaction of these methods as they contribute to the final evaluation is summarized in Figure 3.0.

3.0.1 MODEL OPTIMIZATIONS

Proposed memory systems are evaluated against a competitive and realistic baseline by enforcing iso-accuracy and incorporating common optimizations on the chosen DNNs, as summarized in Table 2.1. All presented energy, performance, and area improvements maintain model accuracy within computed Iso-Training error bounds, as defined in ¹⁹² and briefly described in Chapter 1.1.1. We consider DNNs of different sizes: one CNN model for the small MNIST hand-written digits dataset, two more realistic and well-known DNN models (VGG16 and ResNet50) for the much larger ImageNet dataset, and another VGG-like topology for the mid-size CiFar10 dataset to span the gap between these cases ^{213,88,125,197}. Magnitude-based weight pruning with retraining is used to sparsify DNN weights, as the models are widely known to be over-parameterized. The resulting proportion of zero-valued weights is listed in Table 2.1, and further model optimizations are performed without retraining.

One popular technique to reduce the number of bits required to store each DNN weight is to

reduce precision using fixed-point quantization. Depending on the dynamic range of the DNN weight values, the number of integer and fractional bits can be drastically reduced, even to a few or a single bit per weight at some loss in model accuracy ^{99,14}. Another way to reduce the number of bits required to represent each weight value is to use *k*-means clustering for each layer of the DNN ¹⁹². For the models considered, we found that all the weight values within a given layer can be represented by 16 to 128 unique clustered values at no loss of accuracy. Thus, each weight can be encoded as a 4 to 7 bit cluster index value, with a simple look up table per layer to map indexes back to values. We find clustering uses strictly fewer bits per weight than fixed-point quantization without significant re-training for all DNNs (Table 2.1). For more details on model optimizations and DNN simplification techniques, please see Chapters 1.1.1 and 2.3.

3.0.2 Error Correction and Mitigation

Vulnerability to faults changes when weights are stored using sparse encodings, as explored in Chapter 2.3. A simple strategy to mitigate the increased classification error is to store more vulnerable structures using fewer bits per cell (e.g., SLC or MLC₂)⁶⁰. We additionally explore when it is beneficial to incorporate Hamming-style, parity-based ECC. The configuration used is the lightest-weight ECC considered for NAND flash memories²⁰⁰. Note that if values are binary-encoded in a MLC, a single level-to-level fault is not equivalent to a single bit flip, so for precise error detection and correction, Gray coding is used to represent ECC-protected values in MLCs. MaxNVM additionally proposes and evaluates a lighter-weight error mitigation technique for bitmask-encoded sparse data, which is described in fixmeSection.

NVSim Parameter	Value		
Data Width	8 - 128		
Banks	(1 x 1) - (N x N)		
Mats	(1 x 1) - (N x N)		
	Area		
	Read Latency		
Optimization Targets	Read EDP		
	Read Energy		
	Leakage Power		

NVDLA Baseline	NVDLA-64	NVDLA-1024
Convolutional Buffer Size	128KB	256KB
Number of MACs	64	1024
SRAM Capacity	512KB	2MB
Frequency	ı GHz	1GHz
Datapath Area	0.55mm ²	2.4mm ²
SRAM BW	6 GB/s	25 GB/s
DRAM Read BW		25 GB/s
LPDDR ₄ DRAM Power	100 mW	200 mW

Table 3.0: NVSim, baseline NVDLA parameters ^{62,212,28} for use in system-level evaluation of MaxNVM, taken from ¹⁸³.

3.0.3 Memory Modeling

We integrate new memory cell definitions in NVSim using the energy, performance, and area measurements from our test chip, published work, and SPICE simulations of different sensing circuitry topologies ⁶². NVSim evaluates all possible memory bank configurations for a given capacity, optimization target, and number of bits per cell (Table 3.0). Pareto-optimal points are chosen from NVSim outputs to fit within area, performance, and/or bandwidth constraints according to use case in presented results.

3.1 Maximizing Storage Density and Inference Efficiency (Evaluation)

Leveraging the identification of maximal-density, accuracy-preserving MLC eNVM storage for DNN weights in Chapters 2.2 and 2.3, this section describes the system-level evaluation and implications of integrating such a memory solution with a DNN accelerator system.

3.1.0 Baseline Accelerator Design (NVDLA)

NVIDIA deep learning accelerator (NVDLA) is an industry-grade, open source architecture solution to accelerate DNN inference. Baseline system parameters are given in Table 3.0, and a block diagram is shown in Figure 3.1, left ¹⁷⁸. NVDLA supports CNN and FC layer execution, and all

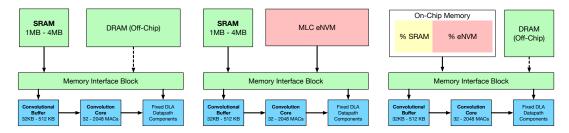


Figure 3.1: Baseline NVDLA (left) vs. Proposed Architectures (center, right) from MaxNVM, taken from ¹⁸³.

computation in MaxNVM evaluations is performed using this datapath. By demonstrating our proposed memory integrated with this system, we highlight the opportunity for eNVM integration with existing systems for efficient DNN inference.

By default, all DNN weights are retrieved per-layer and per-tile from off-chip DRAM before being staged in the convolutional buffer for computation. Any inputs to the network and intermediate results are read from and written to on-chip SRAM of a configurable size in the system design, and the memory interface block manages access both to the on-chip global SRAM buffer and off-chip DRAM resources. Inputs and weights may be configured as 8b or 16b width. The baseline NVDLA system optionally allows for bitmask-encoded sparse weights, and the MaxNVM work additionally considered CSR-encoded sparse weights. Additional architectural characteristics and design parameters are listed in Table 3.0 and described in more detail in ¹⁷⁸.

3.1.1 FULLY ON-CHIP DNN INFERENCE EVALUATION OVERVIEW

MaxNVM considers a completely self-contained inference accelerator that stores all of the weights in on-chip eNVM and does not require external DRAM, as indicated in Figure 3.1, center. This scenario would be especially important for deeply embedded applications such as IoT devices and implanted medical devices where component cost, performance, and energy constraints can be extreme, and entirely on-chip execution could additionally provide privacy benefits \$6,127.

To evaluate the area, energy, and performance of both CTT and RRAM on a per-model basis,

we use model optimization results to determine the number of bits required per encoding and perform fault tolerance analysis. These determine the appropriate number of bits per cell and total number of memory cells required for a given encoding without loss in classification accuracy (covered in Chapters 1.1.1, 2.3). Next, we use our extensions of NVSim to characterize memory systems that are sufficient to accommodate the size of each model under each encoding strategy. Finally, pareto-optimal points for characterized memories of four different design points are identified per model to minimize the read energy-delay-product and area. The resulting system performance is evaluated using the NVDLA performance model and compared to the performance of a baseline configuration relying on off-chip DRAM for weight storage in (Figure 3.1, left).

This study quantifies the increase in die area needed to store the considered DNNs entirely in on-chip memory while maintaining competitive performance and achieving reduced power and reduced energy per inference. The number of cycles to execute an inference per input image (frame) is computed using the NVDLA performance model with the characterized read bandwidth and read latency determined by NVSim for each eNVM array. Some amount of on-chip SRAM is retained in the NVDLA system evaluation (512KB or 2MB in our designated NVDLA configurations, Table 3.0) to manage storage of intermediate values between layers of the DNNs, as the write latency of CTT is prohibitively high (on the order of ms^{115}) and can still be orders of magnitude higher for MLC-RRAM than SRAM write latency (e.g., 160 - 640ns).

3.1.2 Area and Dynamic Read Energy

An interface to eNVM replaces NVDLA's interface to off-chip DRAM for DNN weights, and we demonstrate that aggressive MLC configurations of both CTT and optimistically scaled RRAM can store the sparsely-encoded, ECC protected weights of ImageNet-VGG16 (about 32MB Capacity) in $2mm^2$ and $1.3mm^2$, respectively. Thus, as a result of exhaustive design space exploration and a rigorous evaluation scheme, all DNN weights can reasonably fit in on-chip, non-volatile

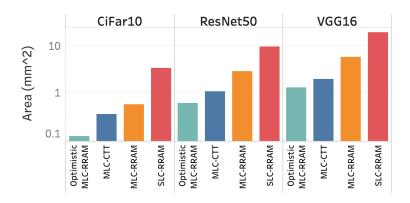


Figure 3.2: Array-Level area of various on-chip memory solutions for storing all DNN weights in eNVM, taken from ¹⁸³.

memory in an area equivalent to 1-2MB of SRAM capacity in modern process nodes. Even SLC-programmed RRAM fits all of ResNet50 (12MB capacity under sparse encoding) in under 10mm² (Figure 3.2). With additional model optimizations, or more relaxed error bounds, MLC configurations could be leveraged even more aggressively at some loss in DNN classification accuracy, depending on use case. This study strictly avoids loss of accuracy. Even so, these models consume reasonable area for on-chip local memory storage using MLC eNVM configurations.

Figures 3.2 and 3.3 summarizes the area and dynamic read energy per access of four distinct onchip memory solutions. The area for each model is the read-energy-delay-product optimal point characterized by our extensions to NVSim that will hold all the DNN parameters in the most optimal encoding for each memory technology as determined in Chapter 2.3. Even relative to storing the same optimized and sparse-encoded weights in SLC-RRAM, the MLC-CTT array requires an average of 9.6× less area while maintaining competitive performance within 10% of the NVDLA baseline. Furthermore, our MLC extension of the SLC-RRAM achieves an average area benefit of over 3.2×, and the optimistically scaled MLC-RRAM (Optimistic MLC-RRAM) enables an average area benefit of 20×. The larger relative area benefit between MLC-CTT and the Optimistic MLC-RRAM for CiFar10-VGG12 (Figure 3.2) derives both from the inherent storage density of

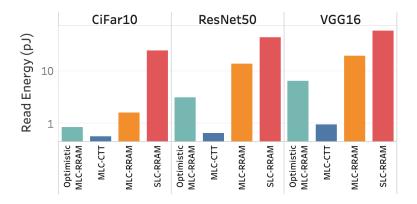


Figure 3.3: Array-Level read energy per access of various on-chip memory solutions for storing all DNN weights in eNVM, taken from ¹⁸³.

the Optimistic MLC-RRAM and the fact that more bits per cell can be safely used with the Optimistic MLC-RRAM (consistently 3 bits per cell) than with MLC-CTT (2 bits per cell) due to their respective fault characteristics and the impact on classification error, as exposed in Chapter 2.3.4.

The dynamic read energy for each of the memory proposals varies by orders of magnitude, even for equivalent-capacity characterized array (Figure 3.3). MLC-CTT is consistently lower energy per access than even the Optimistic MLC-RRAM solution by over $4\times$, and also tends to maintain a lower read latency and higher read bandwidth (up to 9 GB/s).

3.1.3 SYSTEM PERFORMANCE

Considering a real application-specific metric (frames-per-second), even the SLC consistently meets a target of 6ofps, for example, even for the largest models. Though eNVM proposals do not surpass the maximal FPS possible with NVDLA (e.g., Figure 3.4), the best performance for each model consistently exceeds 6o frames per second with the NVDLA-1024 configuration, well above the image processing frame rate for standard motion pictures ¹⁶⁰. The latency overhead of MLC sensing tends to negate the effective bandwidth increase of MLC storage. This is seen in Figure 3.4; SLC-RRAM is competitive with MLC-CTT in terms of frames (i.e., inferences) processed per second. We note

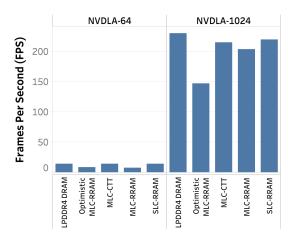


Figure 3.4: Achievable frame rate (inferences-per-second) for baseline NVDLA (LPDDR4 weight memory) vs. Proposed Architectures from MaxNVM, taken from ¹⁸³.

that these performance estimates do not include the decoding overheads for sparse encodings and cluster index values. However, the overhead to reconstruct and decode weight values for both strategies will be minimal and consistent in the baseline and eNVM-based systems.

Figure 3.5 compares the average power consumption and total energy per inference for ResNet50 when operating at maximum performance. We consider two fixed NVDLA baseline datapath configurations given in Table 3.0, for which the power consumption is publicly available ²¹², and assume additional power consumption of LPDDR4 DRAM running at 1GHz is 200mW to form a conservative bound of potential improvements via integration of on-chip eNVM ²⁸.

The NVDLA-64 design is representative of a resource-constrained DNN accelerator, and the energy consumption of memory accesses for weight fetches is *reduced by over 100* \times (DRAM vs. 1 mm^2 of on-chip MLC-CTT), resulting in overall average system power reduction of 3.2 \times . NVDLA-1024 provides considerably higher performance (Figure 3.4) and overall higher efficiency inference (lower dynamic energy per inference), though the average power is higher. In this scenario, we see similar trends in MLC-CTT reducing energy consumption due to weight fetching by up to 13.6 \times . However, due to the higher baseline power of the larger convolutional core and buffer, the total

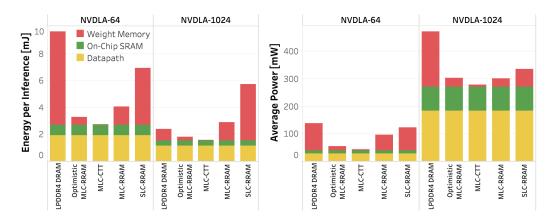


Figure 3.5: Total energy-per-inference and average system power for baseline NVDLA (LPDDR4 weight memory) vs. proposed architectures from MaxNVM, taken from ¹⁸³.

relative system power reduction is up to $1.6 \times$.

Of the three evaluated MLC proposals, MLC-CTT achieves the best system performance, lowest power, and lowest energy per inference. Compared to the optimistically scaled MLC-RRAM, higher read bandwidth and lower energy per read result in 20% overall lower energy per inference for MLC-CTT for NVDLA-64. These demonstrated benefits are relative to energy and performance characteristics of convolutional models using a highly-optimized CNN accelerator datapath that maximizes re-use of fetched values, so energy reduction due to memory fetches would be increasingly beneficial in other resource-constrained contexts that exhibit less re-use of fetched parameters, as explored further in future works, for example in Sections 3.2 and 3.3. Compared to the energy usage of the NVDLA datapath and SRAM in isolation for ResNet50 inference, we see that the power for the eNVM solutions are a fraction of the the compute pipeline for the aggressive MLC proposals (either optimistically scaled MLC-RRAM or MLC-CTT based on our fabricated test chip), and the overall system power reduces by up to 40% with the added benefit of low leakage power and the ability to turn off the eNVM between inferences and retain weight values. Table 3.1 gives a summary of area, energy, and performance for optimal storage for each design point.

Model	Memory Tech	Encoding	BPC	[MB]	Area [mm ²]	Read [ns]	FPS
CiFar10	Opt. MLC-RRAM	BitM+IdxSync	3	4	0.12	5.1	132
	MLC-CTT	BitMask	2	4	0.35	1.6	2286
	MLC-RRAM	BitM+IdxSync	3	4	1.3	4.9	633
	SLC-RRAM	BitMask	I	4	3.4	1.7	2967
VGG16	Opt. MLC-RRAM	CSR+ECC	3	32	1.3	4.2	102
	MLC-CTT	CSR+ECC	3	32	2.0	2.0	142
	MLC-RRAM	CSR+ECC	3	32	5.7	3.2	131
	SLC-RRAM	CSR	I	32	19.2	5.2	147
ResNet50	Opt. MLC-RRAM	BitM+IdxSync	2	I 2	0.6	2.I	147
	MLC-CTT	BitM+IdxSync	2	12	1.0	1.9	215
	MLC-RRAM	BitM+IdxSync	2	12	2.8	1.4	203
	SLC-RRAM	BitMask	I	12	9.6	2.5	219

Table 3.1: Summary of optimal storage per eNVM proposal, characterized per DNN. 'BPC' is the maximum number of bits per cell used across DNN layers, 'MB' is the approximate capacity, 'FPS' is the maximum frames per second for NVDLA-1024, and 'Read' is the Read Latency of each eNVM array in *ns*, taken from ¹⁸³.

3.1.4 REDUCING OFF-CHIP DRAM ACCESS FOR LARGER MODEL SIZES

New DNNs tend to improve accuracy by increasing model size ¹⁹². To understand what happens when a model does not fit in on-chip eNVM, we consider a hybrid solution, shown in Figure 3.1, right. We allocate a fixed on-chip area allowance for all memory (SRAM and eNVM), and weights and activations that do not fit on-chip are stored and fetched from DRAM.

We choose an on-chip memory area budget of 1mm², enough to accommodate about 1MB of SRAM under various NVSim optimization targets. With eNVM, smaller DNNs fit within this area constraint. The results in Figure 3.6 are for the largest DNN, ImageNet-VGG16, and highlight the performance impact of incorporating varying amounts of on-chip space to eNVM, which effectively increases on-chip memory capacity in a constrained environment. Note that the eNVM is not used as cache for DRAM; on-chip CTT/RRAM and DRAM store mutually exclusive sets of model weights and both are fed directly into the accelerator's datapath. By partitioning fixed on-chip memory area between SRAM for intermediate storage and MLC eNVM for weight storage, we reduce costly DRAM accesses while drastically increasing total on-chip memory capacity. For each possible

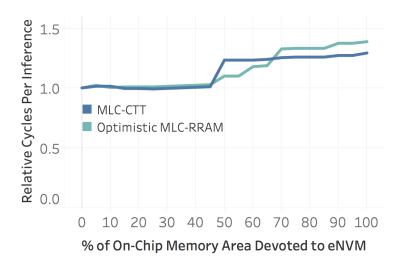


Figure 3.6: Partitioning a fixed on-chip area between SRAM and a highly-dense eNVM solution is strictly beneficial until the SRAM capacity becomes insufficient to store intermediate results, which sharply degrades inference performance (e.g., at 45-60% of area devoted to eNVM), taken from ¹⁸³.

area partition, we use NVSim to characterize the maximal capacity and minimal read latency and energy within that area constraint and select pareto-optimal points.

Depending on how the on-chip memory area is partitioned, inference execution could be bottlenecked by weight retrieval from DRAM for weights not in the eNVM, or by read/write traffic of
intermediates that do not fit in SRAM. We extend the NVDLA performance model to selectively
read certain weights from eNVM rather than DRAM, and maximize eNVM benefits by greedily
selecting the weights of otherwise DRAM-bottlenecked layers to store first. There is some initial
benefit from alleviating the weight retrieval DRAM bottleneck when some amount of eNVM is
allotted, because the number of DRAM accesses reduces and the energy to access either eNVM is
orders of magnitude lower than accessing DRAM. However, performance sharply degrades when
on-chip SRAM storage can no longer hold the working set of intermediate values to feed the convolutional core and execution gets increasingly bottlenecked on writing to and fetching intermediate
values from DRAM. The write characteristics of both alternatives are not sufficient to buffer inter-

mediate values or to re-write weights during inference, as explored in Section 3.1.6.

3.1.5 Intermittent Operation; Benefits of Non-Volatility

As yet, we have demonstrated incredible capabilities in terms of storage density and energy efficiency of MLC eNVM storage for on-chip DNN inference. However, we have not explicitly taken advantage of the non-volatility of the proposed embedded memories, and this property becomes extremely beneficial in intermittent operation use cases. eNVM arrays effectively have a huge additional energy advantage because there is no leakage power (even leakage power for sensing is at least 100× less than SRAM leakage according to NVSim). Though we have thus far only examined the significant benefits in terms of read dynamic energy during inference, another key advantage of both RRAM and CTT memory is that they are non-volatile, and thus can be powered off when not in use and need not incur any leakage power between inferences. This makes these design points and other eNVM technologies a particularly attractive solution for edge devices, as weights can be stored locally, off-line, and without any leakage power cost between inferences, in stark contrast to the constant leakage power required by SRAM or power-hungry DRAM refresh.

Depending on how frequently inferences occur (i.e., required frame rate for an image processing task ^{264,67}), eNVMs have an inherent relative benefit by virtue of not needing to reload weight values when powered on or, alternatively, keeping DRAM and SRAM powered to avoid this cost. Figure 3.7 summarizes how the average energy per inference of the evaluated MLC eNVM solutions compares to the baseline system either remaining fully powered to retain DNN weights ("DRAM always on") vs. waking up the system for each inference ("DRAM wake up"). Each time the baseline system is woken up, there is an energy overhead to load the entire DNN's weights from main memory into DRAM and an increased inference latency to load the entire first DNN layer before the inference task can begin computation, while these costs are not applicable to eNVM technologies that retain their values when powered off between inferences.

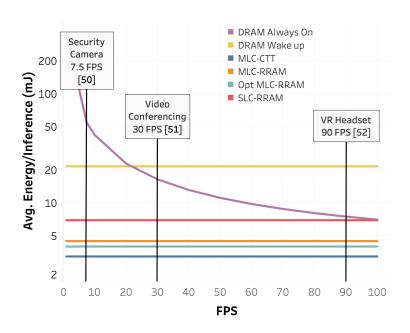


Figure 3.7: Average Energy-per-Inference for baseline NVDLA vs. proposed architectures at varying frame rates; for typical frame rates, it is strictly beneficial to leverage eNVM storage and wake the system per-inference, taken from ¹⁸³.

If the DRAM power stays constant while the frequency of inferences increases (i.e., frames processed per second, FPS), the average energy per inference decreases as the system spends less time idle. Conversely, if the system is woken up for each inference, the average energy per inference is constant so long as the frequency (FPS) can be met by the NVDLA-1024 system, which Figure 3.4 showed is clearly possible for the range of FPS considered. For frame rates below 22 FPS, such as would be relevant for object detection tasks for security camera systems ⁸², it is more energy efficient to wake up the baseline system per inference, but our proposed MLC eNVM solutions maintain a relative 5.3× (MLC-RRAM) to 7.5× (MLC-CTT) reduced energy per inference.

Energy savings are similar for the typical range of frame rates used for video conferencing and other standard image processing tasks (e.g., 30 FPS ²⁶⁴). Though the average energy per inference of the always-on baseline system approaches the performance of MLC eNVM alternatives at higher frame rates, we note that even operating at the FPS mandated to support virtual reality headsets

Model	Memory Technology	Approximate Total Write Time		
	OPTMLCRRAM	1 3 ms		
CiFario	MLC CTT	2.6 minutes		
Cirario	MLC RRAM	3 <i>3ms</i>		
	SLC RRAM	3 <i>ms</i>		
	OPTMLCRRAM	117ms		
ResNet50	MLC CTT	15.7 minutes		
iccsiveryo	MLC RRAM	94 <i>ms</i>		
	SLC RRAM	4.7 <i>ms</i>		
VGG16	OPTMLCRRAM	254 <i>ms</i>		
	MLC CTT	12.2minutes		
	MLC RRAM	636ms		
	SLC RRAM	23 <i>ms</i>		

Table 3.2: Optimistic total time to write all DNN weights per eNVM solution, taken from ¹⁸³.

(at least 90 FPS 67), MLC eNVM proposals achieve $1.7 \times$ (MLC-RRAM) to $2.5 \times$ (MLC-CTT) lower energy per inference. Thus, optimized MLC eNVM solutions are particularly compelling for applications with lower frame-rate requirements, and these benefits would be exaggerated for deeply embedded systems with less frequent wake-ups.

3.1.6 WRITE LATENCY

Table 3.2 gives an approximate total time to write each model's entire set of weights to each eNVM design point, simulating a full weight update after retraining, perhaps, based on best-case-scenario write latency from previous work (e.g., as extracted in the survey presented in Section 1.2). Depending on a given application space and resource constraints, waiting on the order of minutes may or may not be reasonable when re-writing of DNN weights is required. However, in sensor nodes, mobile devices, and other constraint-driven devices in which DNN inference is a key workload, periodic down-time for synchronization and charging may be permissible.

The asymmetric costs of eNVM writes and reads are further discussed, and another key approach is to identify a subset of application data that will be less-fequently-written, but frequently read, to maximize the benefits of on-chip eNVM storage and data reuse. This general strategy is probed

in order to optimize support for multi-task inference in both image-processing (Section 3.2) and natural language processing (Secion 3.3) domains in follow-on works to MaxNVM.

3.1.7 MAXNVM SUMMARY

Table 3.1 provides the energy, area, and performance per model. The results emphasize the need for a comprehensive co-design methodology. Considering the storage density of MLC and energy advantages due to non-volatility, even unoptimized DNNs or other reasonably fault-tolerant applications may leverage these memories. However, choosing the optimal encoding strategy, MLC configuration, and whether to include error mitigation techniques varies between models and eNVM technologies. Thus, a rigorous evaluation of the design space per DNN, as our methodology effectively executes, maximizes efficiency gains.

3.2 MEMTI: Optimizing eNVM for Visual Multi-Task Inference

Embedded non-volatile memories (eNVMs) are under active consideration to provide higher density than SRAM and ameliorate the need for power-hungry DRAM storage. However, the benefits of eNVMs come at the cost of larger write energy and write latency, as exposed and explored in Section 3.1. Moreover, limited eNVM write endurance is an obstacle to the adoption of certain technologies if DNN parameter values require frequent updates, as discussion in Chapter 1. For instance, embedded devices for robotics or augmented reality applications often required a combination of multiple inference tasks, including image classification, object detection, and action recognition. These cases highlight the need for scalable solutions that can flexibly accommodate DNN parameters for multiple tasks.

MEMTI⁵⁹ presented a DNN model and memory co-design solution that leverages a machine learning technique to reduce eNVM writes, while enabling systems to efficiently perform multiple

inference tasks. Maximizing the reuse of the learned parameters across different DNN-dependent vision tasks without re-training enforces the assumption of infrequent writes: parameters shared by multiple tasks are trained and written only once, and therefore are highly suitable for eNVM storage; in contrast, the remaining parameters can be re-trained to accommodate new inference tasks, and stored in SRAM. In addition to the storage density benefits, we evaluate how the process of retraining specific parameters can be used to recover from accuracy loss due to the adoption of denser, fault-prone multi-level eNVM storage. To the best of our knowledge, this was the first attempt to co-design non-volatile memories and DNN models through multi-task learning approaches.

3.2.0 Transfer Learning for DNN and memory co-design

Generalizing deep learning architectures to enable different application domains and more varied inference tasks serves as a way of supporting more powerful and versatile models. For example, prior work in the machine learning community 112 combined several building blocks for translation, speech, and visual inference that can be trained on all desired tasks simultaneously or on each task separately. In either case, however, introducing new inference tasks would require updating the entire set of model parameters. Other works have leveraged the concept of transfer learning to improve the performance of a single DNN on different datasets 260. These approaches are based on the observation that many visual inference tasks share low-level features, such as edge and shape detection, in the front-end layers, and become more task-specific as the computation moves closer to the classification layers. However, in order to preserve inference accuracy, transfer learning approaches either share only a limited number of front-end layers or fine-tune parameters by re-training the transferred features from one inference task to another. A recent proposal applies transfer learning to create a synthesizable fixed-parameter feature extractor 244. However, hard-wiring the feature extractor in logic prevents any future capability of fine-tuning the parameters, limiting the amount of cross-task weight sharing and flexibility.

Dataset	cifar 100	aircraft	daimlerpedcls	gtsrb	ucf101
# images	50K	7K	30K	40K	9K
# classes	100	100	2	43	101
Full model	72.78%	40.98%	99.88%	99.97%	73.77%
Only adapters	79.61%	43.8%	99.51%	99.94%	73.16%
Parameters overhead	10.4%	10.4%	10.1%	10.2%	10.4%
Training speed-up	$4 \times$	$2 \times$	1.35×	3.23×	4.74×

Table 3.3: Summary of dataset characteristics, and maximum training accuracy for the model trained entirely from scratch on each dataset or using residual adapters on a pre-trained network. Pre-trained shared parameters on ImageNet, with 67.65% accuracy, taken from ⁵⁹.

While all these techniques enable a single DNN model to perform different inference tasks, they still require updating a considerable portion of parameters to achieve maximum adaptation. MEMTI pursued a specific transfer learning technique for which the learned parameters can be generalized across multiple vision inference tasks by maximizing DNN parameter reuse and enabling efficient inference on embedded devices. The high degree of DNN parameters reuse reduces memory traffic requirements, which makes non-volatile memories a compelling solution for retaining shared parameters on-chip without incurring costs associated with frequent memory writes.

3.2.1 MULTI-TASK LEARNING MODEL

Our design is based on the DNN architecture presented in 195 , which uses residual adapter modules as a way to parameterize a generic ResNet network. These parametric modules are themselves residual blocks which use 1×1 filters and skip connection. In this setting, the number of domain-specific parameters, which comprises adapter filters, batch normalization, and fully-connected classifier parameters, can be reduced to roughly 10% of the total model size.

For our experiments, we integrate the residual adapter modules in a ResNet26 network. The baseline network is pre-trained on ImageNet, which is standard practice in transfer learning and model fine-tuning techniques. The pre-trained version for ImageNet achieves top-1 accuracy of 67.65%. The ResNet26 weight parameters obtained during pre-training are the backbone of this

multi-task inference system as they are reused for running inference on any additional visual task. The degree of adaptation is tested against five datasets selected to be representative of popular image processing tasks including classification (cifar100, aircraft), object detection (German Traffic Signs, Daimler pedestrian classification), and action recognition (UCF101 Dynamic Images).

Table 3.3 summarizes the best accuracy in the case of the model being either trained entirely from scratch or only for the task-specific parameters. As anticipated, for all datasets, the adapters overhead is around 10%. The accuracy of the network trained using adapters is always better than or comparable to training the entire network independently for each dataset. In addition, we observe that the modified model converges to the best accuracy in fewer training epochs, which results in training speed-up reported in Table 3.3.

3.2.2 Non-volatile Memory Technologies for Multi-Task Inference

The landscape of non-volatile memories includes a wide range of emerging technologies ^{56,31}, as highlighted in Chapter 1.2. These memories are generally characterized by high energy efficiency and high storage density, which can be further increased by programming multiple levels in a single cell. MEMTI focussed on a specific eNVM implementation (RRAM). Various implementations such as phase-change memories (PCM), embedded Flash (eFlash), or ferroelectric memories (FeRAM) can also be used for MLC storage. On the other hand, STT magnetic memories (STT-MRAM), while having the best write and read performance ⁵⁶, are not as suitable because compelling MLC implementations with comparable density have not been demonstrated to date, as described in Chapter 1.2.

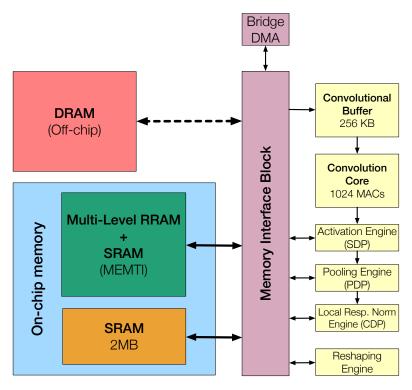


Figure 3.8: NVDLA system diagram, with additional optional interface to multi-level-cell RRAM for on-chip weight storage in a multi-task image processing setting, taken from ⁵⁹.

3.2.3 MEMORY SYSTEM OVERVIEW

In order to complement the properties of residual adapter networks and dense MLC RRAM storage, we propose MEMTI, a Memory system for Efficient Multi-Task Inference. A large fraction of the parameters in a residual adapter network is shared across multiple applications, and can be efficiently stored in MLC RRAM, while application-specific parameters can be stored in SRAM. By partitioning on-chip memory area between RRAM and SRAM, as shown in Figure 3.8, we achieve the best trade-off between storage density for the shared parameters and fast and energy-efficient updates for the task-specific parameters (SRAM). Off-chip DRAM stores multiple sets of task-specific parameters. The resulting memory hierarchy takes advantage of RRAM non-volatility for intermittent operation by powering down the system between inferences. In this scenario, only a small

portion of the model parameters must be written to SRAM during power up or task switching.

Moreover, storing task-specific parameters in more robust memory allows us to mask MLC RRAM faults via retraining.

3.2.4 EVALUATION FRAMEWORK

To evaluate the proposed memory architecture, we quantify the impact of RRAM fault characteristics and MLC encoding on inference accuracy, memory architecture and array properties, and system-level performance. The fault model is derived from previous work integrating eNVM device and circuit-level fault characteristics with DNNs evaluation frameworks to allow for extensive memory and DNN co-design space exploration ^{60,183}. We model MLC RRAM faults based on stochastic level distribution which arise from the random nature of memristors programming. When multiple levels are programmed in a single RRAM cell, the distributions overlap can be used to extrapolate the read fault probabilities for each level, as described in Chapter 2.1.

To evaluate the DNN accuracy under different storage schemes, we use a version of the residual adapter architecture implemented in PyTorch ¹⁸¹. We modified the existing implementation by adding transform functions to manipulate the values according to different multi-level encoding and compression techniques. The resulting error map is then integrated in a DNN evaluation framework to simulate the impact MLC RRAM faults on inference accuracy, an extension of ¹⁹³. The level distributions are extrapolated from measured MLC RRAM characteristics ²⁶⁸. Based on the MLC RRAM fault probabilities, we sample the value of the stored weight matrix based on a predefined multi-level encoding configuration to evaluate the impact on the model accuracy. In addition, we improve the fault model by including the effects of the sensing circuitry on the read error probability, also described in Section 2.1.

The corresponding framework is used to drive the design towards a solution that would minimize the on-chip memory footprint without increasing the inference error. After identifying the

best MLC encoding without loss in accuracy, we perform a memory design space exploration using a modified version of NVSim ⁶², as in MaxNVM ¹⁸³. The impact of off-chip memory accesses is quantified using a model of LPDDR4 DRAM. Finally, we integrate the resulting memory hierarchy with a proven CNN accelerator architecture developed by NVIDIA (NVDLA), which, combined with NVSim results and DRAM estimates, allows us to evaluate the system energy and performance for different application scenarios.

3.2.5 System-level Characterization

As shown in Figure 3.8, the baseline NVDLA system in this particular evaluation comprises a convolutional core with 1024 MAC units fed by a convolutional buffer and supplemented by several additional computational units for pooling and data transformation operations. NVDLA also supports a memory interface block and DMA that fetches model weights per layer from off-chip DRAM and leverages on-chip SRAM (2MB) to buffer inputs and intermediate results of computation between layers in the DNN. We flexibly integrate MEMTI with the NVDLA performance model as an additional memory interface to leverage for model weights, either in addition to or in place of fetching parameters from off-chip DRAM, reusing the more flexible performance model interface put forth in MaxNVM ¹⁸³.

For a competitive multi-task inference application, we evaluate the performance, energy, and area for the NVDLA system when executing three inferences per input frame using three representative visual tasks, namely image classification (cifar100), object detection (gtsrb), and action recognition (UCF101). This series of tasks computed per input frame would be appropriate, for example, for an autonomous vehicle or a drone processing sensory data to understand and interact with the surrounding environment. For this application, we set the target operating frequency to 30 frames per second (FPS), or 90 inference tasks per second, which satisfies a breadth of applications ¹⁶⁰.

3.2.6 Baseline, DRAM-Reliant Evaluation

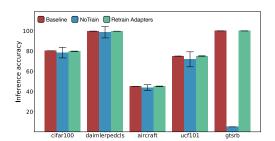
As a baseline case, we assume that the accelerator is continuously processing input frames and fetching both shared and task-specific parameters from off-chip DRAM for each layer's computation. This DRAM-only, always-on operation consumes a total power of 493mW and a peak performance of 749FPS. The estimated power includes datapath, DRAM refresh, and on-chip SRAM leakage. At this stage, the on-chip SRAM is exclusively used for storing the input features and intermediate values. We compute the energy per frame at peak performance to be 1.17mJ.

3.2.7 Provisioning for On-Chip SRAM

We first show the case in which we allocate enough on-chip SRAM to store the entire set of parameters for a single task. For a system designed to run a single inference task, having the option of storing all the network parameters on chip allows to reduce the memory access energy by $40\times$. This result demonstrates the strong impact of off-chip memory access on the entire system energy. These high energy savings are however impractical to realize with SRAM since for a 22nm technology node, we estimate a total area of at least 6.55mm^2 , over $10\times$ larger than the on-chip SRAM area for the baseline NVDLA configuration. Moreover, when we consider the full system energy in the multi-task scenario described above, the periodic parameter updates and SRAM leakage power reduce the energy savings to $0.64\times$.

3.2.8 Improving Storage Density with eNVM

As a first step towards reducing both power consumption and memory footprint we consider storing the weights on-chip using MLC RRAM. Without applying any application-level optimization, a multi-task operation would still require updating all the model parameters stored in RRAM when switching to different inference tasks. While this type of operation has a clear downside dictated



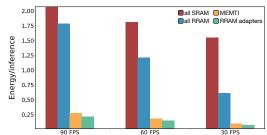


Figure 3.9: Accuracy Results for MEMTI (left) and Energy vs FPS for the different design configurations normalized to the DRAM baseline (right). The power savings of the RRAM-based design for higher frame rates is exacerbated by the frequent RRAM writes, making the design less efficient than the DRAM-based baseline. On the other hand, the energy per inference for MEMTI and RRAM adapters is strictly better than the baseline, taken from ⁵⁹.

by the RRAM write endurance, our system level evaluation exposes other limitations. Although the overall leakage power and on-chip memory area can be reduced to 298mW and 0.347mm² respectively, the energy per inference increases to 14.58mJ. This is caused by the combined effect of RRAM write energy and latency. These examples highlight the need for a solution capable of balancing on-chip memory density and write performance.

3.2.9 MEMTI FOR INTERMITTENT OPERATION

MEMTI reduces memory write costs by replacing the RRAM portion storing the task-specific parameters with SRAM. This is possible thanks to the adoption of residual adapters in the DNN network. For the resulting design, the total system power is 362mW and the energy per inference is 1.56mJ, which is comparable with the baseline. Isolating the costs associated with weight storage compounds the benefits introduced by MEMTI: power consumption decreases by $3.9\times$, with an area overhead of 1.16mm^2 . The resulting peak performance is 429FPS, well above application requirements. Intermittent operation is where MEMTI truly stands out by taking advantage of non-volatility. In this scenario, we fix the operation at 30FPS and power down the system between frames, which reduces the energy per inference by $10.65\times$, as highlighted in Figure 3.9.

	Power	Max FPS	WMem Area	Saved	On-chip	On-chip
	[mW]		[mm ²]	energy	SRAM	RRAM
all DRAM	493	749	_	1×	2MB	_
all SRAM	634	485	6.55	0.64×	8.5MB	-
all RRAM	298	47	0.347	1.62×	2MB	2.2MB
MEMTI	344	429	1.16	10.65×	2.7MB	2MB
RRAM adapters	301	396	1	13.6×	2MB	4MB

Table 3.4: Summary of power, performance and area for the four design configurations considered in this work. The energy savings are normalized to the all DRAM configuration for the intermittent multi-task operation over three tasks running at 30 FPS. On-chip RRAM shows the physical memory capacity (i.e. number of cells), taken from ⁵⁹.

3.2.10 A RRAM-BASED SPECIALIZED DESIGN

Alternatively, we consider the case specifically tailored for the three chosen tasks. Using residual adapters still reduces the weights storage requirements by a factor of 2.3×. However, based on previous results, we use SLC RRAM to store the task-specific parameters and preserve inference accuracy. Therefore, we store the entire set of parameters in MLC and SLC RRAM. Removing the need for additional SRAM reduces the overall power to 343mW, and the area to 1mm². Allocating enough memory for storing all the parameters on chip increases the energy savings compared to the baseline by 13.6×, making this design the most area and energy efficient. Nonetheless, MEMTI maintains the advantage in terms of flexibility and robustness to RRAM errors thanks to ease of reprogrammability for the task-specific parameters, for which the memory capacity is determined only by the network structure and therefore is independent from the breadth of tasks considered in a specific application.

Figure 3.9, right, shows the relationship between FPS and energy per inference normalized to the DRAM case. The all SRAM and all RRAM configurations are heavily penalized by the inability of efficiently implement a multi-task inference system. On the other hand, a co-design of the memory and DNN model using residual adapters shows much higher energy savings compared to the baseline. Table 3.4 summarizes the results at 30 FPS for the different configuration cases.

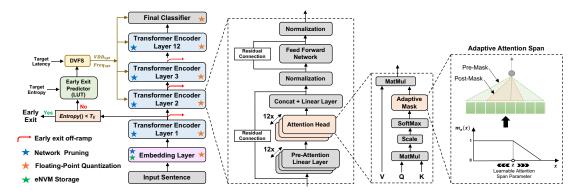


Figure 3.10: EdgeBERT design elements, a comprehensive codesign intersected with the memory density and efficiency considerations of MaxNVM to improve storage density and enable clever on-chip weight storage of ALBERT shared weights, in conjunction with several interdependent algorithmic optimizations and stategic use of dynamic voltage and frequency scaling (DVFS), taken from ²²⁶.

3.3 EDGEBERT: OPTIMIZATIONS FOR MULTI-TASK NLP INFERENCE

Transformer-based language models such as BERT provide significant accuracy improvement to a multitude of natural language processing (NLP) tasks ¹²⁶. However, their hefty computational and memory demands make them challenging to deploy to resource-constrained edge platforms with strict latency requirements. The EdgeBERT project ²²⁶ proposed and evaluated an in-depth algorithm-hardware co-design for latency-aware energy optimizations for multi-task NLP. Edge-BERT employs several optimizations with implications across the computing stack, summarized in Figure 3.10, including entropy-based early exit predication in order to perform dynamic voltage-frequency scaling (DVFS), at a sentence granularity, for minimal energy consumption while adhering to a prescribed target latency. Computation and memory footprint overheads are further alleviated by employing a calibrated combination of adaptive attention span, selective network pruning, and floating-point quantization, as a result of a reimagining of the data format, pruning, and quantization in previous presented studies with the particular characteristics and optimization goals of NLP inference tasks using the ALBERT, transformer-based DNN model ¹²⁶.

Furthermore, in order to maximize the synergistic benefits of these algorithms in always-on and

intermediate edge computing settings, EdgeBERT specialized a 12nm scalable hardware accelerator system, integrating a fast-switching low-dropout voltage regulator (LDO), an all-digital phase-locked loop (ADPLL), as well as, high-density embedded non-volatile memories (eNVMs) wherein the sparse floating-point bit encodings of the shared multi-task parameters are carefully stored. Altogether, latency-aware multi-task NLP inference acceleration on the EdgeBERT hardware system generates up to $7\times$, $2.5\times$, and $53\times$ lower energy compared to the conventional inference without early stopping, the latency-unbounded early exit approach, and CUDA adaptations on an Nvidia Jetson Tegra X2 mobile GPU, respectively.

3.3.0 Co-Designing MLC Storage for Edge NLP Inference

The ALBERT model's hefty storage requirements comprise two different classes of parameters: (1) task-specific encoder weights, and (2) word embedding parameters. Word embeddings are deliberately fixed during fine-tuning and reused across different NLP tasks, making them a prime target for dense, MLC eNVM storage with less frequent writes. The EdgeBERT methodology sought to avoid the energy and latency costs of reloading the word embeddings from off-chip memory for different tasks by storing these shared parameters in embedded non-volatile memories (eNVMs). eNVM storage also enables energy-efficient intermittent computing because the embedding weights will be retained if and when the system-on-chip powers off between inferences. However, despite their compelling storage density and read characteristics, eNVMs exhibit two main drawbacks: potentially high write cost and decreased reliability ¹⁸³.

Fortunately, the word embeddings are acting as read-only parameters on-chip, which makes them highly suitable for eNVM storage, but previous work highlights the need to study the impacts of faulty, highly-dense RRAM storage on DNN task accuracy 193,60,183,59. On the other hand, encoder weights need to be updated when switching across different NLP tasks. To prevent the energy and latency degradation that would follow from updating the encoder weight values in eNVMs, we map

	SLC		MI	.C2	MLC3	
	mean	min	mean	min	mean	min
MNLI	85.44	85.44	85.44	85.44	85.42	85.25
QQP	90.77	90.77	90.77	90.77	90.75	90.61
SST-2	92.32	92.32	92.32	92.32	91.86	90.83
QNLI	89.53	89.53	89.53	89.53	88.32	53.43
Area Density (mm ² /MB)	0.28		0.08		0.04	
Read Latency (ns) 1.21		2.1	1.54		2.96	

Table 3.5: Results of fault injection simulations impact of RRAM embedding storage on overall task accuracy. SLC=single-level cell (1 bit per cell). MLC2= 2 bits per cell. MLC3 = 3 bits per cell, adapted from 226 .

the natural partition of shared and task-specific parameters to eNVMs and SRAMs, respectively 59.

This work specifically considers dense, energy-efficient Resistive RAM (RRAM) arrays as an on-chip storage solution for shared embedding parameters (see Chapter 1.2.0). RRAMs arose as a compelling memory solution for this use case for their relative maturity and demonstrated read characteristics. There is a larger design space of opportunities to be explored with other emerging MLC-capable NVMs, but it is difficult to quickly and effectively identify eNVM solutions for a particular application and system setting, which is a research issue addressed in Chapter 4.

EdgeBERT evaluates the robustness of storing the 8-bit quantized word embeddings in MLC RRAM storage. In order to quantify the trade-offs between storage density and task accuracy, we use cell characteristics of 28nm RRAM programmed with varying number of bits per cell ¹⁹³, and evaluate 100 fault injection trials per storage configuration to identify robust eNVM storage solutions. Resulting per-task accuracy and memory array characteristics are summarized in Table 3.5.

After pruning, we store non-zero compressed embedding weights using a bitmask-style sparse encoding. Previous work demonstrated that DNN weight bitmask values are vulnerable to MLC faults, so the bitmask is protectively stored in lower-risk SLC devices, while we experiment with MLC storage for the non-zero data values ¹⁸³. Table 3.5 uncovers exceptional resilience to storing word embeddings in MLC RRAM. Across many fault injection trials, we observe that MLC2

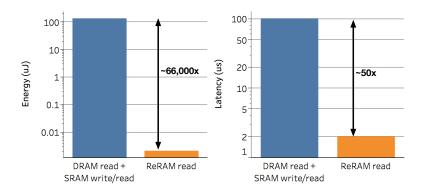


Figure 3.11: Replacing accesses to off-chip DRAM for word embedding storage with embedded RRAM drastically reduces both the memory energy (left) and latency (right) across inference tasks, taken from ²²⁶.

(RRAM programmed at 2 bits-per-cell) does not degrade accuracy across multiple tasks, while MLC3 exhibits potentially catastrophic degradation in minimum accuracy and an appreciable decline in average accuracy for the QNLI task. Based on this observation, the EdgeBERT accelerator system leverages MLC2 ReRAMs for word embedding storage.

3.3.1 RESULTS AND DISCUSSION

Fig. 3.11 illustrates the immense gains of leveraging an MLC eNVM configuration during single-batch inference after SoC power-on. In EdgeBERT, ALBERT embeddings would only need to be read from the integrated ReRAM buffers due to being statically pre-loaded. The conventional operation dictates reading the embedding weights from off-chip DRAM, then writing them to dedicated on-chip volatile SRAM memories so they can be reused for future token identifications. The EdgeBERT approach enforces a latency and energy advantage that is, respectively, 50× and 66,000× greater than the overhead costs in the conventional operation. The non-volatility of this embedded storage means that these benefits can further scale with the frequency of power cycles.

Though EdgeBERT conducted a more limited design space exploration than MaxNVM with respect to eNVM storage solutions, EdgeBERT was able to quickly and efficiently leverage the in-

tuitions and methods of MaxNVM to identify and evaluate MLC eNVM array configurations that satisfied the application and system requirements.

Even in a world of accelerating change, it is still difficult to convince people that new ways of doing things can be better and cheaper.

Grace Hopper, "Computers in the Navy", 1976

4

NVMExplorer: Cross-Stack Memory Design and Optimization

The wide adoption of data-intensive algorithms to tackle today's computational problems introduces new challenges in designing efficient computing systems to support these applications. Hardware specialization has shown potential in supporting state-of-the-art machine learning and

graph analytics across several computing platforms. However, data movement remains a major performance and energy bottleneck. As repeated memory accesses to off-chip DRAM impose an dominating energy cost, embedded memory systems demand increased on-chip storage density and energy efficiency beyond what is currently possible with SRAM.

In recent years, CMOS-compatible, embedded nonvolatile memory (eNVM) research has transitioned from articles and technical reports to manufacturing flows and product lines. These technologies hold incredible promise toward overcoming the memory wall problem. For example, one approach inspired by these new technologies combines the advantages of highly specialized architectures with the benefits of non-volatile memories by leveraging analog compute capabilities 42,201,215,50. On the other hand, the need for optimized on-chip storage solutions and memory innovation applies both to specialized hardware accelerators and for general-purpose CPU systems as well. More broadly, prior works have unveiled incredible potential improvements in storage density and energy efficiency by employing eNVMs across various architecture domains 84,102,183. With many publications showcasing the benefits of eNVM storage technologies, it is critical for system designers to be able to explore their varying capabilities and empower efficient future on-chip storage. Unfortunately, the architecture and broader research community lacks a holistic tool to quantify the system and application-level implications of memory cell technologies and to make informed decisions while navigating the vast eNVM design space.

In the past five years, consistent interest in RRAM and STT was accompanied by emerging solutions with different physical properties such as FeFET-based memories, as discussed in Chapter 1.2. Each published example offers compelling and distinct trade-offs in terms of read and write characteristics, storage density, and reliability. In addition, the space of eNVM technologies is constantly evolving with certain technologies moving out of fashion or into production. Given the fluidity and complexity of this design space, application experts and system designers need to be able to evaluate which cell technologies are most likely to provide better efficiency, higher storage density, or im-

provements on other key metrics in the context of different computing demands. Similarly, device designers and memory architects need high-level guidance to co-design their innovations toward more practical and maximally beneficial future, heterogeneous memory systems.

NVMExplorer is an end-to-end design space exploration framework that addresses key cross-stack design questions and reveals future opportunities across eNVM technologies under realistic system-level constraints, while providing a flexible interface to empower further investigations ¹⁸⁵. For example, a user can specify constraints such as target application latency, area budget, and memory access characteristics to identify a subset of eNVM proposals which are best matched for those design goals. Furthermore, the output of this framework can be analyzed to identify the limiting cell-level parameters and array characteristics for each eNVM. To enable ease-of-use and iterative studies, NVMExplorer also provides a web-based, interactive data visualizations for analyzing cell-level properties, array-level characteristics, and application-level impacts across eNVM technologies. Many of these data visualizations were made publicly available, and underlying evaluation tools are open-source ¹⁸⁵.

After describing NVMExplorer, this chapter briefly presents a related project highlighting the potential impact of effective, accessible data visualization for efficient system design and optimization ¹⁸⁴. Next, this chapter reviews several application-driven case studies to explore and analyze eNVM storage solutions for DNN inference acceleration, graph processing, and general-purpose compute. Each eNVM is viable in certain contexts, and the most compelling eNVM is dependent on application behavior, system constraints, and device-level choices. This finding suggests the existence of many possible architecture-device co-design opportunities, also including specialization of application-level constraints and characteristics, which is the focus of Section 4.4. Interestingly, the presented co-design studies reveal both opportunities and potential disconnects between the priorities of architects and device-designers in maximizing memory efficiency in a realistic system context.

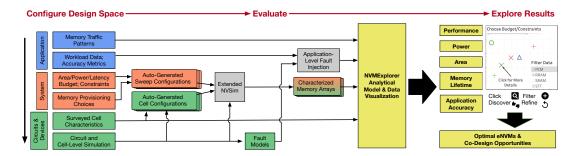


Figure 4.0: NVMExplorer Overview; There are three main stages, as discussed in Section 4.0: (1) a user configures their design space of interest, either with specific constraints and characteristics or leveraging broad, provided configurations; (2) the evaluation automatically generates corresponding configurations, runs a backend memory array simulator and/or application-level fault injection trials, and uses an analytical model to extrapolate results to application-level and system-level metrics of interest; (3) a user explores results interactively, discovering opportunities for optimization and optimal memory system configurations, as well as filtering and refining the design space, adapted from ¹⁸⁵.

4.0 An Efficient, Extensible Design Space Exploration Framework

At a high level, NVMExplorer is a comprehensive design space exploration (DSE) framework integrating application-level characteristics, system constraints, and circuit and device parameters in a publicly-available, simple-to-use flow. The overall structure of NVMExplorer relies on three stages, described in more details in the following subsections:

- A comprehensive cross-stack configuration interface to specify the design space of interest.
 This configuration spans the computing stack from application (blue) and system (orange) down to circuits and devices (green).
- 2. An evaluation engine which automatically generates configurations, simulates memory arrays, processes application behavior, computes key metrics such as performance, power, area, accuracy, and lifetime, and generates meaningful visualizations. Evaluation steps which extend existing tools are shaded grey in Fig. 4.0.
 - (a) First, a modified and extended version of NVSim extrapolates cell characteristics to memory array-level metrics

- (b) Additionally, application data and memory fault models can be connected to an application-level fault injection tool to quantify the impacts of memory faults
- (c) Then, the analytical model estimates overall dynamic power, latency, and memory lifetime as dictated by application use-case and system settings
- 3. An interactive, web-based visualization tool to aide discovering, filtering and refining eNVM design points.

4.0.0 Cross-Stack Configuration

To evaluate and compare eNVM solutions in system settings, it is not just cell or even array-level characteristics of a particular technology that matter. Rather, viable solutions depend on the area, power, and latency budget of a system and how applications running on that system interact with the memory. NVMExplorer provides a rich interface for configuring key application, system, and circuit and device parameters.

At the application level, the user inputs information about memory traffic, which may include the number of read and write operations, their proportion relative to the total number of memory accesses, and how accesses are spread out over execution time. These configuration parameters may be fixed values (e.g., characterization results of a specific workload) or provided as ranges to generate generic memory traffic patterns. With either type of application input, NVMExplorer will analytically evaluate a given memory specification over a range of application scenarios. Some applications may have additional demands or metrics which are tightly related to memory technology characteristics. For example, machine learning applications or approximate computing methods may trade-off relaxed accuracy for performance and energy, and NVMExplorer also provides an interface for designers to study the application interactions and implications of fault-prone eNVM solutions.

At the system level, the user has the freedom to evaluate a wide variety of memory configura-

tion options by either setting performance, power, and area constraints and optimization goals or by choosing memory array specifications such as capacity, multi-level programming, bank configuration, and more. The circuits and devices level of the design space configuration comprises pertechnology memory cell characteristics, in addition to sensing and programming circuitry choices. NVMExplorer also provides a database of eNVM cell configurations derived from ISSCC, IEDM, and VLSI publications, but it is also possible (and encouraged!) for users to extend the current NVMExplorer database with new simulation-based (*i.e.* SPICE or TCAD models), measured, or projected circuit and device properties. Once the full-stack specifications are set, NVMExplorer automatically generates configuration files, which are used as input to the evaluation engine.

4.0.1 EVALUATION ENGINE

NVMExplorer natively supports a wide range of design space configurations including application choices and generic memory traffic patterns, system constraints, and eNVM cell characteristics. However, the user is also free to configure their own design space. Given the auto-generated cell and system-level sweep configurations, the evaluation engine produces memory array architecture characterizations and computes application- and system-level power, performance, area, and reliability metrics. NVMExplorer combines a customized memory array simulator, an application-level fault injection tool, and an analytical model to perform these evaluations.

To characterize memory arrays, NVMExplorer relies on a customized version of NVSim, a previously validated tool to compute array-level timing, energy, and area ⁶². NVMExplorer builds on existing efforts to extend NVSim to support multi-level cells and ferroelectric-based eNVMs ^{204,183}. In addition, NVMExplorer provides a more user-friendly configuration interface to ease data collection and post-processing. We introduce the capabilities of NVMExplorer in comparing eNVMs in Section 4.0.2.

Results of cell-level and circuit-level simulations are used to parameterize fault models and per-

form application-level fault injection, as in Chapter 2. For performance estimations, in lieu of cycle-accurate simulation, NVMExplorer utilizes a long-pole, bandwidth driven model that takes memory access latency and available read/write bandwidth and compares aggregated access latency per workload execution and per second of execution to workload access statistics. This is similar in spirit to performance models in ^{66,178}, and it serves the primary purpose of identifying memory solutions that cause application slowdown, rather than predicting precise latency implications. To extract other critical application-level metrics, such as energy, NVMExplorer aggregates the read and write access energy based on the number of application accesses and array energy-per-access with the leakage power, scaling according to use-case and wake-up frequency for intermittent operation. Similarly, memory lifetime is extrapolated by applying the average reported endurance to the write access pattern per workload and the use-case.

4.0.2 EXAMPLE ARRAY-LEVEL COMPARISON

Figure 4.1 presents example array characterization output generated by NVMExplorer after evaluating various eNVM configurations implemented in a 22nm node. The design points are color-coded to highlight optimistic (green), pessimistic (red), or reference (blue) designs across surveyed publications per cell technology, as described in more detail in Section 4.0.7. The figure also reports the characteristics of 16nm SRAM as a comparison point. For each technology, we show array characterization under different optimization goals, which result in a variety of internal array architectures (e.g., bank arrangement and column muxing for read-out, as summarized in Chapter 1.0). For example, we observe a wide range for the read-energy-per-bit of an iso-capacity SRAM array. This result reflects the effect of different array optimization targets (read energy-delay product, write characteristics, area) on the internal bank configuration and periphery overhead.

This preliminary study already provides a few key takeaways. Each eNVM is able to attain read access characteristics competitive with SRAM, with the exception of an array characterized with

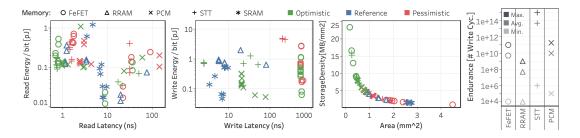


Figure 4.1: NVMExplorer example array characterization (4MB, various optimization targets), including read and write characteristics per access, storage density, and projected endurance based on the minimum, maximum, and average reported endurance across survey results in Chapter 1.2, taken from ¹⁸⁵.

pessimistic underlying PCM cell characteristics. However, write access characteristics vary dramatically across published eNVM examples, in addition to the range of reported endurance per technology. The tension between these properties and potential storage density (even in the absence of multi-level cell programming) indicates that array-level comparison in isolation may guide a system designer towards sub-optimal solutions. For example, a FeFET-based memory may seem a fitting choice for high-density, read-performant storage, but performance and energy efficiency of those memories are highly shaped by application traffic patterns and underlying cell assumptions. Thus, the cross-stack nature of data exploration supported by NVMExplorer is essential in guiding system-level choices and further investigation.

4.0.3 FAULT MODELING AND RELIABILITY STUDIES

In addition to characterizing memory performance, power, area, and lifetime, NVMExplorer extends previously validated efforts in application-level fault injection to provide an interface for fault modeling and reliability studies ¹⁹³. Users can provide an expected error rate or more detailed, technology-specific fault models and storage formats to perform fault injection trials on application data stored in different eNVMs. To quantify the impact on application-specific metrics of accuracy, the fault injection tool is tightly integrated with application libraries for data-intensive workloads,

including PyTorch for DNNs and SnapPY for graph processing ^{181,135}, as well as numpy for generic application data. As a demonstration, we perform SPICE simulation and extract fault characteristics associated with single-level vs. multi-level cell (SLC vs. MLC) programming and sensing circuitry characteristics. In this work, we consider a subset of eNVMs, namely, RRAM, CTT, and FeFET, whose fault characteristics could be derived from existing modeling efforts ^{183,204,60}. We use our extended fault injection framework to simulate the impact of storing workload data in SLCs vs. MLCs and with varying sensing circuitry designs and parameters. Armed with these additional capabilities, NVMExplorer can replicate the results of previous considerations of eNVM storage reliability ¹⁸³, in addition to providing a broader platform for studying the interactions between programming choices, cell characteristics, and application accuracy.

4.0.4 Data Visualization as a Cornerstone of Design Space Exploration

The figures in this work are snapshots from NVMExplorer's interactive web-based data visualizations, built using Tableau ²²⁵, many of which are freely available ¹⁸⁵. A screenshot of one such view is provided in Figure 4.2. In each study, results are filtered and constrained according to system optimization priorities and application use cases, as described in the text. The basic NVMExplorer data visualization dashboard presents power, performance, area, and memory lifetime results across all user-configured sweep results (e.g., many application traffic patterns, array provisioning choices, and/or eNVM cell configurations) alongside array-level metrics for a holistic design exploration experience. A user can filter results in terms of important constraints (e.g., latency or accuracy targets, power or memory area budget) and identify design points of interest. While several features of these visualizations are evident in the figures in this work, including dynamic filtering across plots, click-and-drag to narrow the design space, and pop-up details about results, the reader is encouraged to use their imagination in how they might explore and filter the data shown in alternative ways according to their interests, questions, or confusions.

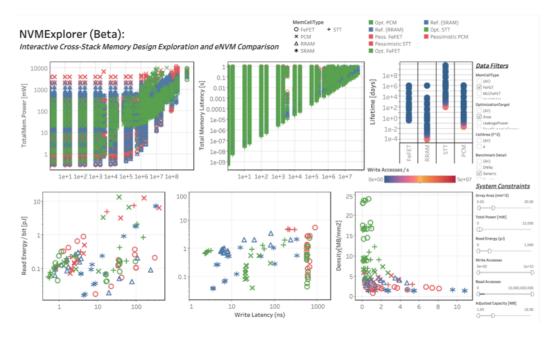


Figure 4.2: NVMExplorer dashboard summary, as appeared on live webpage in November 2021.

NVMExplorer is designed to be navigable and extensible to allow for users to conduct studies which are relevant to the design spaces they care about. A user can easily explore tweaking different device parameters or architecture constraints to see how it affects the results and metrics. This holistic approach makes application-architecture-device co-design easily accessible because it directly exposes trade-offs between device parameters and architecture metrics like performance and power for specific application settings. Using the interactive data visualization web tool, users can quickly filter results to understand results in an intuitive way and to conduct one-off studies related to a specific co-design target. Sections 4.1-4.3 explore different application-driven case studied using NVMExplorer, and Section 4.4, explores different architecture-device co-design opportunities. The application-driven case studies include a case study on eNVMs for DNN inference on IoT devices, a system agnostic graph workload study, and a study on an eNVM-based last-level cache. These case studies demonstrated some specific questions that can be answered using NVMExplorer. The co-design opportunities include both device-driven opportunities as well as architecture-driven

opportunities. The device-driven opportunities include alternative FeFET fabrication choices, trading area efficiency for increased performance, and variation in multi-level programming advantage across technologies. The architecture-driven opportunities includes intermittent operation and write buffering.

4.0.5 CHAMPVis: Comparative, Hierarchical Performance Analysis

Effective and accessible data visualization can accelerate the process of designing and optimizing efficient computing systems. In this section, I briefly present a data visualization and analysis tool I developed towards identifying performance bottlenecks in general-purpose systems called CHAM-PVis ¹⁸⁴. Additionally, I'll describe how the design and implementation of CHAMPVis informed the importance of providing an accessible, interactive simulation and data visualization framework as part of NVMExplorer.

The first step to system and algorithmic optimization is identifying performance bottlenecks. The computer systems community has made significant strides in building tools for bottleneck analysis. TopDown ²⁵⁷ records the performance impact of these events and outlines a structured, hierarchical methodology to categorize critical bottlenecks. At the highest level, modern computers are organized into three major stages: (1) the frontend, which determines what work to perform, (2) the processing unit, which computes the work, and (3) backend, which stores the final output for future use in memory. Each of these stages can be broken down into a hierarchy of sub-stages, up to 4-levels, down to the utilization of individual hardware components (e.g., floating point vector computation unit or L1 cache for a CPU core). This methodology was adopted in commercial tools including Intel VTune ¹⁰³. However, these tools only consider a single application and hardware platform at a time, making comparative analysis laborious and time-consuming. CHAMPVis is a web-based visualization platform that enables using TopDown-style hierarchical performance analysis across multiple applications.

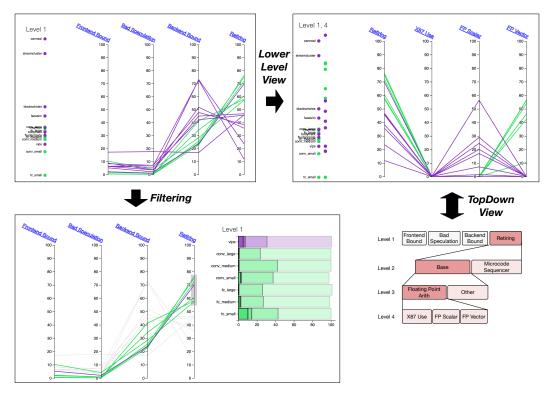


Figure 4.3: CHAMPVis representative figures; a user may flexibly explore TopDown ²⁵⁷ statistics across many workloads or hardware platforms, filtering to identify and classify performance bottlenecks at a glance, taken from ¹⁸⁴.

Supporting comparative performance analysis across applications and hardware platforms is key for datacenter-scale optimizations. Typical datacenters contain thousands of machines which are grouped into tens of machine types, and each essential application spans thousands of lines of code ⁸⁷. By analyzing a diverse set of applications or hardware platforms, performance optimizations can be applied to a wider range of inefficiencies. For example, Google and Facebook apply TopDown performance analysis within their datacenters ^{114,219}. These works led to solutions that reduced total datacenter cycles by up to 2%. However, researchers are currently limited to studying only the most abstract level in TopDown — deeper evaluations would require visualization techniques tailored to datacenter-scale hierarchical comparisons.

To enable productive, efficient, and user-friendly performance comparisons, I helped to develop

CHAMPVis, from which some representative images are presented in Figure 4.3. CHAMPVis facilitates identifying similarities and differences across different software applications and hardware platforms. CHAMPVis was developed considering several key design decisions such as how to visually relate performance metrics at different levels of the hierarchical hardware structure, how to guide users through the analysis process, and how to compare the performance of different applications and hardware platforms across multi-dimensional metrics. CHAMPVis allows users to directly interact with and extract useful details from hierarchical performance analysis data while simultaneously making effective comparisons. This closes the loop between identifying an important performance bottleneck and comparing results across software applications or hardware platforms. This will enable datacenter system engineers and software application developers to more productively optimize performance. Thinking critically about how to represent and interact with data for effective exploration and analysis in the CHAMPVis project inspired the development of openly-available and interactive data visualizations as part of the NVMExplorer methodology.

4.0.6 Comparison to, Integration with Other Tools and Frameworks

Previous work in evaluating eNVMs can be characterized as either focusing on device-level and array-level evaluations, or providing in-depth cross-stack analysis for a particular combination of eNVM and application target. Table 4.0 codifies the key differences between NVMExplorer and some related works. Survey works such as the Stanford Memory Trends ²²⁰ maintain a list of key parameters, like storage capacity and write energy, while previously validated array-level characterization tools, such as NVSim ⁶², characterize timing, energy, and area of eNVM-based memory structures. DESTINY¹⁸⁷ modifies NVSim to evaluate 3D integration.

To evaluate eNVMs in a system setting, prior work typically integrates NVSim with a system simulator. DeepNVM/DeepNVM++ 102,101 enables cross-layer modeling, optimization, and design space exploration of MRAM-based technologies in the context of GPU cache for DNNs using

		Tech. Surveys		Array Simulators		Arch-Specific Frameworks			
		IRDS 100	Trends 220	NVSim 62	Destiny 187	Neuro-Sim+34	NVMain 188	Deep-NVM++101	NVMExplorer
NVM	RRAM	√	√	√	√	✓			√
	STT		√	√	✓	✓	✓	✓	√
	SOT		√					✓	√
	PCM		√	√	√	✓	✓		√
	CTT								√
	FeRAM	√	√						√
	FeFET		√			✓			✓
Circuits	MLC					√			√
	Fault					/			,
	Modeling					v			V
Architectural Simulator / Use Case						PIM + DNNs	gem 5	GPGPU-sim + DNNs	Analytical; many included
App-Aware Evaluation	Accuracy					✓			√
	Memory Lifetime	1					✓		√
	Operating Power	1				√	✓	✓	✓
	Latency					✓	✓	✓	✓

Table 4.0: NVMExplorer leverages existing efforts by extending NVSim, while enabling cross-stack DSE across multiple use cases and domains, including more breadth than previous works, and providing a unified platform to explore and iterate design, adapted from ¹⁸⁵.

GPGPUSim. NVMain ¹⁸⁸ enables evaluation of eNVM-based main memory using gem; ²⁰. NeuroSim+³⁴ focuses on evaluation of processing-in-memory for DNN inference and training. While these frameworks are great examples of domain-specific explorations and evaluations, NVMExplorer can evaluate a variety of system and application domains, in addition to offering reliability analysis, additional metrics such as memory lifetime, and a database of technology cell characteristics and configurable device parameters.

Existing works such as these provide limited or otherwise domain-specific design space exploration frameworks. The time required to sift through non-volatile memory roadmaps, learn the growing number of NVM simulators, and piece together toolflows to answer questions about specific architectures and applications is throttling the pace of co-design between architects and device designers and inhibiting the advancement of technologically-heterogeneous memory systems. In contrast, NVMExplorer's cross-stack and analytical approach provides efficient evaluation of critical application-aware metrics such as application accuracy and resulting memory lifetime with sufficient fidelity to guide future design studies. NVMExplorer offers more breadth by including application-, system-, and device-level considerations, and accommodating a wider range of devices without requiring a separate system simulator.

NVMExplorer's modular design allows for efficient and effective integration of other underlying memory characterization and simulation tools as the research space conitnues to grow and evolve. For example, at the configuration stage, NVMExplorer can interface with profiling tools and memory trace generators to designate application traffic rates and patterns of interest. For example, memory access statistics generated using the Sniper Multi-Core Simulator ²⁶ are used as representative cache behavior in Section 4.3. Also at the configuration stage, NVMExplorer has to potential to directly interface with cell, device, and circuit simulation tools (e.g., HSPICE) to configure memory cell characteristics, programming settings, and array component properties and overheads. At the evaluation stage, NVMExplorer is sufficiently modular that the memory array-level characterization tool can be exchanged for alternative simulators to evaluate distinct memory solutions, such as 3D-stacked architectures using Destiny ¹⁸⁷ or cryogenic operation using CryoMem ¹⁶³. Both array characterization results and application-level metrics can then be harnessed at the exploration stage to guide iteration of the design space or directly as inputs to a broader simulation framework for more detailed evaluation (e.g., to parameterize a new memory system component in gem5 ²⁰, or to take advantage of existing tools for thermal analysis ⁸⁵).

4.0.7 TECHNOLOGY LANDSCAPE AND EXAMPLE CELL PARAMETERIZATION

NVMExplorer provides a broad survey of published eNVM examples, which can be parameterized so that systems experts can make meaningful, high-level comparisons across technologies despite different underlying trade-offs and maturity. We validated this approach per-technology against fabricated memory arrays ¹⁸⁵.

This work was informed by a compilation of device-level and array-level data across eNVM technologies, as summarized in Chapter 1.2. We sourced the majority of the cell-level parameters from ISSCC, IEDM, and VLSI publications and focus primarily on works from 2017-2020 to reflect the most recent range of achievable behavior per technology. Previous efforts detailed the physical

properties and limitations per technology ³¹, while NVMExplorer focuses on compiling sufficient cell-level details and leaning on existing technology models to provide a broad and practical database of cell definitions. While we hope these extracted cell definitions are helpful to the community in calibrating the current state-of-the-art, NVMExplorer is extensible as the design space continues to evolve, as demonstrated in Section 4.4.

Comparing eNVMs at varying stages of development and with varying underlying physical properties is a challenging task. The case studies in this work aim to provide high-level guidance and relative judgments about which eNVM cell technologies are worthy of further investigation under specific system and application constraints. Thus, rather than focus in on specific, physically accurate cell configurations, NVMExplorer provides the bounds of what is conceivable per eNVM technology across the full range of published recent academic work. We liken identifying and evaluating these bounds per-technology to forming the poles of a tent that encompasses the full extent of eNVM properties, so we call the extrema in terms of cell-level characteristics (i.e., smallest, lowest read energy, best retention vs. largest cell size, lowest endurance) the device-level "tentpoles". In an actively evolving technology space, this approach allows us to make meaningful classifications about which technologies are potentially adoptable solutions. These modeling choices are classified into two fixed cell configurations for applicable technologies. We validate that the "tentpoles" of the cell-level design space result in array-level characterization that provides coverage of published memory array properties, as discussed in Section 4.0.8.

For the technology classes most represented in our survey, we compute which published example has the best-case and worst-case storage density in terms of Mb/F², and this data serves as the foundation of the bounds of the cell-level design space; those points which are most and least dense across recent published examples. Any critical cell-level parameters not reported with those cell definitions are assigned values (e.g., read characteristics and programming settings) using the best (lowest power, highest efficiency) or worst (highest power, lowest efficiency) value per metric across

all other recent publications with sufficient supporting data. These best-case and worst-case technologies per class form the tentpoles of the underlying cell design space, and we label these fixed cell definitions as "optimistic" or "pessimistic" accordingly. For the purposes of the case studies presented in published works ¹⁸⁵, all array- and application-level results are produced using these fixed underlying optimistic and pessimistic cell properties, though we note that a user of NVMExplorer can draw either on these constructed, bounding example cells or on the full database of surveyed configurations, or on fully customized definitions with respect to cell size, access properties, and operating conditions (e.g.,read/write voltage, temperature). Corresponding fault models and error rates for reliability studies are extracted after optimistic vs. pessimistic cell-level properties are fixed, as discussed in one of the presented case studies.

This approach is helpful for many reasons: for one, these extremes help us answer exploratory questions about what we will likely see in the near future; secondly, comparing the best-case of one technology to the worst-case of another can help gauge less mature technologies against more mature reference points; thirdly, if such optimistic configurations are untenable or even pessimistic configurations are attractive in a specific system setting, we can build confidence for further exploration and more detailed modeling efforts without implementing and attempting to meaningfully compare many many cell definitions with insufficient data. A limitation of this methodology is that inherent trade-offs between certain parameters for a technology may not be linked (e.g., area, latency, and retention for STT); however, this amalgam of cell properties represent the full spectrum of achievable characteristics per technology, rather than specific fabricated results. As a point of additional comparison, the results shown in the following studies include a reference cell configuration for RRAM as a relatively mature eNVM, with parameters derived from a specific industry result ¹²⁸.

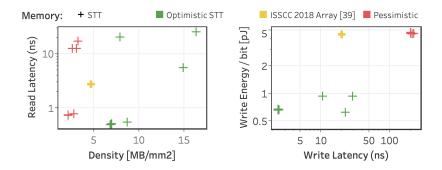


Figure 4.4: Example memory-array-level validation of reference (yellow)⁶¹ vs. optimistic (green) and pessimistic (red) underlying cell 'tentpole' configurations; metrics of a fabricated STT array from the literature⁶¹ fall within the bounds of those generated using our device-level 'tentpole' characterization, taken from ¹⁸⁵.

4.0.8 VALIDATION

Our array-level area, energy, and latency characterizations rely on the previously-validated procedures of NVSim to extrapolate cell-level configurations and array design constraints to optimized memory layouts and properties ⁶². However, in employing our "tentpole" approach, it is critical that we verify that array-level results using our optimistic and pessimistic underlying cell characteristics fully cover and match expectations of existing manufactured and published eNVM solutions.

Whenever possible, we select publications with array-level characterizations for a given technology, and compare those results to iso-capacity memory arrays modeled through our "tentpole" approach. Figure 4.4 shows an example of such an exercise. We compare a 1MB STT-RAM array published at ISSCC in 2018 ⁶¹ to optimistic and pessimistic STT design points produced by NVM-Explorer. Here, we note that our tentpole results effectively represent the range of actual array properties by producing metrics that are both higher and lower, but similar in magnitude, to the reference STT-RAM array. The studies presented in this work consider only validated configurations for which we were able to either complete this validation exercise or run SPICE-level simulations. It is worth noting that NVMExplorer is set up to evaluate all cell technologies in Table 1.1 (e.g., though SOT is a compelling emerging solution, our survey found insufficient array-level data for validation,

so it is omitted in this Chapter's evaluations). System validation and application characteristics are derived from existing, state-of-the-art references.

4.1 Supporting DNN Inference Under Varying Operating Conditions

Prior studies have demonstrated the potential benefits of eNVM storage for Deep Neural Network (DNN) inference accelerators ^{183,59,239}, albeit with limited scope in terms of eNVM technologies and cross-stack parameters considered. NVMExplorer empowers researchers to approach a broader set of questions that compare eNVMs in different storage scenarios (e.g., limited to weights vs. storage of DNN parameters and intermediate results) and system constraints (e.g., strict area budget, or power budget). In this section, we consider two distinct use cases for a DNN inference accelerator: continuous operation, as in image processing per frame of a streamed video input, and intermittent operation, where the system is woken up per inference task and can leverage the non-volatility of eNVM by retaining DNN parameters on-chip in power-off state between inferences.

4.1.0 CONTINUOUS OPERATION

We consider the commonly-used and well-studied NVDLA²¹² as a base computing platform (as in Chapter 3.1) and compare its 2MB SRAM with iso-capacity eNVMs. We use the NVDLA performance model ¹⁷⁸ to extract realistic memory access patterns and bandwidth requirements of the on-chip buffer. More specifically, we evaluate the power and performance of accesses to on-chip memory storing ResNet26 weights for single-task image classification using the ImageNet dataset vs. multi-task image processing, comprising object detection, tracking, and classification, at a consistent frame rate of 60 frames-per-second, as is typical for HD video. We additionally consider the impact of storing activations in eNVM, but this ostensibly ignores endurance limitations.

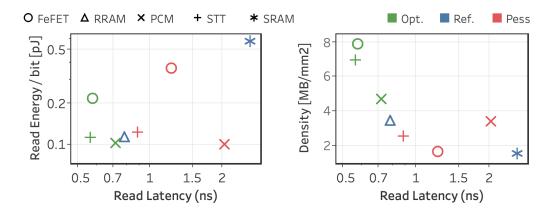


Figure 4.5: Read characteristics and storage density for 2MB arrays, provisioned to replace on-chip SRAM for NVDLA, taken from ¹⁸⁵.

4.1.1 ARRAY CHARACTERISTICS

First, we observe the read and storage density characteristics for 2MB arrays using the cell-level tent-poles of several promising eNVM technology classes, as shown in Figure 4.5 compared with SRAM. Notice that read energy effectively divides arrays into two tiers. STT, PCM, and RRAM offer lower read energies and competitive read latencies vs. SRAM. In contrast, FeFET-based eNVMs suffer from higher read energies, but optimistic FeFET offers the highest storage density with low latency. At similar low latency, optimistic STT offers 6× higher density over SRAM. PCM and RRAM outperform SRAM in terms of both read latency and storage density. While such comparative insights can readily be extracted from this pair of plots, there are other important dimensions to also consider, and NVMExplorer facilitates more comprehensive analyses that consider the impact of application priorities and system-level use cases on eNVM design decisions.

4.1.2 APPLICATION-LEVEL METRICS

Figure 4.6, left, summarizes total operating power (both dynamic access and leakage power) for the 2MB memory arrays characterized in Figure 4.5 and accessed according to traffic patterns of dif-



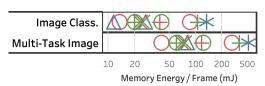


Figure 4.6: The most energy-efficient eNVM varies under different DNN inference use cases, such as continuous (left, operating power) vs. intermittent (right, reporting energy per input image frame); these results exclude eNVM solutions that are unable to meet application latency and accuracy targets, taken from ¹⁸⁵.

ferent ResNet deployment scenarios, i.e., single- vs. multi-task and weights-only vs. storing both weights and activations. These results exclude eNVM candidates that cannot support 60 FPS operation nor maintain DNN accuracy targets. Recall NVMExplorer includes fault injection wherein high eNVM fault rates can degrade model accuracy to unacceptable levels. While not explicitly shown here, NVMExplorer exposes numerous additional interactions for users to probe and build intuition. For example, while total memory power increases as the number of accesses per frame increases to compute multiple tasks, the ratio of read-to-write traffic stays roughly the same. Hence, the relative power of eNVM arrays also remains similar. In particular, PCM, RRAM, and STT all offer over 4× reduction in total memory power over SRAM. One important reason for this is that SRAM leakage power will dominate compared to eNVM solutions, even under high traffic. Of the energy-efficient solutions, STT offers best performance (lowest application latency per frame). In contrast, optimistic FeFET offers higher storage density while maintaining 6oFPS and a 1.5-3× power advantage over SRAM.

4.1.3 Intermittent Operation

Let us now consider eNVM storage for two additional use cases that alter system-level optimization goals and corresponding eNVM selection, further highlighting the flexibility and ease of exploration

the NVMExplorer framework offers. A major advantage of storing DNN weights in eNVMs is that non-volatility supports intermittent operation that powers off the accelerator between inferences. Using SRAMs would either consume leakage power to keep the weights memory powered on or consume power to restore the weights from off-chip memory, e.g., by incurring a latency and energy penalty by fetching from DRAM. In this use case, we provision monolithic eNVM storage to hold all DNN weights (e.g., up to 32MB for Natural Language Processing (NLP) tasks). For image processing, all weight memory accesses are to eNVM, eliminating the wake-up latency and power associated with loading parameters on-chip, in addition to reducing distance between compute system and higher-capacity memory. Previous work demonstrated that careful optimization between DNN properties and MLC eNVM storage can further increase storage density ^{183,204}.

Figure 4.6, right, compares the resulting memory-energy-per-inference across eNVMs for both single-task image classification and multi-task image processing, as determined by the total number of accesses to retrieve all DNN weights over the course of processing one input frame. The lowest-energy technology choice differs between the single vs. multi-task inference and, perhaps more interesting, both are eNVM candidates with *lower* storage density (RRAM and pessimistic FeFET), as opposed to the highest density options (STT and optimistic FeFET), which hints at a cross-stack prioritization of read performance as opposed to cell size reduction, as probed further in Section 4.4. We repeat this study for single task vs. multi-task natural language processing using the ALBERT network, a relatively small-footprint, high-accuracy, transformer-based DNN ¹²⁶.

We find that frequency of wake-up and specific target application traffic patterns play a critical role in selecting preferred eNVM candidates. To further study this result, we dig into the implications of intermittent operation and compare the total energy versus the number of inferences per day, showing a continuum of wake-up frequency that may arise (e.g., deployed solar-powered agricultural sensors or satellites, or a voice-enabled assistant executing NLP tasks on wake-up). The left plot of Figure 4.7 shows total memory energy as a function of inferences per day for image classi-

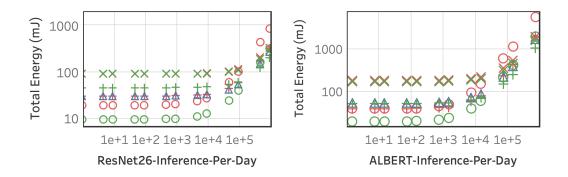


Figure 4.7: The eNVM storage solution (iso-capacity arrays provisioned per task, optimized for ReadEDP) that minimizes total memory energy consumption varies according to system wake-up frequency and DNN inference task; All solutions shown maintain application accuracy and a < 1s latency per inference, taken from ¹⁸⁵.

fication. Here, total memory energy is presented as a proxy for device battery life. From the figure, we observe that when the number of inferences per day is sufficiently low (less than 1e5), optimistic FeFET yields the lowest energy. At higher wake-up frequency, optimistic STTs take over because of the relatively lower energy-per-access. Figure 4.7, right, investigates the impact on an NLP workload. While results are similar, optimistic STT emerges as the best technology at lower inference rates (as compared to image classification), because because sentence classification is a particularly intensive task, requires higher operating power than ResNet26.

Table 4.1 summarizes the preferred eNVM technology across different use cases and tasks, with "Opt. eNVM" indicating the preferred choice under optimistic underlying cell characteristics and "Alt. eNVM" indicating the preferred technology assuming pessimistic assumptions and reference points, and table entries for intermittent operation are selected at a fixed wake-up rate. As shown by these results, non-volatility of on-chip storage resources is particularly compelling for resource-constrained systems that experience intermittent power supply (e.g., deployed solar-powered agricultural sensors or satellites) or otherwise operate intermittently (e.g., a voice-enabled assistant executing NLP tasks on wake-up). Across a range of device wake-up frequencies and per-wake-up compute patterns, we observe that several eNVMs become compelling, and the preferred NVM

Use Case	Inference Task	Data Storage	Priority	Opt. eNVM	Alt. eNVM
		Weights Only	Low Power	PCM	PCM
	Single-Task	weights Only	High Density	FeFET	CTT
	Image Classification	Weights + Acts	Low Power	PCM	RRAM
Continuous		weights + Acts	High Density	STT	RRAM
(6oIPS)		Weights Only	Low Power	PCM	RRAM
	Multi-Task	weights Only	High Density	FeFET	CTT
	Image Processing	Weights + Acts	Low Power	STT	RRAM
		weights + fiets	High Density	STT	RRAM
	Single-Task	Weights Only	Low Energy/Inf	RRAM	RRAM
	Image Classification	weights Only	High Density	FeFET	CTT
	Multi-Task	Weights Only	Low Energy/Inf	FeFET	FeFET
	Image Processing	weights Only	High Density	FeFET	CTT
Intermittent		Embeddings Only	Low Energy/Inf	RRAM	RRAM
(1IPS)	Sentence Classification	Linbeddings Only	High Density	FeFET	CTT
	(ALBERT)	All Weights	Low Energy/Inf	STT	RRAM
		Till Weights	High Density	FeFET	CTT
	Multi-Task NLP	All Weights	Low Energy/Inf	STT	RRAM
	(ALBERT)	7111 Weights	High Density	FeFET	CTT

Table 4.1: Summary of preferred eNVM under varying DNN use case, task, storage strategy, and optimization priority. Several distinct eNVM solutions are best suited in different cases under either optimistic underlying cell characteristics ("Opt eNVM") or considering only reference and pessimistic configurations ("Alt eNVM"), taken from ¹⁸⁵.

choice for further investigation varies depending on both of these factors.

4.2 Complementing and Accelerating Graph Processing

Our second case study explores the potential benefits of using eNVMs for graph processing, which imposes an entirely different set of constraints in terms of memory read and write characteristics. Graph processing comprises many read-dominated tasks with less predictable data reuse than DNNs (e.g., search kernels), but still involves write traffic and, overall, is incredibly data-intensive in terms of memory bandwidth and capacity. As an initial exploration of compatibility and viability between graph processing workloads and eNVM storage solutions, we consider the total power and resulting memory lifetime per technology under generic traffic patterns covering the range of read and write bandwidths for critical graph tasks, as described in previous workload characterization efforts ¹⁷.

As a proof of concept in a specific system, we additionally evaluate eNVM storage solutions under

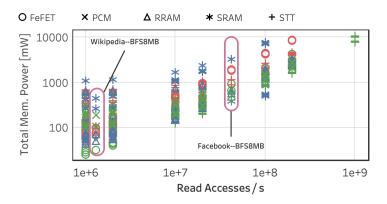


Figure 4.8: Memory power for traffic patterns encompassing graph processing demands, including specific graph kernels as labeled. The lowest power solution depends on the expected read traffic, taken from ¹⁸⁵.

access patterns for benchmarks executed on a domain-specific accelerator 83.

4.2.0 Analysis for generic traffic patterns

We consider different memories experiencing a range of generic traffic patterns representing graph processing kernels (i.e., read access rates from 1-10GB/s and write access rates from 1-100MB/s)¹⁷. NVMExplorer provides a wide array of critical metrics to compare and user-configurable visualizations to extract important trends and limitations. For example, in Figures 4.8 and 4.9, we choose to display total memory power against read traffic, as number of read accesses becomes a dominant factor in total power for read-dominated workloads, and total memory latency against write traffic, as overall performance for several eNVMs is strongly determined by write traffic.

4.2.1 POWER

As shown by Figure 4.8, total memory power generally increases with read access rate and the lowest power solution depends on the application traffic load. For applications that exhibit fewer than 10^7 read accesses per second, optimistic FeFET is a clear winner, while pessimistic FeFET and RRAM are next best candidates. On the other hand, for higher rates of read traffic (e.g., $> 10^8$), optimistic

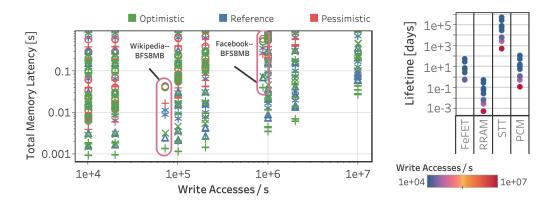


Figure 4.9: Memory latency and projected lifetime for traffic patterns encompassing graph processing demands, including specific graph kernels as labeled. FeFET solutions fail to match SRAM performance. STT provides superior performance and memory lifetime, taken from ¹⁸⁵.

STT is best. For mid-range read access rates, PCM and RRAM are also viable solutions sometimes offering the lowest power solution. However, this relationship alone does not dictate memory technology choice.

4.2.2 LATENCY

A slightly different and more consistent story emerges when we analyze the impact of different eN-VMs on overall memory latency (both read and write) versus write access rates, shown by Figure 4.9. While there is a clear preference for optimistic STT, RRAM and optimistic PCM are also worth considering. In contrast, most pessimistic eNVM technologies and all FeFET-based solutions are significantly inferior, even failing to match SRAM performance for many traffic patterns. When we additionally consider projected memory lifetime, STT emerges the clear winner overall. Note that the right chart of Figure 4.9 plots the memory lifetime assuming continuous operation at a particular write access rate. Hence, the highest write traffic always yields the lowest lifetime. While RRAM seemed promising based on performance and power, it has the worst endurance and lowest lifetimes.

4.2.3 Analysis for domain-specific systems

In addition to relying on generic traffic to represent the full range of expected load of graph processing, NVMExplorer can also be leveraged to answer a more specific design question: For performance targets and traffic patterns to a specific storage resource in a graph processing accelerator system, which eNVMs offer compelling characteristics that warrant further investigation? To this end, Figures 4.8 and 4.9 also include points, identified in pink, corresponding to memory traffic to run breadth-first search on two different social network graphs ¹³⁵. Traffic patterns are extracted from throughput and accesses reported for the compute stream of a domain-specific graph processing accelerator utilizing an 8MB eDRAM scratchpad ⁸³. In the baseline system, about 90% of the energy is spent on the eDRAM scratchpad (not including DRAM controller energy), with an operating power of at least 3.1W at the 32nm process technology node as reported from Cacti ^{83,91}. We analyze the benefits of replacing the 8MB eDRAM scratchpad with an iso-capacity eNVM array provisioned to meet the cited latency target (1.5ns).

If we exclude RRAM due to low lifetime projections, FeFET, PCM, and STT all offer significantly lower memory power (about 2-10× lower than SRAM) and even pessimistic STT offers consistent performance. These observations, based on a realistic graph processing use case extracted from prior work, are consistent with the results generated using generic traffic patterns. Again, optimal technology choice depends on higher, system-level optimization goals, and NVMExplorer provides critical insights in the presence or absence of a specific system simulator results.

If the high-level goal is to maximize storage density, FeFET is highly attractive, but severely limited by poor write latency (unable to meet application latency expectations under the higher range of traffic patterns). Rather than prematurely eliminating FeFET, designers can leverage NVMExplorer to study the impact of relaxing or adapting application targets or to explore co-design solutions that target improvements to the underlying technology or architecture, as in Section 4.4.0.

4.3 Probing General-Purpose Applications: eNVM as LLC

Improved density and energy efficiency could revolutionize general-purpose on-chip storage, and recent efforts have endeavored to replace high-performance memories, like caches, with eNVM-based alternatives ^{123,84,101}. However, caches must handle a large volume of writes depending on the application, so the achievable write latency and endurance per eNVM comes to the forefront of design considerations. While the improved density and energy efficiency of eNVMs could revolutionize general-purpose on-chip storage, the open question of achievable endurance and write access characteristics per technology cannot be overlooked and needs to be a primary factor in determining a cache technology replacement.

In this study, we consider the last-level cache (LLC) of a high-performance desktop processor, similar to Intel's 14nm, 8-core Skylake. The memory hierarchy includes a private 32 KiB L1I\$; a private 32 KiB L1D\$; a private 512 KiB L2\$ (non-inclusive, write-back); and a shared ring 16MiB L3\$ with 64 B line, 16 ways (inclusive and write-back). The system includes DRAM with 2 channels, 8 B/cycle/channel, 42cycles + 51 ns latency. Representative application behavior comes from SPECrate CPU2017 (integer and floating point), and we warm-up the cache for 500M instructions and simulate for 1-billion instructions in detail using the Sniper simulator 24,26. This provides application modeling data for a 16MB LLC (e.g., reads, writes, execution time per benchmark) that are inputs to the application configuration interface of NVMExplorer. When replacing the last level cache with an NVM, we use an iso-capacity configuration. Another possibility is to fill up the same amount of physical area that an equivalent SRAM-based last level cache would consume (which would lead to significant capacity advantage for NVMs).

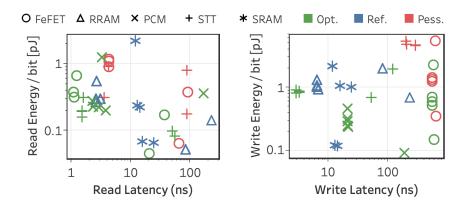


Figure 4.10: Array access characteristics in for consideration of replacing (iso-capacity) a 16MB LLC, taken from ¹⁸⁵.

4.3.0 ARRAY CHARACTERISTICS

First we focus on the array characteristics of the different memory technologies in isolation, as shown in Figure 4.10. From the left plot, we note a competitive range of read energy and read latency does not reveal a clear winner. For example, if read energy per access is highest priority, FeFET, RRAM or even SRAM offer array configurations that trade access latency for energy efficient, while STT and optimistic FeFET offer pareto-optimal read characteristics. For writes, a PCM-based last level cache appears to minimize energy per access. On the other hand, only STT and RRAM are able to beat SRAM write latency. Again, we find array characteristics in isolation do not offer sufficient guidance to choose the best eNVM for LLC, and NVMExplorer allows us to go further. In terms of storage density, STT shows over an order of magnitude reduction in area compared to SRAM (from about $10mm^2$ to about $1mm^2$ across optimization targets), though under pessimistic underlying cell characteristics, we note that PCM or RRAM offers higher density that STT.

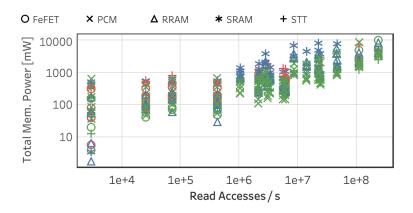


Figure 4.11: Memory operating power under continuous operation across SPEC benchmark traffic to a 16MB LLC shows preferred eNVM depends on traffic demands and optimization goal. All solutions shown meet per-benchmark read/write demands. For high-traffic benchmarks, STT provides lowest power, lowest latency, and longest projected lifetime, taken from ¹⁸⁵.

4.3.1 Power, Performance, and Lifetime

Now we start to focus on the effects of the actual application behavior on array characteristics in the context of the last level cache. It is, therefore, important to also consider system-level application behavior. Figures 4.11 and 4.12 show the resulting power, performance, and lifetime when using different eNVMs as LLC and assuming memory traffic from SPEC2017 benchmarks. Figure 4.11 shows total memory power versus read access rate, where each column of points corresponds to a particular benchmark traffic pattern. We again see that the lowest power eNVM solution depends on the traffic pattern. In broad terms, RRAM and FeFET fair better for lower read access rates while PCM is better for higher rates until STT emerges best for the highest rates. In terms of memory access latency (Figure 4.12) with respect to write access rates, STT is usually the best choice, though arrays unable to meet application bandwidth are excluded. Lastly, Figure 4.12, right, compares lifetimes across the eNVM technologies for a range of write access rates. Again, STT offers the best longevity on average. However, depending on the use case of your desktop machine, PCM and FeFET may also be compelling options (e.g., if read-dominated workloads such as xalancbmk

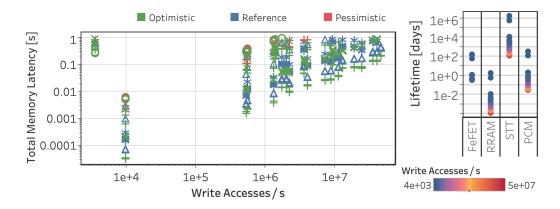


Figure 4.12: Memory latency and projected lifetime under continuous operation across SPEC benchmark traffic to a 16MB LLC shows preferred eNVM depends on traffic demands and optimization goal. All solutions shown meet perbenchmark read/write demands. For high-traffic benchmarks, STT provides lowest power, lowest latency, and longest projected lifetime, taken from ¹⁸⁵.

dominate an end user's desktop machine, rather than more write-intensive scientific computing workloads). RRAM, on the other hand, does not appear viable as an LLC.

Interestingly, the winning technology is dependent on the read accesses per second seen by the last level cache only when they are sufficiently low. Even more so is the fact that it is actually the pessimistic STT technology which outperforms all others for the case where the read accesses per second is 10⁴. For higher values of read accesses per second, the optimistic version of STT is the clear winner. In terms of performance, total memory latency is consistently lowest for the optimistic STT regardless of number of write accesses per second.

4.4 Case studies in eNVM, System, and Application Co-Design

Each scenario presented so far in this chapter presented unique optimization goals and system priorities and, in each case, we compare how each eNVM's power, performance, and area fairs relative to similarly-provisioned SRAM or DRAM in a baseline system. The first case study showed that each eNVM candidate has distinct advantages as scratchpad memory for a DNN accelerator, depending

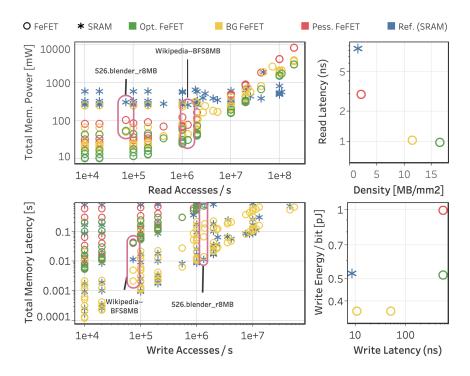


Figure 4.13: Back-gated (BG) FeFETs provide the high density and low operating power required by graph processing benchmarks while maintaining SRAM-comparable performance, and these solutions begin to close the performance gap between non-BG FeFET and other memory technologies across SPEC2017 benchmarks, taken from ¹⁸⁵.

on task and use cases (Table 4.1). The second case study explored – in a system-agnostic manner – promising eNVMs across expected memory traffic of graph analytics tasks and reveals that STT, PCM, and FeFET memories are all compelling solutions depending on guiding optimization target. Third, STTs emerge as the most attractive candidate to replace SRAMs in the LLC of a modern desktop CPU, consistent with prior studies 84,98,7.

NVMExplorer enables systematic studies that traverse the immense design space and provide navigable results, which can be progressively refined down to the most promising eNVM candidates. NVMExplorer's outputs offer insights into fundamental limitations per eNVM that may either be addressed by innovations at the device level or architectural solutions, and such co-design opportunities are the focus of the following discussions.

4.4.0 Device-Driven Co-Design of FeFET-based eNVM

Previous FeFET-based device characterization and modeling efforts have exhibited write pulses on the order of 100*ns*-1*µs*. However, alternative FeFET fabrication strategies in early development stages, such as back-gated FeFETs²⁰⁵, offer compelling potential advancements in write latency (10*ns* programming pulse) and projected endurance (10¹²). Section 4.2 and Chapter 2.4 noted that a primary limitation of FeFETs in the context of graph processing is an inability to meet the application latency targets under higher write traffic. Thus, using the underlying cell properties of back-gated FeFETs reported in ²⁰⁵, we can rapidly re-examine the viability of FeFET-based memory and probe whether this change could make a difference in the viability of FeFET-based memory for graph processing and other workloads of interest.

Figure 4.13 shows the total memory power and total memory latency of an 8MB memory array of back-gated FeFETs (in yellow) compared to using previous FeFET models (red, green) and SRAM (blue). We examine these metrics under a range of read and write traffic patterns which are inclusive of the graph benchmarks described in Section 4.2 and the SPEC benchmarks used in Section 4.3, but here showing access patterns for an 8MB capacity LLC. Three different array organizations are tested which represent three different optimization targets: read latency, read energy, and read energy-delay-product (EDP). The underlying array-level characterization is shown in Figure 4.13, right. From the array characterization, we observe that the back-gated FeFETs show a slight increase in read energy per access and slight decrease in storage density compared to prior state-of-the-art cells. However, we observe that they enable comparable application latency to SRAM across a wide range of write traffic where previous FeFET versions fall short. Furthermore, back-gated FeFETs results in the lowest operating power over most of the range of read accesses per second, including for the example graph processing benchmark, Wikipedia–BFS8MB.

Based on these observations, we posit that back-gated FeFET memory may close the performance

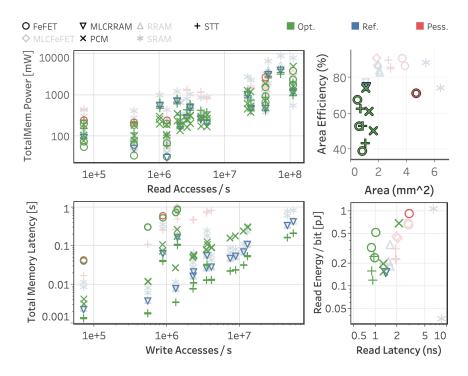


Figure 4.14: Results for 8MB arrays are filtered according to a maximum area efficiency (top right). Arrays with lower area efficiency are highlighted across all views and tend to result in low memory latency and power across many traffic scenarios, taken from ¹⁸⁵.

gap between prior FeFETs and other memory technologies (including SRAM) and unlock additional application domains. NVMExplorer's ability both to quickly and efficiently gauge the impact of cell-level innovations and to match emerging device designs to compelling use cases can enable productive future co-design collaborations. This feedback loop is mutually beneficial in providing direct motivation for further device development and encouraging system designers to integrate more energy-efficient, highly dense on-chip memory.

4.4.1 TRADE AREA EFFICIENCY FOR PERFORMANCE

One theme we can highlight across the architecture-driven case studies in Sections 4.1-4.3 is that the subset of characterized results that exhibit lower area efficiency (i.e., internal array architectures

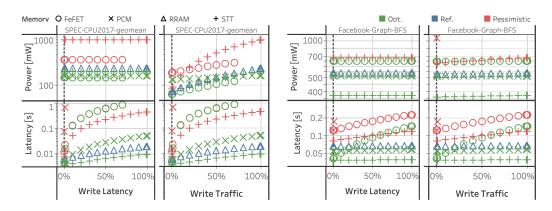


Figure 4.15: Masking write latency or reducing write traffic via introduction of a write caching scheme could make a broader set of eNVM technologies viable, taken from ¹⁸⁵.

that do less amortization of periphery and sensing overhead) also tend to result in lower total memory latency across many traffic scenarios. This is perhaps counter-intuitive given the effort spent in the devices community to manufacture very small cell sizes. We also note that in Figure 4.14, where such design points are highlighted across the plots, that slight advantages in terms of energy-peraccess (e.g., Opt. STT and PCM compared to FeFET) tend to correlate to large total power advantages in high-traffic scenarios. As such, pointing out to device designers the greater relative impact of reduced energy per access rather than decreased cell size could usher in a more productive, product-ready set of eNVM technologies. Additionally, we observe that reducing energy per write access for STT and RRAM would drastically improve their relative power advantage for data-intensive applications, even at a cost of relatively lower area efficiency or storage density.

4.4.2 Write buffering changes the performance landscape

In conjunction with technology innovations to reduce write latency, adoption of a wider set of eN-VMs in general-purpose computing contexts could be made possible by employing existing architectural techniques to mask poor write characteristics. For example, in an effort to extend memory lifetime and mask the performance impact of write access, a more performant technology (e.g., SRAM,

or STT) could be employed as a write-buffer. Rather than employ a costly and engineering-intensive cycle-accurate simulator to gauge the impact of provisioning a write buffer, NVMExplorer enables an analytical study under user-specified traffic patterns to narrow the space of eNVMs worthy of further simulation and design effort. This approach answers high-level questions regarding whether write-buffering could make a difference in making additional eNVMs viable for applications with significant write traffic, and, if so, how much benefit would need to be extracted?

For illustrative purposes, we consider a simple write cache that would hold write requests to the eNVM, write back to eNVM when the buffer is full, and allow in-place updates in the case of multiple writes to the same address before an update to eNVM. Figure 4.15 shows the results for this study for SPEC2017 and Facebook-Graph-BFS. Just buffering the writes will mask the effective write latency experienced by the system, while a write cache that allows updates could additionally reduce traffic and extend lifetime. In particular, we look at the effects of masking write latency and reducing write traffic on total memory latency and power. We observe that for Facebook-Graph-BFS, if the write traffic load is reduced by at least half, FeFET emerges as a performant option, while STT remains the lowest power solution for this particularly high-traffic workload. STT and RRAM are still the optimal technology choices for SPEC2017 in terms of performance, but write-buffering could empower FeFETs as a lower-power alternative if latency could be masked or write traffic to the eNVM could be reduced by at least 25%. Write buffering could additionally benefit RRAM-based solutions by effectively extending the memory lifetime in the face of otherwise-problematic endurance characteristics.

4.4.3 Revisiting Fault Models, Error Mitigation

While programming multiple bits per memory cell is an important strategy for increasing storage density across many eNVMs, previous work has revealed that MLC eNVMs may exhibit significantly higher fault rates that must be carefully considered in conjunction with application re-

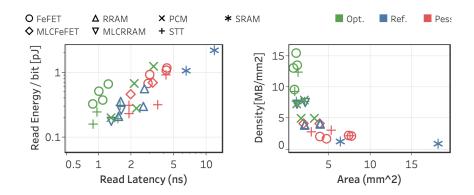


Figure 4.16: When we consider multi-level cells (MLC)s and filter out memory solutions that don't provide acceptable ResNet18 inference accuracy after fault injection, we note MLC RRAM is denser and more performant than SLC RRAM, while MLC FeFET is only sufficiently reliable for larger cell sizes (red), taken from ¹⁸⁵.

silience ^{183,204}. NVMExplorer enables efficient and broad probing of reliability vs. storage density by providing an application-agnostic fault injection tool and templates for technology-specific fault models. To demonstrate, we quantify the application accuracy for ResNet18 image classification under weight storage in SLC vs. 2-bit MLC across multiple technologies for which there exists sufficient cell and circuit level data to produce detailed fault models. The density vs. reliability trade-off is distinct for each technology. For example, Figure 4.16 displays 8MB and 16MB characterized arrays, including 2-bit MLC RRAM and 2-bit MLC FeFET, filtered such that only those arrays meeting application latency requirements and maintaining image classification accuracy are included. Note that these results replicate previous efforts that indicate that image classification inference is robust to 2-bit MLC RRAM storage (we also verified this for CTT-based memories with fault modeling details provided in ^{183,60}), while we show that MLC FeFET devices only exhibit acceptable accuracy for larger cell sizes. This is because smaller FeFETs are more difficult to program reliably due to device-to-device variation ²⁰⁴. Portions of NVMExplorer were leveraged to quantify cell- and circuit-level trade-offs specific to MLC FeFETs in greater depth to determine optimal cell provisioning and writing schemes for target applications ²⁰⁴.

I never am really satisfied that I understand anything; because, understand it well as I may, my comprehension can only be an infinitesimal fraction of all I want to understand about the many connections and relations which occur to me.

Ada Lovelace

5

Conclusion: Future Memory System Opportunities and Innovations

As technology scaling reaches physical limits and critical applications become increasingly data-intensive, fundamental improvements in efficiency are bound to effective and innovative memory system design choices. This thesis has reviewed and revealed a variety of concrete de-

sign choices and system opportunities for improving memory density and efficiency, particularly highlighting the promise of integrating embedded non-volatile memory solutions in a variety of computing systems. In addition to specific system solutions and key design choices, a central contribution of this work is to develop and make available methodologies and tools to empower future exploration of technology innovations, application characteristics, system design, and the critical intersections and interactions of these choices across the computing stack.

Several key themes have emerged in the discussion of this dissertation which will continue to be critical to future memory system design. In this section, I will describe the themes of reliability, flexibility, and technology-awareness and how they can and should inform future work in heterogeneous memory systems. Additionally, I will introduce other innovations and design considerations, such as temperature-aware evaluation, floorplan-aware memory architecture design for accelerators, environmentally-aware analysis of technology and fabrication choices, and exciting integration choices such as 3D-stacked memory and chiplet-based systems, and describe how future opportunities in these research spaces intersect and are made more immediately explorable, tangible, and feasible by the methods, tools, and processes described in this thesis.

5.0 Themes to Unlock Future Memory Efficiency

In this section, I review several over-arching design principles that emerged through the discussion of my dissertation work, and I comment on how these themes and principles will continue to shape the field of computer architecture, including some specific projects and studies I have considered that are in-progress or otherwise outside the scope of this work.

5.0.0 RELIABILITY

Chapter 2 probes the many ways not only that technology-level and system-level choices (e.g., MLC programming, data encoding) have direct and potentially catastrophic impacts on target application (in terms of accuracy, performance, and more), but also that exposing and exploring the distinct application-level impacts of varying system and technology reliability in the memory system can unlock incredible benefits in terms of increased memory density and memory system efficiency. These findings were concretized and integrated in a broader set of memory system evaluations in Chapters 3 and 4. In fact, as computing systems naturally gravitate towards increased specialization and heterogeneity in the face of the end of Moore's law, critical examination of reliability assumptions prompts a huge opportunity to *relax* and/or *customize* storage settings, data formats, and technology choices towards the specific needs and scope of particular applications and domains of interest.

This thesis has heavily focused on the cross-stack consideration of potentially decreased reliability in emerging memory technologies in conjunction with the relative fault tolerance of deep neural network inference as an application space relevant to myriad computing domains and systems. However, presented work in general-purpose compute settings (Chapter 4.3) and towards acceleration of graph processing and broader classes of machine learning (Chapters 4.2, 3.2, 3.3) suggest that the bleeding edge of memory efficiency in a variety of domains and system settings will require the proposed approach to re-thinking and re-calibrating resilience analysis and mitigation strategies in tight connection with memory technology design choices and application properties. While there is evidence that the specific software tools for fault modeling, fault injection, and incorporating reliability into system-level design objectives that I've contributed to and developed through my PhD work (namely, Ares 193 and nvmFI within NVMExplorer 185) will continue to carry the frontier and enable fascinating and essential future studies in memory reliability, I hope that the exposure and discussion of the interactions between data format (including numerical representations and sparse

encodings) and low-level memory cell and circuit optimizations (e.g., MLC programming and sensing design) will exhibit lasting impact on future memory solutions.

5.0.1 FLEXIBILITY

While a significant portion of the system solutions and memory architectures proposed throughout this thesis are heavily customized towards a single application or a subset of an application space, an increasingly crucial vector of any proposed architecture is the balance of customization via co-design vs. flexibility to support a sufficient range of workloads. The identification and segmentation of a 'sufficient range' of supported applications or workloads is highly system-dependent, and increasingly heterogeneous systems require careful consideration of not only what to accelerate and dedicate computational units towards, but also where and how memory should be allocated, physically distributed on a chip, and accessed and shared among system resources to best support end-to-end applications. In this light, the definition of flexibility alludes not only to a range of application characteristics, but also to the programmability and abstraction built into a customized system in terms of system support (e.g., user interface, development stack, memory management protocols) and circuit-level support (e.g., dynamic allocation and selection of programming settings, perhaps based on reliability concerns).

Ongoing work aims to taxonomize and sufficiently standardize both the design process for the next generation of accelerators and the appropriate dataflow, memory allocation, and scheduling paradigms to maximize utilization and efficiency ²²⁴. However, this flavor of co-design is as yet insufficient without critically examining data movement energy and opportunities for increased overall energy efficiency in the form of decoupled, physically distributed memory banks that are interchangeable and dynamically allocatable to emulate the memory hierarchy and partitioning that best supports a given workload of interest at compile-time. I'm investigating such a proposal in an orthogonal but conceptually relevant research effort that aims to identify where and when the the

fracturing and physical distribution of a memory array architecture is worthwhile. In this proposal, smaller, individually accessible and addressable buffer units are cheaper per-access, but the relative cost and lack of area efficiency of physically distributed memory resources among compute units must be overcome by a combination of sufficiently stripped-down network routing (made possible by mostly-static scheduling paradigms ¹⁸²) and effective data mapping to maximize data reuse to physically co-located data buffers. A key finding in this effort is that the individual buffer units, to maximize potential efficiency, must be flexibly able to be defined and partitioned according to application properties, and then data allocation must be optimally identified during compile-time, through a cross-computing-stack design methodology.

5.0.2 TECHNOLOGY-AWARENESS

Across a variety of contexts and leveraging a range of underlying memory technologies, this thesis has reviewed and evaluated concrete strategies to expose or otherwise customize lower-level design choices (i.e., memory cell, circuit, array optimizations) to the broader system architecture and even to the design and optimization of an application of interest to unlock striking gains. While reliability and flexibility are two critical considerations that can and should be built into future design goals and metrics, the core of computer architecture innovation and memory system efficiency in the future will be exploring how, when, and to what extent to bring awareness of technology-level details into the fore. The results presented throughout this thesis unearth the ways in which this is exacerbated as one considers integrating eNVMs, but the perspective of co-design and technology-awareness is similarly critical in understanding, adapting, and deriving benefits from advanced system design, fabrication, and integration choices more broadly, as discussed next.

5.1 Innovations on the Horizon, and How to Leverage Them

In this section, I briefly review opportunities and limitations for a set of broad future research thrusts, in each case contextualizing in terms of their viability for heterogeneous future memory systems and the efficacy of the methods and findings presented in this dissertation.

5.1.0 Temperature-Aware System Analysis and Evaluation

One design parameter largely overlooked in the preceding chapters is the impact of temperature. There are several ways in which thermal effects may impact analysis of the viability, effectiveness, and efficiency of proposed memory system solutions. For example, recent work observed that hotspots and uneven heat distribution in CPUs are exacerbated at advanced technology nodes (e.g., 7nm vs. 14nm or 16nm), leading to performance degradation, reliability issues, and potential damage to components ⁸⁵. Further study is required to know whether such issues may be compounded or possibly alleviated by replacing high-leakage-power SRAM with an eNVM, and whether sensitivity of temperature fluctuations would impact eNVM susceptibility to faults in different system contexts. Additionally, environmental conditions and cooling methods of the system can have significant performance and power implications as well. In fact, recent work has proposed cryogenic operation (i.e., cooling an entire chip via contact with liquid nitrogen) as a potentially power-efficient solution for off-chip DRAM and even for the CPU, drastically reducing the effective leakage and improving performance of the cache subsystem ¹⁶³.

As additional innovations and system proposals continue to emerge, it is essential to incorporate temperature-dependencies and to contextualize potential benefits in a broader system context. For example, future work should investigate whether potential power overheads of cooling can be sufficiently reduced to justify cryogenic operation, and how relative power benefits of cryo-operation may compare to the energy efficiency and low leakage of eNVM solutions for the LLC.

5.1.1 Environmental-Impact-Aware System Analysis and Evaluation

In the face of our changing climate, the indisputably high carbon footprint of fabrication, development, and use of technology should be front-of-mind of researchers aiming to shape and inform the next generation of computing systems. Moore's Law describing the number of transistors per integrated chip has always, at heart, been an economic argument, in terms of justifying the research and development towards increased density and performance, but the time is past for technology scaling efforts and innovations to be dually justified by economic incentives and environmental impacts.

Ongoing work has identified design challenges and opportunities in quantifying carbon footprint and designing future devices with environmental impact in-the-loop ^{80,245}. The modeling efforts recently put forth to quantify environmental impacts of computing can and should be integrated with the design tools for memory systems put forth in this thesis.

As yet, the environmental impact both in terms of embodied carbon and operation-related footprint of devices integrating eNVMs has not been analyzed, and shaping optimization goals and
metrics to include minimization of carbon emissions is a crucial and imminent change in mindset.

Quantifying and justifying changes to manufacturing, procurement of unique materials, and integration choices become central questions in judging the viability, promise, and limitations of each
of the eNVMs studied in this thesis. For example, one proposal to elide the costs and ensuing carbon footprint of fabrication in advanced technology nodes is to instead roll back to a less advanced
node (e.g., 22nm-45nm) for the dual benefit of reduced complexity and resources in manufacturing,
as well as research and development, and lower power density during operation. Considering the
open questions in terms of efficient scaling across memory technologies and system solutions, there
is much ground to cover and many delicate, cross-computing-stack interactions to consider in reducing carbon footprint of end-to-end systems. Namely, increasing conversation and collaboration
from materials engineers and industry fabrication plants up to computer architects and application

developers could change future computing systems by leveraging device-workload interactions and identifying potential concessions in the hallowed metrics of performance and cost in order build greener technology.

5.1.2 3D-INTEGRATED MEMORY

3D-stacked dies as well as monolithic 3D integration of memory (either standalone or integrated with compute layers) exhibits exciting potential performance and density benefits, but comes with design challenges ^{41,73}. Probing this trade-off per-memory-technology is a critical first step, but realistic analysis of end-to-end scalability, efficiency, and environmental impact will be needed. Many proposals already exist in this space and enable dramatic bandwidth and significant performance, but such systems must be carefully calibrated to identify when and where it is an appealing design choice (e.g., in light of potential thermal implication, potential additional costs, and more). There are ample, near-future opportunities to identify co-design choices in the space of 3D-stacked and monolithic 3D implementations of various memory technologies to augment the memory subsystem of both general purpose and highly specialized computing systems.

5.1.3 Chiplets – Implications for Memory Design Space Exploration

There are many well-documented limitations of large, monolithic chip designs, both in terms of yield and fabrication process at advanced technology nodes and in the drastically increasing complexity, time, and engineering effort involved to design and validate such a design. Multi-Chip Modules (MCMs), is a compelling emerging alternative design paradigm comprised of individually designed and fabricated chiplets, then integrated into a larger system ²⁰². A chiplet-based design introduces challenges in the efficiency of inter-chiplet communication, but has been demonstrated to scale well in supporting applications requiring beefy compute and high on-chip memory capacity, as

for deep neural network inference.

A chiplet-based integration strategy is also an opportunity to combine and communicate among distinct chip designs, including those fabricated with different processes or those dedicated for different designs and purposes (e.g., a subset of distributed chiplets for technologically heterogeneous memory resources, such as those integrating different NVM technologies). For this reason, MCMs are a compelling design paradigm for integrating and making efficient use of tightly-coupled eNVM resources in a large system, and the proposed tools and methodologies in this dissertation could be extended to model the interactions and access properties of a chiplet-based design. The relevance of varying reliability, flexibility, and technology-awareness are heightened in this context. The critical intersections of application access characteristics, data formats, and network properties unveiled in this work will remain relevant to these critical design space explorations and system developments.

5.2 Summary: Cross-Computing-Stack Memory Efficiency Opportunities

Future memory system efficiency will clearly be guided by reliability, flexibility, and technology-awareness. The work presented in this dissertation can be effectively complemented by improved metrics and methodologies (e.g., quantifying environmental impact of design choices, incorporating temperature-awareness and integration innovations). Continuing the attitudes and design methods developed in this thesis will be critical to transformations and efficiency gains in future memory systems. In particular, exposing technology characteristics to higher levels of the computing stack necessitates conversation and approachable methods for collaborative design among application experts, system architects, circuit designers, device physicists, and more. Building accessible, extensible, and thoughtful frameworks and software tools for design space exploration and technology modeling is a bridge towards a more diverse and creative era of memory architecture solutions, to which I'm excited to shape and contribute as a researcher.

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